



SVR ENGINEERING COLLEGE

Approved by AICTE & Permanently Affiliated to JNTUA

Ayyalurmetta, Nandyal – 518503. Website: www.svrec.ac.in

Department of Electronics and Communication Engineering



(19A04402P) ELECTRONIC CIRCUITS –ANALYSIS AND DESIGN LABORATORY

II B.Tech (ECE) - II Semester - 2020-21 – R19



STUDENT NAME	
ROLL NUMBER	
SECTION	



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DEPARTMENT OF **ELECTRONICS AND COMMUNICATION ENGINEERING**

CERTIFICATE

ACADEMIC YEAR: 2020-21

This is to certify that the bonafide record work done by

Mr./Ms. _____ bearing

H.T.NO. _____ of II B. Tech II Semester in the

Electronic Circuits- Analysis and Design Laboratory

Faculty In-Charge

Head of the Department

LIST OF EXPERIMENTS:

1. MOSFET Amplifier
 - a. Design and simulate MOSFET (Depletion mode) amplifier using PSPICE /Multisim and study the Gain and Bandwidth of amplifier
 - b. Design common source MOSFET (Enhance mode) amplifier with discrete components and calculate the bandwidth of amplifier from its frequency response
2. JFET Amplifier
 - a. Design and simulate common source FET amplifier using PSPICE /Multisim and study the Gain and Bandwidth of amplifier
 - b. Design common source FET amplifier with discrete components and calculate the bandwidth of amplifier from its frequency response
3. Common Emitter Amplifier (Self bias Amplifier)
 - a. Design and simulate a self- bias (Emitter bias) Common Emitter amplifier using PSPICE /Multisim and study the Gain and Bandwidth of amplifier
 - b. Design voltage divider based Common Emitter amplifier with discrete components and calculate the bandwidth of amplifier from its frequency response.
4. Design and simulate two stage RC coupled amplifier for given specifications. Determine Gain and Bandwidth from its frequency response curve.
5. Design and simulate Darlington amplifier. Determine Gain and Band width from its frequency response curve.
6. Design and Simulate CE – CB Cascode amplifier. Determine Gain and Bandwidth from its frequency response curve
7. Design and simulate voltage series feedback amplifier for the given specifications. Determine the effect of feedback on the frequency response of a voltage series feedback amplifier.
8. Design and simulate current shunt feedback for the given specifications. Determine the effect of feedback on the frequency response of a current shunt feedback amplifier.
9. Design and simulate RC Phase shift oscillator and Wien bridge oscillator for the given specification. Determine the frequency of oscillation.
10. Design and simulate Hartley and Colpitts oscillators for the given specifications. Determine the frequency of oscillation.
11. Design and simulate class A power amplifier and find out the efficiency. Plot the output waveforms.
12. Design and simulate class B push-pull amplifier and find out the efficiency. Plot the output waveforms.
13. Design and simulate single tuned amplifier. Determine the resonant frequency and bandwidth of a tuned amplifier.
14. Design and simulate double tuned amplifier. Determine the resonant frequency and bandwidth of a tuned amplifier.

Note: Design & simulate any 12 experiments with Multisim / PSPICE or equivalent software and verify the results in hardware lab with discrete components.

ECE DEPT VISION & MISSION PEOs and PSOs

Vision

To produce highly skilled, creative and competitive Electronics and Communication Engineers to meet the emerging needs of the society.

Mission

- Impart core knowledge and necessary skills in Electronics and Communication Engineering through innovative teaching and learning.
- Inculcate critical thinking, ethics, lifelong learning and creativity needed for industry and society
- Cultivate the students with all-round competencies, for career, higher education and self-employability

I. PROGRAMME EDUCATIONAL OBJECTIVES (PEOS)

PEO1: Graduates apply their knowledge of mathematics and science to identify, analyze and solve problems in the field of Electronics and develop sophisticated communication systems.

PEO2: Graduates embody a commitment to professional ethics, diversity and social awareness in their professional career.

PEO3: Graduates exhibit a desire for life-long learning through technical training and professional activities.

II. PROGRAM SPECIFIC OUTCOMES (PSOS)

PSO1: Apply the fundamental concepts of electronics and communication engineering to design a variety of components and systems for applications including signal processing, image processing, communication, networking, embedded systems, VLSI and control system

PSO2: Select and apply cutting-edge engineering hardware and software tools to solve complex Electronics and Communication Engineering problems.

III. PROGRAMME OUTCOMES (PO'S)

- 1. Engineering knowledge:** Apply the knowledge of mathematics, science, engineering fundamentals, and an engineering specialization to the solution of complex engineering problems.
- 2. Problem analysis:** Identify, formulate, review research literature, and analyze complex engineering problems reaching substantiated conclusions using first principles of mathematics, natural sciences, and engineering sciences.
- 3. Design/development of solutions:** Design solutions for complex engineering problems and design system components or processes that meet the specified needs with appropriate consideration for the public health and safety, and the cultural, societal, and environmental considerations.
- 4. Conduct investigations of complex problems:** Use research-based knowledge and research methods including design of experiments, analysis and interpretation of data, and synthesis of the information to provide valid conclusions.
- 5. Modern tool usage:** Create, select, and apply appropriate techniques, resources, and modern engineering and IT tools including prediction and modeling to complex engineering activities with an understanding of the limitations.
- 6. The engineer and society:** Apply reasoning informed by the contextual knowledge to assess societal, health, safety, legal and cultural issues and the consequent responsibilities relevant to the professional engineering practice.
- 7. Environment and sustainability:** Understand the impact of the professional engineering solutions in societal and environmental contexts, and demonstrate the knowledge of, and need for sustainable development.
- 8. Ethics:** Apply ethical principles and commit to professional ethics and responsibilities and norms of the engineering practice.
- 9. Individual and team work:** Function effectively as an individual, and as a member or leader in diverse teams, and in multidisciplinary settings.
- 10. Communication:** Communicate effectively on complex engineering activities with the engineering community and with society at large, such as, being able to comprehend and write effective reports and design documentation, make effective presentations, and give and receive clear instructions.
- 11. Project management and finance:** Demonstrate knowledge and understanding of the engineering and management principles and apply these to one's own work, as a member and leader in a team, to manage projects and in multidisciplinary environments.
- 12. Life-long learning:** Recognize the need for, and have the preparation and ability to engage in independent and life-long learning in the broadest context of technological change.

IV. COURSE OBJECTIVES

- To provide a practical exposure for design & analysis of electronic circuits for generation and amplification of input signal.
- To learn the frequency response and finding gain, input & output impedance of multistage amplifiers
- To Design negative feedback amplifier circuits and verify the effect of negative feedback on amplifier parameters.
- To understand the application of positive feedback circuits & generation of signals.
- To understand the concept of design and analysis of Power amplifiers and tuned amplifiers
- To construct and analyze voltage regulator circuits.

V. COURSE OUTCOMES

After the completion of the course students will be able to

Course Outcomes	Course Outcome statement	BTL
CO1	Understand Characteristics and frequency response of various amplifiers	L1
CO2	Analyze negative feedback amplifier circuits, oscillators, Power amplifiers, Tuned amplifiers.	L3
CO3	Determine the efficiencies of power amplifiers.	L2
CO4	Design RC and LC oscillators, Feedback amplifier for specified gain and multistage amplifiers for Low, Mid and high frequencies.	L4
CO5	Simulate all the circuits and compare the performance.	L5

VI. COURSE MAPPING WITH PO'S AND PEO'S

Course Title	P01	P02	P03	P04	P05	P06	P07	P08	P09	P010	P011	P012	PS01	PS02
Electronic circuits- Analysis and Design	2.8	2.6	2.4	2.4	2.2	1.8	1.4	1.2	1.6	1.0	2.2	1.6	2.2	1.8

VII MAPPING OF COURSE OUTCOMES WITH PEO'S AND PO'S

Course Title	P01	P02	P03	P04	P05	P06	P07	P08	P09	P010	P011	P012	PS01	PS02
CO1	3	3	3	2	3	1	1	1	2	1	3	2	3	2
CO2	2	2	2	2	1	1	2	1	1	1	2	1	3	3
CO3	3	3	2	3	2	3	2	1	2	1	2	2	2	1
CO4	3	2	3	2	2	2	1	2	1	1	3	1	2	2
CO5	3	3	2	3	3	2	1	1	2	1	1	2	1	1

I N D E X

Max. Marks per each experiment : 5

Sl. No.	Name of the Experiment	Page No.	Date Of Performed	Date Of Submission	Marks Obtained	Signature of Lab incharge
	Off the Syllabus :	----	-----	-----	----	-----
-----	Using Simulation software & Hardware	----	-----	-----	----	-----
1	JFET common source amplifier	9				
2	BJT-Common Emitter amplifier	15				
3	Two stage RC coupled amplifier	21				
4	Darlington pair amplifier	27				
5	CE-CB Cascode amplifier	31				
6	Voltage series feedback amplifier	37				
7	Current shunt feedback amplifier	43				
8	Single tuned voltage amplifier	49				
9	Class-A Series FED power amplifier	57				
10	Complementary symmetry Class B push-pull poweramplifier	63				
11.A	RC Phase shift Oscillator	69				
11.B	Wein Bridge Oscillator	75				
12.A	Colpitt's oscillator	81				
12.B	Hartley Oscillator	87				
	Total Marks obtained :					
	Average Marks obtained :					
	Beyond the Syllabus :	----	-----	-----	----	-----
13.	Bootstrapped Emitter follower	93				
14.	Astable Multivibrator using Transistors	99				

----- Index Continued -----

Experiment No. : 1

Date :

Name of the Experiment : FET - COMMON SOURCE (CS) AMPLIFIER

AIM :

- 1). To obtain the frequency response of *Common Source FET amplifier* using Software and Hardware
- 2). To calculate the band width of this amplifier.

APPARATUS :

Software :

1. System ----- 1 No.
2. Multisim software

Hardware :

- 1). Function generator(*FG*) ----- 1 No.
- 2). Cathode Ray Oscilloscope(*CRO*) ----- 1 No.
- 3). Regulated Power Supply (*RPS*) : Dual channel, (0-30)V, 1A ----- 1 No.
- 4). Probes ----- 1 No.
- 5). Bread board ----- 1 No.
- 6). Connecting wires : ----- A few Nos.

COMPONENTS :

- 1). Transistor BF 245 / BF W11 ----- 1 No.
- 2) Carbon fixed resistors
 - a). 1.8K Ω , ½W ----- 1 No.
 - b). 2.2K Ω , ½W ----- 2 No.
 - c). 100K Ω , ½W ----- 1 No.
- 3). Capacitors
 - a). 0.22 μ F ----- 2 No.
 - b). 33 μ F ----- 1 No.

THEORY :

Small signal amplifiers can also be made using Field Effect Transistors. These devices have the advantage over bipolar transistors of having an extremely high input impedance along with a low noise output making them ideal for use in amplifier circuits that have very small input signals.

The design of an amplifier circuit based around a junction field effect transistor or “JFET”, (N-channel FET for this tutorial) or even a metal oxide silicon FET or “MOSFET” is exactly the same principle as that for the bipolar transistor circuit used for a Class A amplifier circuit we looked at in the previous tutorial.

Firstly, a suitable quiescent point or “Q-point” needs to be found for the correct biasing of the JFET amplifier circuit with single amplifier configurations of Common-source (CS), Common-drain (CD) or Source-follower (SF) and the Common-gate (CG) available for most FET devices.

Common Source JFET Amplifier **as** this is the most widely used JFET amplifier design.

The amplifier circuit consists of an N-channel JFET, but the device could also be an equivalent N-channel depletion-mode MOSFET as the circuit diagram would be the same just a change in the FET, connected in a common source configuration. The JFET gate voltage V_g is biased through the potential divider network set up by resistors R1 and R2 and is biased to operate within its saturation region .

The junction FET takes virtually no input gate current allowing the gate to be treated as an open circuit. Then no input characteristics curves are required.

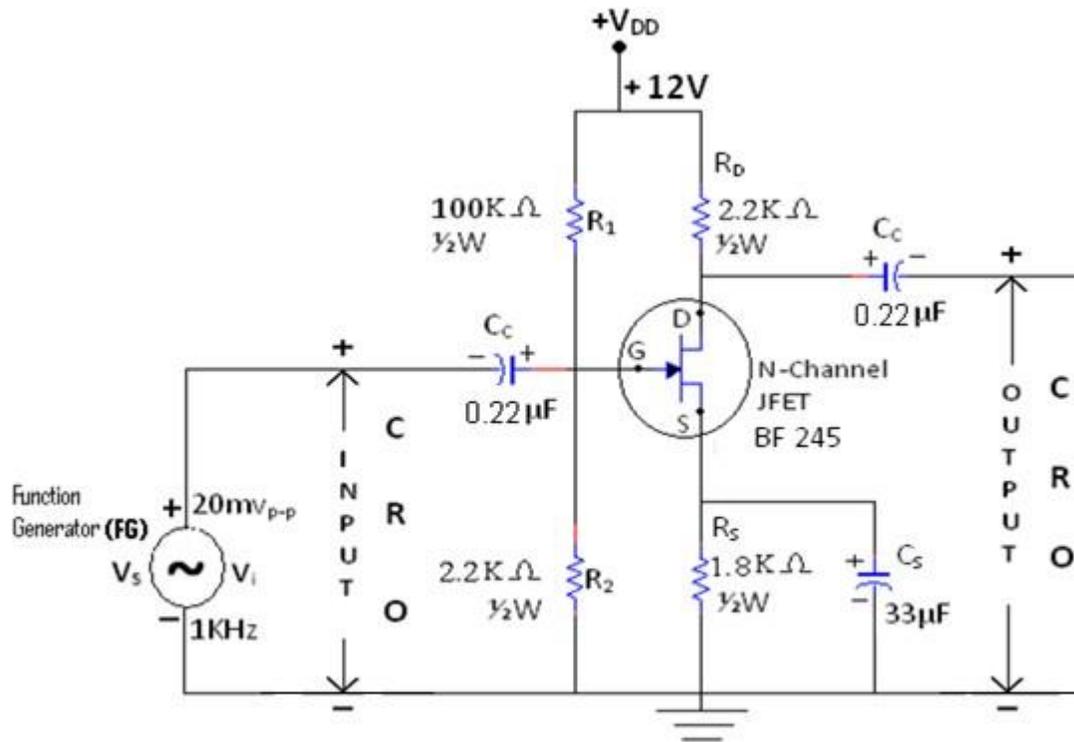
CIRCUIT DIAGRAM – SOFTWARE & HARDWARE :

Figure : Circuit diagram of Common Source FET Amplifier

PROCEDURE – SOFTWARE :

We have picked up the components from the components bar as per above circuit.

2. Made the connections as per the above circuit diagram by using the components which we have picked up.
3. Set the input signal as *sine wave form which is having the value 20mV_{P-P}* as constant in the function generator.
4. Initially set the input signal frequency value is 1KHz in the function generator.
5. To simulate the circuit clicked on *run option* through *execute button* in *tool bar*.
6. We have seen the *sine wave* on the **CRO** screen as o/p signal.
7. Calculated the *peak to peak voltage (V_{O(p-p)})* and noted down in the tabular form Against the column of 1KHz.
8. Stopped the simulation by clicked on *run option* through *execute button* in the *tool bar*.
9. Repeat the same procedure from points 7 to 9 for the corresponding frequency values by setting in the function generator for the following steps,
20Hz, 100Hz, 200Hz, 1KHz, 200KHz, 400KHz, 600KHz, 1180KHz, 1MHz, 100MHz, 500MHz. in the function generator.
10. Observed the graph for *frequency Vs amplitude* through the *AC Analysis*.
11. Finally shut down the system safely.
12. We have observed that, the graph which is drawn by manually is same to the graph which is obtained from the *AC Analysis*.
13. Now calculated and noted down the values of *voltage gain (A_V)* and *gain in dB* to the corresponding values of *output voltage (V_O)* & *input voltage (V_i)* by using the formulas given below,
$$\text{Voltage gain } (A_v) = V_o / V_i \quad \text{and} \quad \text{Gain in dB} = 20 \log_{10}(A_v)$$
14. Plotted the graphs (frequency response curves) as per

below a). frequency on X-axis & gain in dB on Y-axis.

PROCEDURE – HARDWARE :

- 1). Connected the circuit as per the circuit diagram.
- 2). Then switched ON the *function generator* and *CRO*; but don't switched ON the *RPS*.
- 3). Now Kept the *AC/GND/DC* switch is at *AC* position.
- 4). Initially kept the 1KHz. frequency by varying the frequency control in the *function generator*.
- 5). Now applied the peak to peak amplitude of a sine wave is of $20mV_{p-p}$ by varying the amplitude control in the *function generator* through observing in the *CRO*.
- 6). Kept this input value as $20mV_{p-p}$ constant up to the completion of the experiment
Otherwise the wrong output would occurred.
- 7). Now switched ON the *RPS* and set the 10V in it i.e. $V_{CC} = 10V$.
- 8). Varied the different frequency steps of 20Hz, 100Hz, 200Hz, 1KHz, 200KHz, 400KHz, 600KHz, 1180KHz, 1MHz.
by adjusted the frequency control in the *function generator* and noted down the corresponding values of output signal i.e. peak to peak amplitude of sine wave by observing in the *CRO*.
- 9). Now switched OFF the *RPS*, *function generator* and *CRO*.
- 10). Then calculated the *voltage gain* $A_v = V_o/V_i$ & *gain in dB* $= 20\log_{10}(A_v)$ and noted down the values in the specified columns of the tabular column.
- 11). Plotted the graphs (frequency response curves) as per below, a). frequency on X-axis & gain in dB on Y-axis.
b). frequency on X-axis & voltage gain on Y-axis.
- 12). Calculated the *band width* from the above two (frequency response curves) graphs by using the formula $f_2 - f_1$ which is given under the heading of *parameters*.

TABULAR COLUMNS – SOFTWARE & HARDWARE :

Input Voltage (V_i) = 20 mV_{P-P} is constant for all

readings For Software :

For Hardware :

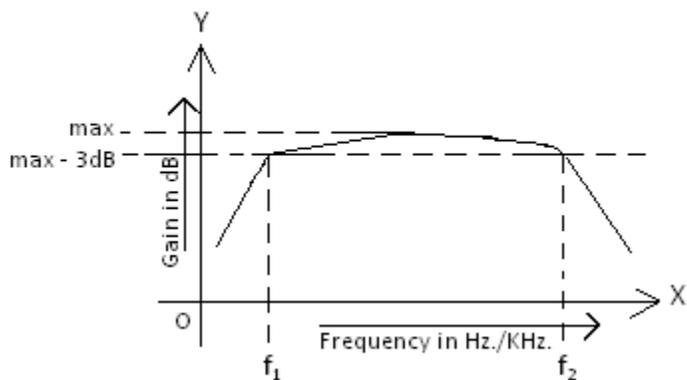
Sl. No.	Frequency In Hz/KHz.	Output Voltage (V_o) In mVolts.	Voltage gain $A_v = V_o/V_i$	Gain in dB = $20\log_{10}(A_v)$	Frequency In Hz/KHz.	Output Voltage (V_o) In mVolts.	Voltage gain $A_v = V_o/V_i$	Gain in dB = $20\log_{10}(A_v)$
1	20 Hz.							
2	100 Hz.							
3	200 Hz.							
5	1 KHz.							
6	200KHz.							
7	400KHz.							
8	600KHz.							
To be continued in next page								

Sl. No.	Frequency In Hz/KHz.	Output Voltage (V _o) In mVolts.	Voltage gain A _v = V _o /V _i	Gain in dB = 20log ₁₀ (A _v)		Frequency In Hz/KHz.	Output Voltage (V _o)In mVolts.	Voltage gain A _v = V _o /V _i	Gain in dB = 20log ₁₀ (A _v)
9	800KHz.								
10	1 MHz.								
11	100 MHz					-----	-----	-----	-----
12	500MHz.					-----	-----	-----	-----

EXPECTED GRAPHS – SOFTWARE & HARDWARE :

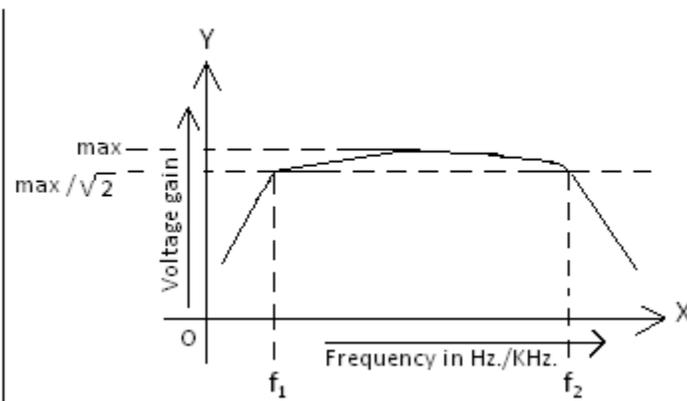
A). Frequency response curve

For frequency verses gain in dB.



B). Frequency response curve

For frequency verses voltage gain.



PARAMETERS – SOFTWARE & HARDWARE :

- 1). Band width of frequency response curve for frequency verses gain in dB \Rightarrow
 $= f_2 - f_1$
- 2). Band width of frequency response curve for frequency verses voltage gain
 $= f_2 - f_1 \Rightarrow$

RESULT – SOFTWARE & HARDWARE :

We have obtained the frequency response curves of *Common Source FET Amplifier (CSFET)* for frequency verses gain in dB & frequency verses voltage gain and calculated the band width of both of them. The band width values are given below,

- 1). Band width of frequency response curve for frequency verses gain in dB. =
- 2). Band width of frequency response curve for frequency verses voltage gain =

VIVA VOICE QUESTIONS:

1. What is the Difference between BJT and FET?

2. What is Amplifier?

3. What is Band Width?

4. What are the applications of CS FET Amplifier?

5. FET is which controlled device?

6. Mention FET characteristics.

7. What are the configurations of FET?

8. What are the classifications of FET?

9. Which configuration mostly used in FET?

10. What is dB?

Experiment No. : 2

Date :

Name of the Experiment : **BJT - COMMON EMITTER (CE) AMPLIFIER****AIM :**

1). To obtain the frequency response of *Common Emitter amplifier* using Hardware and Software2). To calculate the band width of this amplifier.

APPARATUS :**Software :**

1. System 1 No.
2. Multisim software

Hardware :

- 1). Function generator(*FG*) 1 No.
- 2). Cathode Ray Oscilloscope(*CRO*) 1 No.
- 3). Regulated Power Supply (*RPS*) : (0-30)V, 1A Dual channel 1 No.
- 4). Probes 1 No.
- 5). Bread board 1 No.
- 6). Connecting wires : A few Nos.

COMPONENTS :

- 1). Transistor BC 547 1 No.
- 2) Carbon fixed resistors
 - a). 47K Ω , 1/2W 1 No.
 - b). 10K Ω , 1/2W 1 No.
 - c). 4.7 K Ω , 1/2W 1 No.
 - d). 1 K Ω , 1/2W 1 No.
- 3). Capacitors
 - a). 0.22 μ F 2 No.
 - b). 33 μ F 1 No.

THEORY :

A transistor in which the emitter terminal is made common for both the input and the output circuit connections is known as common emitter configuration. When this configuration is provided with the supply of the alternating current (AC) and operated in between the both positive and the negative halves of the cycle in order to generate the specific output signal is known as common emitter amplifier.

In this type of configuration the input is applied at the terminal base and the considered output is to be collected across the term Voltage Gain

The ratio of the output voltage generated when the input voltage applied decides the voltage gain of the common emitter amplifier.

Characteristics

The characteristics of the common emitter configuration amplifier configuration are as follows

The voltage gain value obtained for the common emitter amplifier is medium.

It also consists of the current gain in the medium range.

Because of both the voltage and the current gains the power gain value of this configuration is referred to be high.

There is some resistance value at the inputs as well as the output but in this configuration it is maintained at the medium value.

Applications

1. These amplifiers are preferably used as the current amplifier than a voltage amplifier as it has more current gain than the voltage gain.
2. In the radio frequency circuitry this configuration is preferred.
3. For the lower values of noise and its amplification this configuration is preferred.

CIRCUIT DIAGRAM – SOFTWARE & HARDWARE :

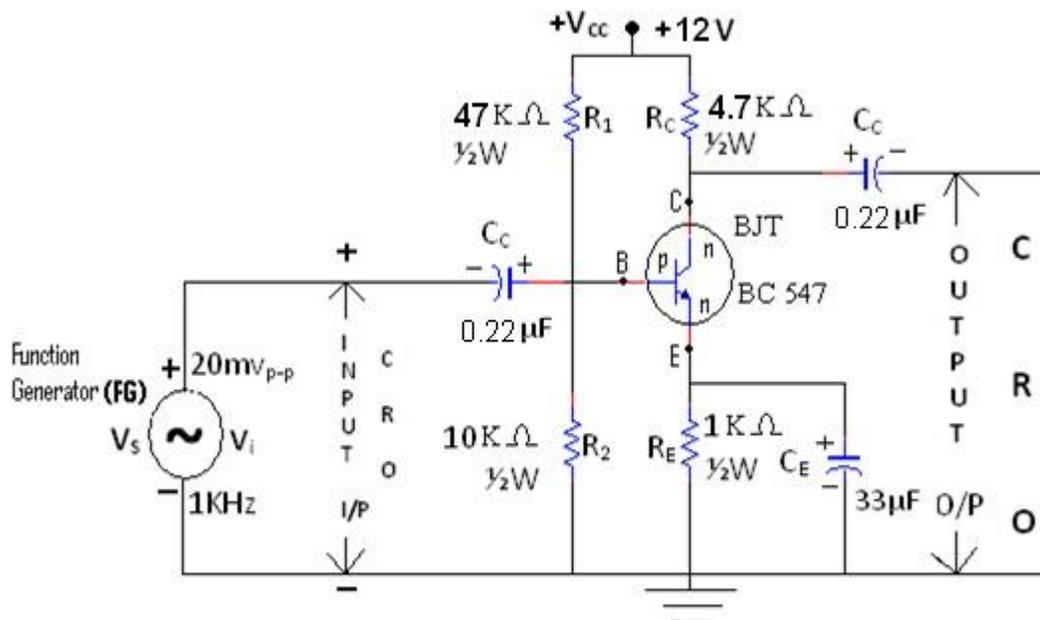


Figure: Circuit diagram of Common Emitter(CE) amplifier.

PROCEDURE – SOFTWARE :

1. We have picked up the components from the components bar as per above circuit.
2. Made the connections as per the above circuit diagram by using the components which we have picked up.
3. Set the input signal as *sine wave form* which is having the value $20\text{mV}_{\text{P-P}}$ as constant in the function generator.
4. Initially set the input signal frequency value is 1KHz in the function generator.
5. To simulate the circuit clicked on *run option* through *execute button* in *tool bar*.
6. We have seen the *sine wave* on the **CRO** screen as o/p signal.
7. Calculated the *peak to peak voltage* ($V_{O(p-p)}$) and noted down in the tabular form Against the column of 1KHz .
8. Stopped the simulation by clicked on *run option* through *execute button* in the *tool bar*.
9. Repeat the same procedure from points 7 to 9 for the corresponding frequency values by setting in the function generator for the following steps, 20Hz , 100Hz , 200Hz , 1KHz , 200KHz , 400KHz , 600KHz , 1180KHz , 1MHz , 100MHz , 500MHz . in the function generator.
10. Observed the graph for *frequency Vs amplitude* through the *AC Analysis*.
11. Finally shut down the system safely.
12. We have observed that, the graph which is drawn by manually is same to the graph which is obtained from the *AC Analysis*.

13. Now calculated and noted down the values of *voltage gain* (A_v) and *gain in dB* to the corresponding values of *output voltage* (V_o) & *input voltage* (V_i) by using the formulas given below,

$$\text{Voltage gain } (A_v) = V_o / V_i \quad \text{and} \quad \text{Gain in dB} = 20\log_{10}(A_v).$$

- 14). Plotted the graphs (frequency response curves) as per below
- frequency on X-axis & gain in dB on Y-axis.
 - frequency on X-axis & voltage gain on Y-axis.

PROCEDURE – HARDWARE :

- Connected the circuit as per the circuit diagram.
- Then switched ON the *function generator* and *CRO*; but don't switched ON the *RPS*.
- Now Kept the *AC/GND/DC* switch is at *AC* position.
- Initially kept the 1KHz. frequency by varying the frequency control in the *function generator*.
- Now applied the peak to peak amplitude of a sine wave is of $20\text{mV}_{\text{p-p}}$ by varying the amplitude control in the *function generator* through observing in the *CRO*.
- Kept this input value as $20\text{mV}_{\text{p-p}}$ constant up to the completion of the experiment
Otherwise the wrong output would occurred.
- Now switched ON the *RPS* and set the 10V in it i.e. $V_{\text{CC}} = 10\text{V}$.
- Varied the different frequency steps of 20Hz, 100Hz, 200Hz, 1KHz, 200KHz, 400KHz, 600KHz, 1180KHz, 1MHz. by adjusted the frequency control in the *function generator* and noted down the corresponding values of output signal i.e. peak to peak amplitude of sine wave by observing in the *CRO*.
- Now switched OFF the *RPS*, *function generator* and *CRO*.
- Then calculated the *voltage gain* $A_v = V_o/V_i$ & *gain in dB* $= 20\log_{10}(A_v)$ and noted down the values in the specified columns of the tabular column.
- Plotted the graphs (frequency response curves) as per below,
 - frequency on X-axis & gain in dB on Y-axis.
 - frequency on X-axis & voltage gain on Y-axis.
- Calculated the *band width* from the above two (frequency response curves) graphs by using the formula $f_2 - f_1$ which is given under the heading of *parameters*.

TABULAR COLUMNS :

Input Voltage (V_i) = 20 mV_{P-P} (0.02V) is constant for all readings.

For Software :

For Hardware :

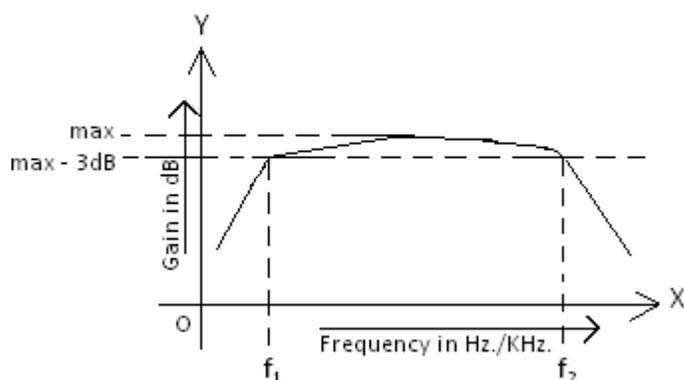
Sl.No.	Frequency In Hz/KHz.	Output Voltage (V_o) In mVolts.	Voltage gain $A_v = V_o/V_i$	Gain in dB = $20\log_{10}(A_v)$		Frequency In Hz/KHz.	Output Voltage (V_o) In mVolts.	Voltage gain $A_v = V_o/V_i$	Gain in dB = $20\log_{10}(A_v)$
1	20 Hz.								
2	100 Hz.								
3	200 Hz.								
5	1 KHz.								
To continued in next page									

Sl.No.	Frequency In Hz/KHz.	Output Voltage (V _O) In mVolts.	Voltage gain A _v = V _o /V _i	Gain in dB = 20log ₁₀ (A _v)		Frequency In Hz/KHz.	Output Voltage (V _O)In mVolts.	Voltage gain A _v = V _o /V _i	Gain in dB = 20log ₁₀ (A _v)
6	200KHz.								
7	400KHz.								
8	600KHz.								
9	1180KHz.								
10	1 MHz.								
11	100 MHz					-----	-----	-----	-----
12	500MHz.					-----	-----	-----	-----

EXPECTED GRAPHS – SOFTWARE & HARDWARE :

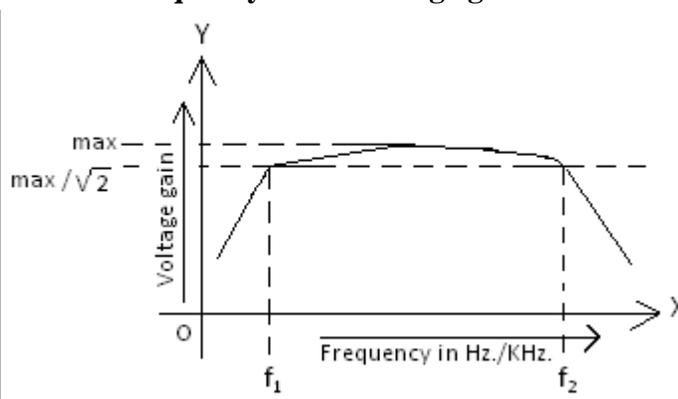
A). Frequency response curve

For frequency verses gain in dB.



B). Frequency response curve

For frequency verses voltage gain.



PARAMETERS – SOFTWARE & HARDWARE :

- 1). Band width of frequency response curve for frequency verses gain in dB. = $f_2 - f_1 =$
- 2). Band width of frequency response curve for frequency verses voltage gain = $f_2 - f_1 =$

RESULT –SOFTWARE & HARDWARE :

We have obtained the frequency response curves of *Common Emitter Amplifier (CE)* for frequency verses gain in dB & frequency verses voltage gain and calculated the band width of both of them. The band width values are given below,

- 1). Band width of frequency response curve for frequency verses gain in dB. =
- 2). Band width of frequency response curve for frequency verses voltage gain =

VIVA VOICE QUESTIONS:

1. What is BJT?

2. What are the applications of BJT?

3. What is Early Effect?

4. Define alpha and beta DC amplification factors of BJT.

5. Briefly explain reach through effect.

6. Draw the symbols for BJT.

7. Explain the transistor operation with the help of four regions

8. Explain base width modulation of a transistor

9. Compare CB, CE, CC configurations of a transistor.

10. A transistor has CE current gain of 100. If the collector is 40 mA. What is the value of emitter current?]

Experiment No. : 03

Date :

Name of the Experiment : TWO STAGE RC COUPLED AMPLIFIER

AIM :

To verify / plot the frequency response curve *and* to find the band width. of a *two stage RC coupled Amplifier* using software and hardware

APPARATUS :**Software :**

1. System..... 1 No.
2. Multisim software

Hardware :

- 1). Function generator(*FG*) ----- 1 No.
- 2). Cathode Ray Oscilloscope(*CRO*) ----- 1 No.
- 3). Regulated Power Supply (*RPS*) : (0-30)V, 1A Dual channel ----- 1 No.
- 4). Probes ----- 1 No.
- 5). Bread board ----- 1 No.
- 6). Connecting wires : ----- A few Nos.

COMPONENTS :

- 1). Transistor BC 547-----2 No.
- 2) Carbon fixed resistors
 - a). 47K Ω , 1/2W ----- 2 No.
 - b). 10K Ω , 1/2W ----- 2 No.
 - c). 4.7 K Ω , 1/2W ----- 2 No.
 - d). 1 K Ω , 1/2W ----- 2 No.
- 3). Capacitors
 - a). 0.22 μ F ----- 4 No.
 - b). 33 μ F ----- 2 No.

THEORY :

RC coupling is the most widely used method of coupling in multistage amplifiers. ... In this case the resistance R is the resistor connected at the collector terminal and the capacitor C is connected in between the amplifiers. It is also called a blocking capacitor, since it will block DC voltage.

Advantages :

The following are the advantages of RC coupled amplifier. The frequency response of RC amplifier provides constant gain over a wide frequency range, hence most suitable for audio applications. The circuit is simple and has lower cost because it employs resistors and capacitors which are cheap.

Gain :

The gain of an amplifier is increased by connecting the amplifiers in cascaded manner. The output of one stage is connected to the input of next stage through the coupling capacitor. It increases the overall gain of the amplifier and decreases the overall bandwidth of the amplifier.

Applications :

Optical Fiber Communications. Public address systems as pre-amplifiers. Controllers. Radio or TV Receivers as small signal amplifiers.2

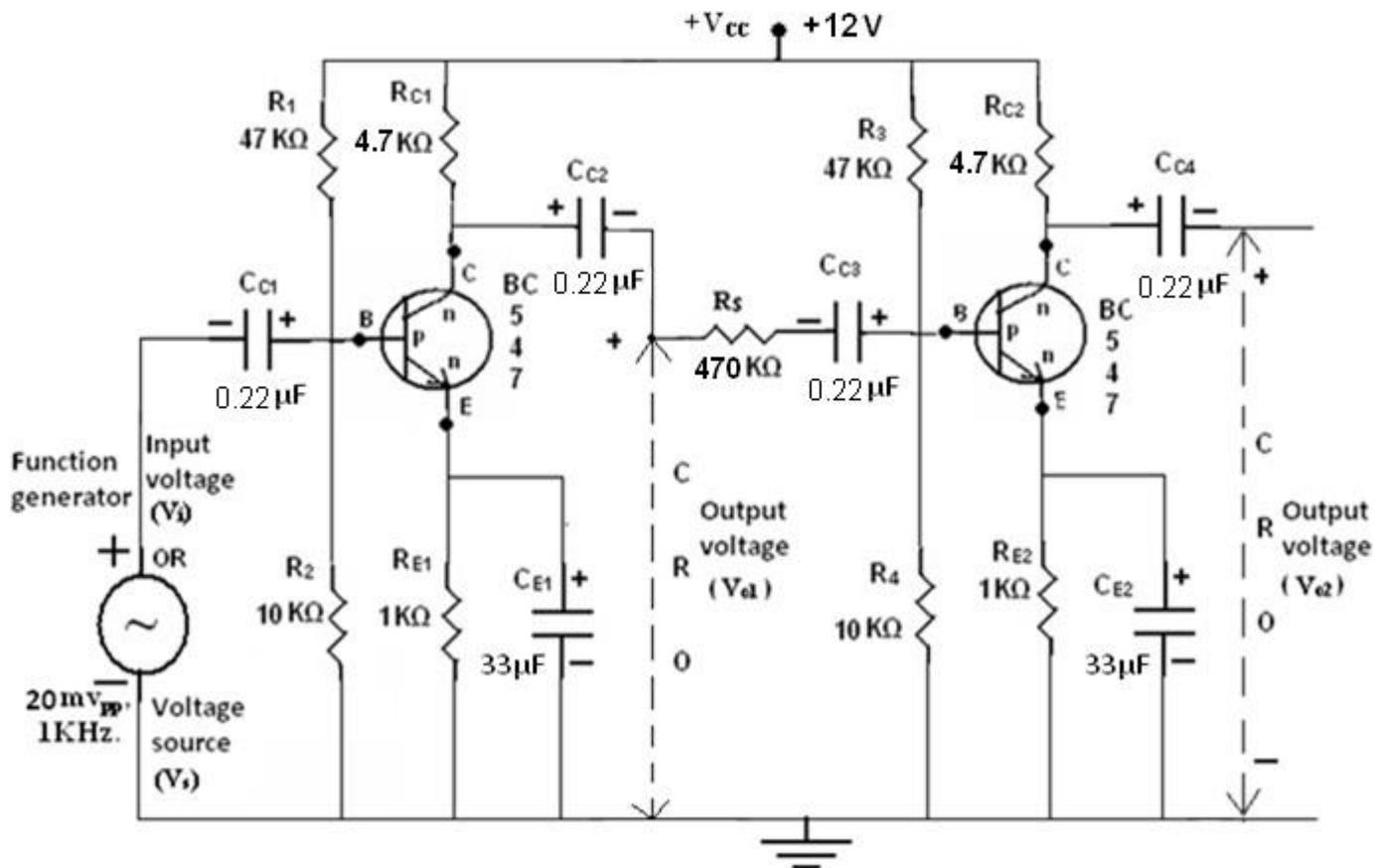
CIRCUIT DIAGRAM : SOFTWARE & HARDWARE :

Figure : Circuit diagram of Two stage RC coupled amplifier.

PROCEDURE - SOFTWARE :

1. We have picked up the components from the tool bar as per above circuit in Multisim software.
1. Made the connections as per the above circuit diagram by using the components which have picked up.
3. Connected the CRO across the capacitor C_{C2} .
4. Set the input signal value as $20\text{mV}_{\text{P-P}}$, 1KHz in the function generator as constant and V_{CC} as 12V .
5. To simulate the circuit click on execute / run button in tool bar.
6. We have seen the *Sine wave* on the CRO which is connected at o/p of the single stage as O/P signal.
7. Noted/ observed the readings for o/p voltage (Peak to Peak) of output signal in CRO by varying the different frequency steps (i.e. 20Hz , 100Hz , 200Hz , 500Hz , 1KHz , 100KHz , 200KHz , 400KHz , 600KHz , 1180KHz , 1MHz , 100MHz , 500MHz .) of the input AC signal in function generator.
8. Noted the above readings to the corresponding frequency steps in the tabular form of *Single stage RC Couple Amplifier*.
9. Stop the simulation by click on *Run button* in *tool bar*.
10. Now click on CRO which is connected at o/p of 2nd stage and click on *Run button*.
11. Noted the above readings to the corresponding frequency steps in the tabular form of *Two stage RC Couple Amplifier*.
12. Stop the simulation by click on *Run button* in *tool bar*.

13. Observed the graph *frequency Vs amplitude* through the AC Analysis for *Two stage RC Coupled Amplifier*.
14. Shut down the system safely.
15. Calculated and noted the Voltage gain by using the formula of V_o / V_i and Gain in dB by using the formula of $20\log_{10}(A_v)$ in the tabular form of both *Single stage* and *Two stage RC Coupled amplifiers*.
16. Drawn the graph for which the *frequency* on X-axis and *Gain in dB* on Y-axis for both *RC Coupled Amplifier* circuits..
17. We have calculated the bandwidth of both *RC Coupled amplifiers* from that graph as per given formula,

$$\text{Band width for Single stage RC Coupled Amplifier (BW)} = f_2 - f_1$$

$$\text{Band width for Two stage RC Coupled Amplifier (BW)} = f_4 - f_3$$
18. We have observed that, the graph which is drawn by manually is same to the graph which is obtained from the AC Analysis for both *RC coupled amplifiers*.

PROCEDURE - HARDWARE :

1. We have connected the circuit as per the circuit diagram which is shown above.
 2. Initially connected the probe across the function generator as per shown in the circuit diagram to set the input signal.
 1. Switched ON the CRO and function generator.
 2. Applied the input signal as *sine wave form* having the values of 20m_{p-p} 1KHz. from the function generator by observing in the CRO.
 3. Removed the probe from that place and connected it across the C_{C2} to observe the output of single stage.
 4. Switched ON the RPS and kept the +12V as V_{CC}.
 5. Kept the amplitude of the input signal as constant as 20mV_{p-p} for all frequency steps.
 6. Noted down the values of output voltage in terms of peak to peak voltages by varying the different frequency steps in the function generator which are given below, 20Hz, 100Hz., 200Hz., 500Hz, 1KHz, 100KHz, 200KHz, 400KHz, 600KHz, 1180KHz, 1MHz.
 7. The above readings noted in the tabular form of *single stage RC coupled amplifier*.
 8. Disconnect the probe from C_{C2} and reconnected it across C_{C4} to observe the output of second stage.
 9. Repeat the same procedure from the step 6 to 8 for tabular form of *Two stage RC Coupled Amplifier*.
 10. Now calculated and noted down the values in the tabular form of *single stage RC Coupled Amplifier* as per given below,
 - a). Voltage gain (A_v) = V_o / V_i and Gain in dB = $20\log_{10}(A_v)$.
 - b). Plotted the graph between *frequency on X- axis* and *gain in dB on Y- axis*.
 - c). Band width from the graph by using the formula- *Band width* = $f_2 - f_1$
 11. Now calculated and noted down the values in the tabular form of *Two stage RC Coupled Amplifier* as per given below,
 - a). Voltage gain (A_v) = V_o / V_i and Gain in dB = $20\log_{10}(A_v)$.
 - b). Plotted the graph between *frequency on X- axis* and *gain in dB on Y- axis*.
 - c). Band width from the graph by using the formula- *Band width* = $f_4 - f_3$

TABULAR COLUMN – FOR SINGLE STAGE RC COUPLED AMPLIFIER :

Input Voltage (V_i) = 20 mV_{P-P} (0.02V) is constant for all readings.

For Software :

For Hardware :

Sl.No.	Frequency In Hz/KHz.	Output Voltage (V_o) In mVolts.	Voltage gain $A_v = V_o/V_i$	Gain in dB = $20\log_{10}(A_v)$		Frequency In Hz/KHz.	Output Voltage (V_o) In mVolts.	Voltage gain $A_v = V_o/V_i$	Gain in dB = $20\log_{10}(A_v)$
1	20 Hz.								
2	100 Hz.								
3	200 Hz.								
5	1 KHz.								
6	200KHz.								
7	400KHz.								
8	600KHz.								
9	1180KHz.								
10	1 MHz.								
11	100 MHz					-----	-----	-----	-----
12	500MHz.					-----	-----	-----	-----

TABULAR COLUMN - TWO STAGE RC COUPLED AMPLIFIER :

Input Voltage (V_i) = 20 mV_{P-P} (0.02V) is constant for all readings.

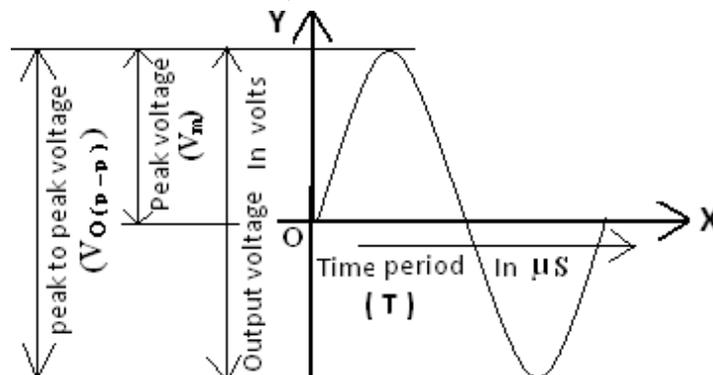
For Software :

For Hardware :

Sl.No.	Frequency In Hz/KHz.	Output Voltage (V_o) In mVolts.	Voltage gain $A_v = V_o/V_i$	Gain in dB = $20\log_{10}(A_v)$		Frequency In Hz/KHz.	Output Voltage (V_o) In mVolts.	Voltage gain $A_v = V_o/V_i$	Gain in dB = $20\log_{10}(A_v)$
1	20 Hz.								
2	100 Hz.								
3	200 Hz.								
5	1 KHz.								
6	200KHz.								
7	400KHz.								
8	600KHz.								
9	1180KHz.								
10	1 MHz.								
11	100 MHz					-----	-----	-----	-----
12	500MHz.					-----	-----	-----	-----

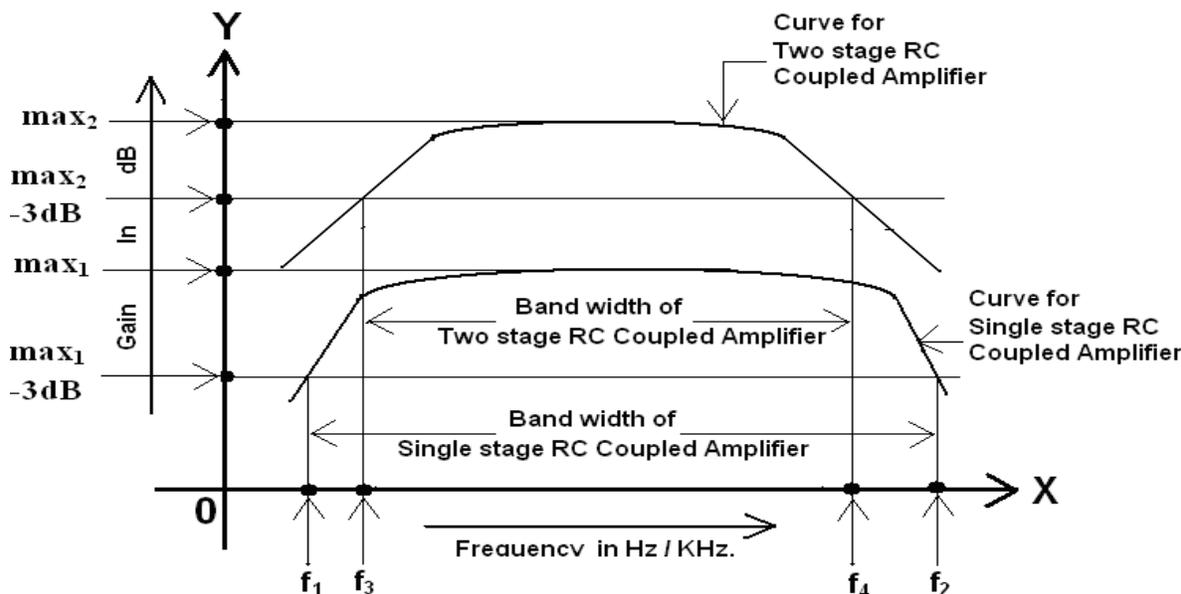
EXPECTED WAVEFORM – SOFTWARE & HARDWARE :

I have got the *Sine wave form* on the CRO as output signal for single stage as well as for Two stage *RCCoupled Amplifiers* which is shown below,



EXPECTED GRAPH – SOFTWARE & HARDWARE :

The following graph shows the frequency response curves of both *Single stage & Two stage RC coupled Amplifiers*.



CALCULATIONS – SOFTWARE & HARDWARE :

- 1). Band width “single stage RC coupled amplifier = $f_2 - f_1 =$
- 2). Band width “two stage RC coupled amplifier = $f_4 - f_3 =$

CONCLUSION – SOFTWARE & HARDWARE :

1. I have observed that
 - a). The bandwidth of *Two stage RC coupled amplifier* is less as compared to *Single stage RC coupled amplifier* and
 - b). The gain of *Two stage RC coupled amplifier* is more as compared to *Single stage RC coupled amplifier*

RESULT – SOFTWARE & HARDWARE :

I verified / drawn the frequency response curve and found the bandwidth values of a *single stage & two stage RC coupled amplifiers*. The band width values are,

- 1). Band width of *single stage RC coupled amplifier* = _____
- 2). Band width of *two stage RC coupled amplifier* = _____

VIVA VOCE QUESTIONS :

1. Applications of Darlington pair Amplifier.
 2. Applications of Multi stage amplifiers?
 3. Mention Advantages of Multistage Amplifiers.
 4. What is Band Width?
 5. What is Frequency Response?
 6. Need for multi stage amplifier?
 7. What are the different coupling schemes?
 8. Applications of Multi stage amplifiers?
 9. Mention Advantages of Multistage Amplifiers.
 10. What is Band Width?
-

Experiment No. : 04

Date :

Name of the Experiment : DARLINGTON PAIR AMPLIFIER

AIM :

To obtain the frequency response curve of *Darlington pair amplifier* using software & hardware

APPARATUS :**Software :**

1. System
2. Multisim software

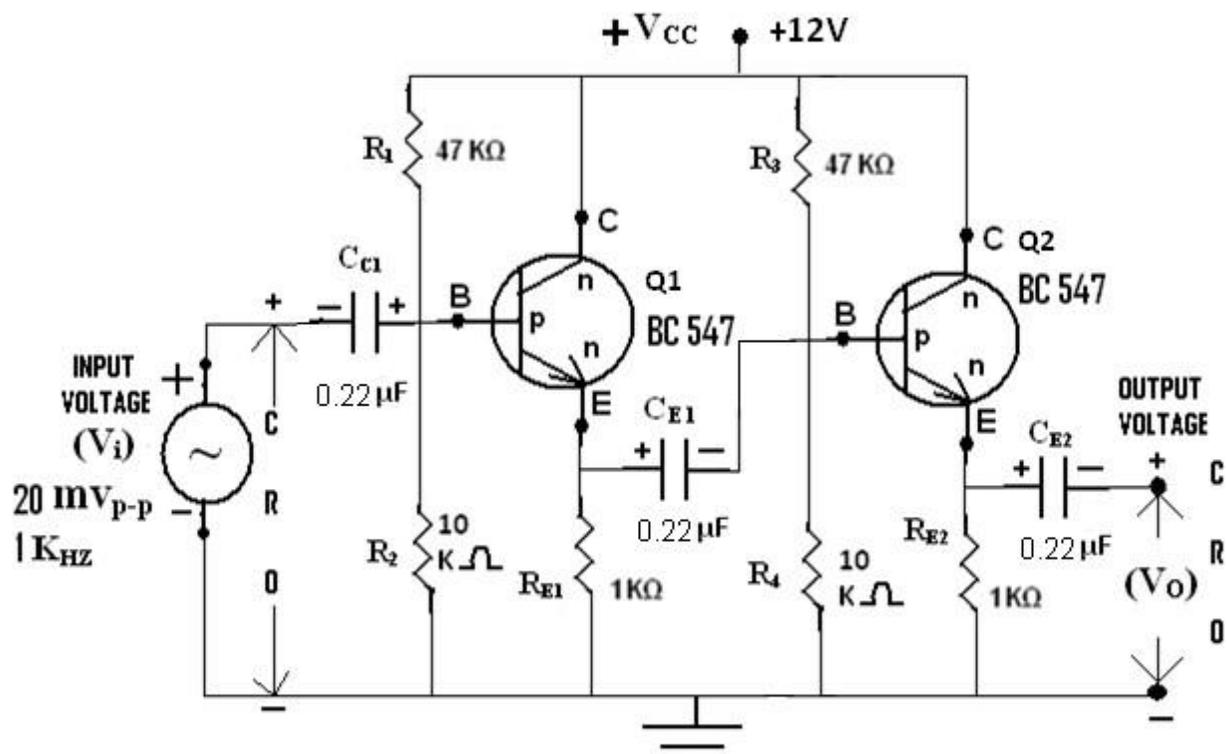
Hardware :

- 1). Transistor a). BC547 NPN ----- 2 No.
- 2). Resistors a). 47K Ω ----- 2 No.
b). 10 K Ω ----- 2 No.
d). 1 K Ω ----- 2 No.
- 3). Capacitors a). 0.22 μ F ----- 3 No.

THEORY :

Darlington Pair amplifier circuit is a connection of two transistors which acts as a single unit with overall current gain equal to the multiplication of the individual current gains of the transistors. Darlington pair **transistor amplifier circuit** is very popular in electronics. Clearly, it is an **Amplifier circuit**. In this article, we are going to discuss the theory and the applications of Darlington pair amplifier.

The current gain of Darlington pair amplifier is almost equal to the product between the current gains of individual transistors. If β_1 and β_2 be the current gains of individual transistors then overall current gain of Darlington pair amplifier = $\beta_1\beta_2$.

CIRCUIT DIAGRAM – SOFTWARE & HARDWARE :

PROCEDURE – SOFTWARE :

1. We have picked up the components from the components bar as per above circuit.
2. Made the connections as per the above circuit diagram by using the components which we have picked up.
3. Connected the CRO across the capacitor C_{E2} of second stage.
4. Set the input signal as *sine wave form which is having the value* $20\text{mV}_{\text{P-P}}$ as constant in the function generator.
5. Initially set the input signal frequency value is 1KHz in the function generator.
6. To simulate the circuit clicked on *run option* through *execute button* in *tool bar*.
7. We have seen the *sine wave* on the **CRO** screen as o/p signal.
8. Calculated the *peak to peak voltage* ($V_{O(p-p)}$) and noted down in the tabular form Against the 3 column of 1KHz.
9. Stopped the simulation by clicked on *run option* through *execute button* in the *tool bar*.
10. Repeated the same procedure from points 6 to 9 for the corresponding frequency values by setting in the function generator for the following steps,
20Hz, 100Hz., 200Hz., 500Hz, 1KHz, 200KHz, 400KHz, 600KHz, 1180KHz, 1MHz, 100MHz, 500MHz.
in the function generator.
11. Finally shut down the system safely.
12. Now calculated and noted down the values of *voltage gain* (A_v) and *gain in dB* to the corresponding values of *output voltage* (V_o) & *input voltage* (V_i) by using the formulas given below,
 $Voltage\ gain\ (A_v) = V_o / V_i$ and $Gain\ in\ dB = 20\log_{10}(A_v)$.

PROCEDURE – HARDWARE :

1. We have connected the circuit as per the circuit diagram which is shown above.
2. Initially connected the probe across the function generator as per shown in the circuit diagram to set the input signal.
3. Switched *ON* the *CRO* and *function generator*.
4. Applied the input signal as *sine wave form* of $20\text{mV}_{\text{P-P}}$, 1KHz. from the function generator by observing in the CRO.
5. Later removed the probe from that place and connected it across the capacitor C_{E3} to observe the output.
6. Switched *ON* the *RPS* and kept the 12V as V_{CC} .
7. Kept the amplitude of the input signal as constant as $20\text{mV}_{\text{P-P}}$ for all frequency steps.
8. Noted down the values output voltage of output signal in terms of peak to peak voltages by varying the different frequency steps in the function generator which are given below,
20Hz, 100Hz., 200Hz., 500Hz, 1KHz, 100KHz, 200KHz, 400KHz, 600KHz, 1180KHz, 1MHz.
9. Repeated the same procedure for point 8 for corresponding frequency values.
10. Now calculated and noted down the values of *voltage gain* (A_v) and *gain in dB* to the corresponding values of *output voltage* (V_o) & *input voltage* (V_i) by using the formulas given below,
 $Voltage\ gain\ (A_v) = V_o / V_i$ and $Gain\ in\ dB = 20\log_{10}(A_v)$.

TABULAR COLUMN :

Input Voltage (V_i) = $20\text{mV}_{\text{P-P}}$ (0.02V) is constant for all readings.

For Software :

For Hardware :

Sl.No.	Frequency In Hz/KHz.	Output Voltage (V _o) In mVolts.	Voltage gain A _v = V _o /V _i	Gain in dB = 20log ₁₀ (A _v)		Frequency In Hz/KHz.	Output Voltage (V _o) In mVolts.	Voltage gain A _v = V _o /V _i	Gain in dB = 20log ₁₀ (A _v)
1	20 Hz.								
2	100 Hz.								
3	200 Hz.								
4	1 KHz.								
5	200KHz.								
6	400KHz.								
7	600KHz.								
8	1180KHz.								
9	1 MHz.								
10	100 MHz					-----	-----	-----	-----
11	500MHz.					-----	-----	-----	-----

EXPECTED GRAPH – SOFTWARE & HARDWARE :

Note : We can't draw the graph and could not find the band width for this experiment, because there is no amplification.

CONCLUSION :

We have formed the circuit of Darlington pair amplifier by connected two common collector amplifiers in two stages. The input impedance of two stage common collector amplifier i.e. Darlington pair amplifier is very high as compared to single stage common collector amplifier. Due to this reason only the voltage gain of Darlington pair amplifier is less than as compared to single stage common collector amplifier.

RESULT – SOFTWARE & HARDWARE :

I have obtained the voltage gain and gain in db at different frequencies of a *Darlington pair amplifier*.

VIVA VOICE QUESTIONS:

1. Applications of Darlington pair Amplifier.
2. Applications of Multi stage amplifiers?
3. Mention Advantages of Multistage Amplifiers.
4. What is Band Width?
5. What is Frequency Response?
6. Compare CB,CE, CC configurations of a transistor
7. Explain the transistor operation with the help of four regions
8. What is cascade Amplifier?
9. Explain base width modulation of a transistor
10. Which Amplifier is having CC-CC configuration?

Experiment No. : 05

Date :

Name of the Experiment : **CE – CB - CASCODE AMPLIFIER****AIM :**

- 1). To obtain the frequency response of *CE – CB cascode amplifier* using Software and Hardware
- 2). To calculate the band width of this amplifier.

APPARATUS :**Software :**

1. System..... 1 No.
2. Multisim software

Hardware :

- 1). Function generator(**FG**) ----- 1 No.
- 2). Cathode Ray Oscilloscope(**CRO**)----- 1 No.
- 3). Regulated Power Supply (**RPS**) : (0-30)V, 1A Dual channel ----- 1 No.
- 4). Probes----- 1 No.
- 5). Bread board ----- 1 No.
- 6). Connecting wires :-----A few Nos.

COMPONENTS :

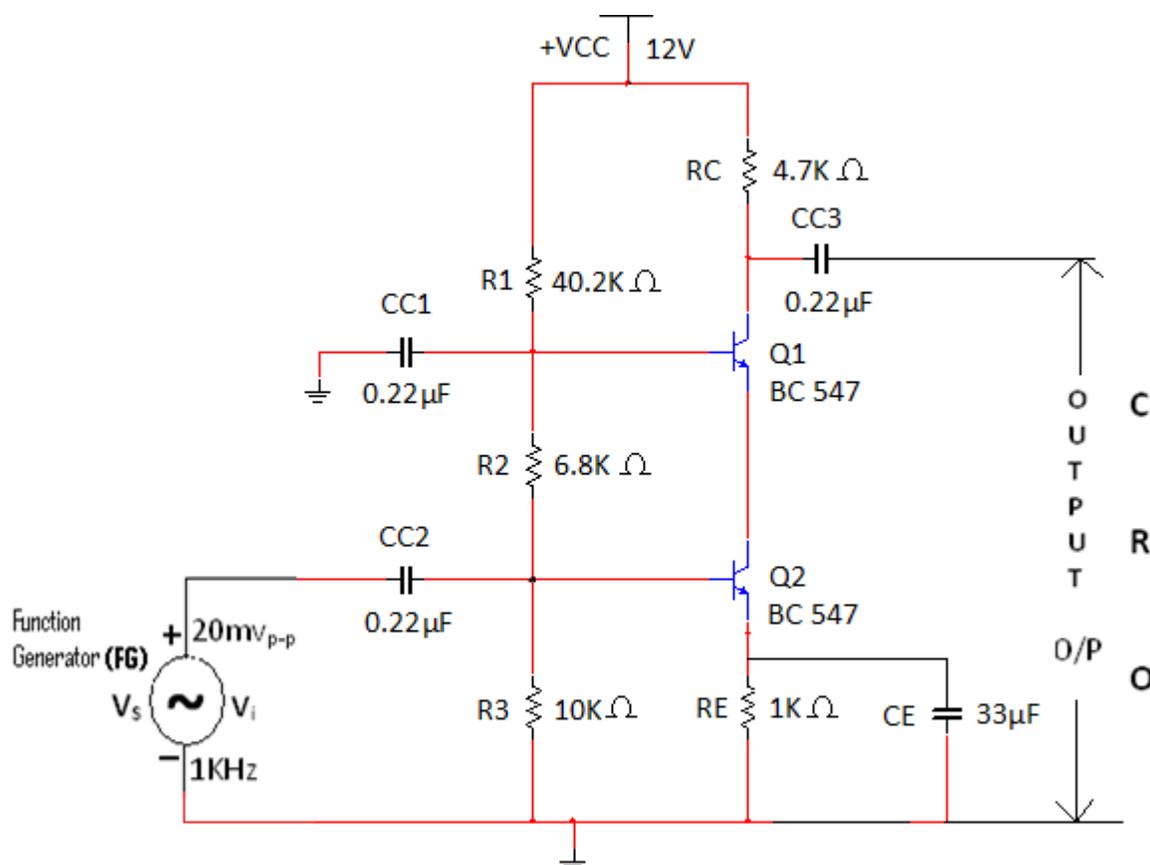
- 1). Transistor BC 547 ----- 1 No.
- 2) Carbon fixed resistors
 - a). 47K Ω , 1/2W ----- 1 No.
 - b). 40.2K Ω ----- 1 No.
 - c). 10K Ω , 1/2W ----- 1 No.
 - d). 6.8 K Ω , 1/2W ----- 1 No.
 - e). 4.7 K Ω , 1/2W ----- 1 No.
 - f). 1 K Ω , 1/2W ----- 1 No.
- 3). Capacitors
 - g). 0.22 μ F ----- 3 No.
 - h). 33 μ F ----- 1 No.

THEORY :

While the C-B (common-base) amplifier is known for wider bandwidth than the C-E (common-emitter) configuration, the low input impedance (10s of Ω) of C-B is a limitation for many applications. The solution is to precede the C-B stage by a low gain C-E stage which has moderately high input impedance (k Ω s).

The stages are in a cascode configuration stacked in series, as opposed to cascaded for a standard amplifier chain.

The key to understanding the wide bandwidth of the cascode configuration is the Miller effect. **The Miller effect is the multiplication of the bandwidth robbing collector-base capacitance by voltage gain A_v .**

CIRCUIT DIAGRAM – SOFTWARE & HARDWARE :**PROCEDURE – SOFTWARE :**

1. We have picked up the components from the components bar as per above circuit.
2. Made the connections as per the above circuit diagram by using the components which we have picked up.
3. Set the input signal as *sine wave form* which is having the value $20\text{mV}_{\text{P-P}}$ as constant in the function generator.
4. Initially set the input signal frequency value is 1KHz in the function generator.
5. To simulate the circuit clicked on *run option* through *execute button* in *tool bar*.
6. We have seen the *sine wave* on the **CRO** screen as o/p signal.
7. Calculated the *peak to peak voltage* ($V_{O(p-p)}$) and noted down in the tabular form Against the column of 1KHz .
8. Stopped the simulation by clicked on *run option* through *execute button* in the *tool bar*.
9. Repeat the same procedure from points 7 to 9 for the corresponding frequency values by setting in the function generator for the following steps, 20Hz , 100Hz , 200Hz , 1KHz , 200KHz , 400KHz , 600KHz , 1180KHz , 1MHz , 100MHz , 500MHz . in the function generator.
10. Observed the graph for *frequency Vs amplitude* through the *AC Analysis*.
11. Finally shut down the system safely.
12. We have observed that, the graph which is drawn by manually is same to the graph which is obtained from the *AC Analysis*.

13. Now calculated and noted down the values of *voltage gain* (A_V) and *gain in dB* to the corresponding values of *output voltage* (V_O) & *input voltage* (V_i) by using the formulas given below,
Voltage gain (A_V) = V_o / V_i and *Gain in dB* = $20\log_{10}(A_V)$.
- 14). Plotted the graphs (frequency response curves) as per below
- frequency on X-axis & gain in dB on Y-axis.
 - frequency on X-axis & voltage gain on Y-axis.

PROCEDURE – HARDWARE :

- Connected the circuit as per the circuit diagram.
- Then switched ON the *function generator* and *CRO*; but don't switched ON the *RPS*. Now Kept the *AC/GND/DC* switch is at *AC* position.
- Initially kept the 1KHz. frequency by varying the frequency control in the *function generator*.
- Now applied the peak to peak amplitude of a sine wave is of $20\text{mV}_{\text{p-p}}$ by varying the amplitude control In the *function generator* through observing in the *CRO*.
- Kept this input value as $20\text{mV}_{\text{p-p}}$ constant up to the completion of the experiment Otherwise the wrong output would occurred.
- Now switched ON the *RPS* and set the 10V in it i.e. $V_{\text{CC}} = 12\text{V}$.
- Varied the different frequency steps of 20Hz, 100Hz, 200Hz, 1KHz, 200KHz, 400KHz, 600KHz, 800KHz, 1MHz. by adjusted the frequency control in the *function generator* and noted down the corresponding values of output signal i.e. peak to peak amplitude of sine wave by observing in the *CRO*.
- Now switched OFF the *RPS*, *function generator* and *CRO*.
- Then calculated the *voltage gain* $A_V = V_o/V_i$ & *gain in dB* = $20\log_{10}(A_V)$ and noted down the values in the specified columns of the tabular column.
- Plotted the graphs (frequency response curves) as per below,
 - frequency on X-axis & gain in dB on Y-axis.
 - frequency on X-axis & voltage gain on Y-axis.
- Calculated the *band width* from the above two (frequency response curves) graphs by using the formula $f_2 - f_1$ which is given under the heading of *parameters*.

TABULAR COLUMNS :

Input Voltage (V_i) = $20\text{mV}_{\text{p-p}}$ (0.02V) is constant for all readings.

For Software :

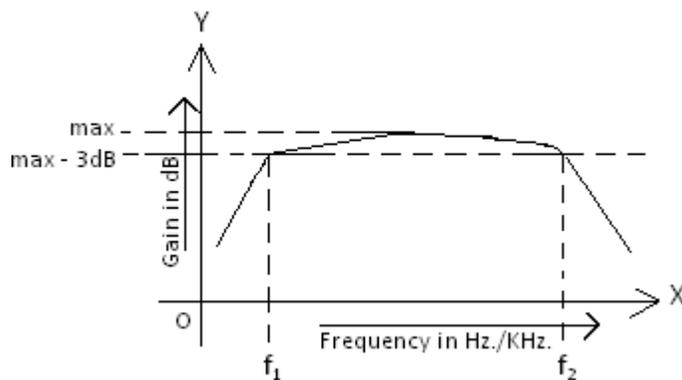
For Hardware :

Sl. No.	Frequency In Hz/KHz.	Output Voltage (V_o) In mVolts.	Voltage gain $A_V = V_o/V_i$	Gain in dB = $20\log_{10}(A_V)$	Frequency In Hz/KHz.	Output Voltage (V_o) In mVolts.	Voltage gain $A_V = V_o/V_i$	Gain in dB = $20\log_{10}(A_V)$
1	20 Hz.							
2	100 Hz.							
3	200 Hz.							
----- To be continued in next page -----								

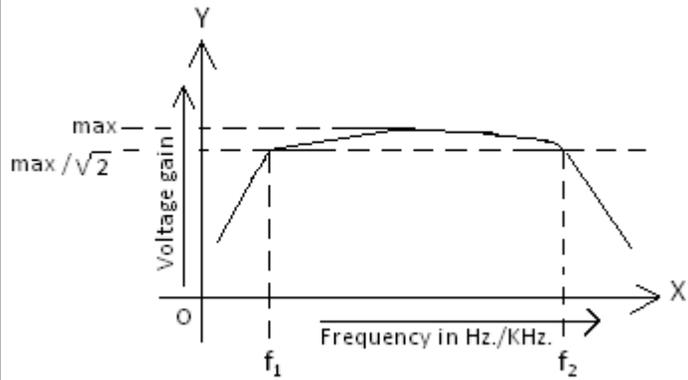
Sl. No.	Frequency In Hz/KHz.	Output Voltage (V _O) In mVolts.	Voltage gain A _v = V _o /V _i	Gain in dB = 20log ₁₀ (A _v)	Frequency In Hz/KHz.	Output Voltage (V _O)In mVolts.	Voltage gain A _v = V _o /V _i	Gain in dB = 20log ₁₀ (A _v)
4	1 KHz.							
5	200KHz.							
6	400KHz.							
7	600KHz.							
8	1180KHz.							
9	1 MHz.							
10	100 MHz				-----	-----	-----	-----
11	500MHz.				-----	-----	-----	-----

EXPECTED GRAPHS – SOFTWARE & HARDWARE :

A). Frequency response curve
For frequency verses gain in dB.



B). Frequency response curve
For frequency verses voltage gain.



PARAMETERS – SOFTWARE & HARDWARE :

- 1). Band width of frequency response curve for frequency verses gain in dB. = $f_2 - f_1 =$
- 2) Band width of frequency response curve for frequency verses voltage gain = $f_2 - f_1 =$

RESULT –SOFTWARE & HARDWARE :

We have obtained the frequency response curves of *CE-CB cascode Amplifier* for frequency verses gain in dB & frequency verses voltage gain and calculated the band width of both of them. The band width values are given below,

- 1). Band width of frequency response curve for frequency verses gain in dB. =
- 2) Band width of frequency response curve for frequency verses voltage gain =

1. What is cascode Amplifier?
2. CE-CB configuration is having which Amplifier?
3. Applications of Multi stage amplifiers?
4. Mention Advantages of Multistage Amplifiers
5. What is Band Width?
6. What is Frequency Response?
7. What is cascade Amplifier?
8. Explain the transistor operation with the help of four regions
9. Explain base width modulation of a transistor
10. Compare CB,CE, CC configurations of a transistor.

Experiment No. : 6**Date :****Name of the Experiment : VOLTAGE SERIES FEEDBACK AMPLIFIER****AIM :**

To design and obtain the frequency response of *Voltage series feedback* using software and hardware.

APPARATUS :**Software :**

1. System 1 No.
2. Multisim software

Hardware :

- 1). Function generator(*FG*) ----- 1 No.
- 2). Cathode Ray Oscilloscope(*CRO*) ----- 1 No.
- 3). Regulated Power Supply (*RPS*) : (0-30)V, 1A Dual channel ----- 1 No.
- 4). Probes ----- 1 No.
- 5). Bread board ----- 1 No.
- 6). Connecting wires : ----- A few Nos.
 - 1). Transistor a). BC547 NPN ----- 1 No.
 - 2). Resistors a). 47K Ω ----- 1 No.
 - b). 10 K Ω ----- 1 No.
 - d). 1 K Ω ----- 1 No.
 - 3). Capacitors a). 0.22 μ F ----- 2 No.

THEORY :**Feedback :**

Feedback is said to exist in an amplifier circuit, when a fraction of the output signal is returned or fed back to the input and combined with the input signal. If the magnitude of the input signal is reduced by the feed back, the feed back is called negative or degenerative. If the magnitude of the input signal is increased by the feed back, such feed back is called positive or regenerative.

Definition :

When any increase in the output signal results into the input in such a way as to cause the decrease in the output signal, the amplifier is said to have negative feedback. ... In Voltage-Series feedback, the input impedance of the amplifier is increased and the output impedance is decreased.

True voltage :

Voltage series feedback amplifier have the difference voltage, $V_{id} = V_{in} - V_f$. Therefore, the feedback voltage always opposes the input voltage and is out of phase by 180° with respect to input voltage. Hence, the feedback is said to be negative.

Benefit of high input impedance :

1. It provides good amplification to the input signal otherwise we get low voltage and that leads to low amplification
2. It minimizing the loading effect on input and thus significant amount of input voltage signal is maintained for amplification.

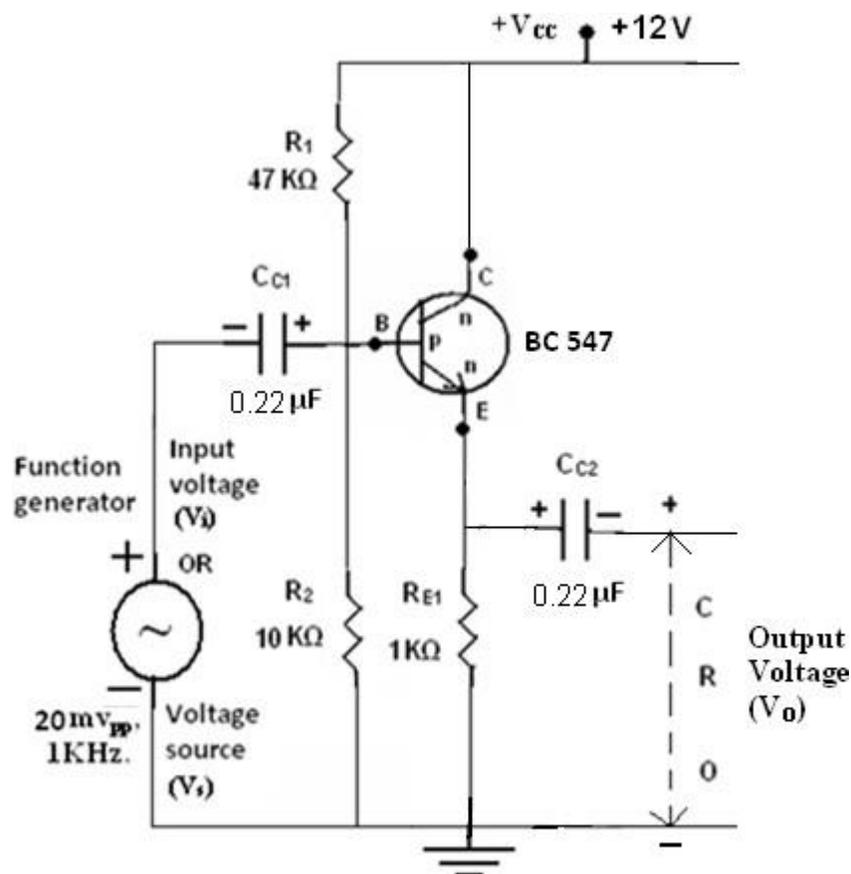
CIRCUIT DIAGRAM – SOFTWARE & HARDWARE :

Figure : Circuit diagram of Voltage series feed back amplifier

PROCEDURE – SOFTWARE :

1. We have picked up the components from the components bar as per above circuit.
2. Made the connections as per the above circuit diagram by using the components which we have picked up.
3. Connected the CRO across the *Emitter capacitor to ground*.
4. Set the input signal as *sine wave form* 20mV_{P-P} and 1KHz. as constant in the function generator.
5. To simulate the circuit clicked on *run option* through *execute button* in *tool bar*.
6. We have seen the *sine wave* on the **CRO** screen as o/p signal.
7. Calculated the *peak to peak voltage* ($V_{O(p-p)}$) and noted down in the tabular form against the column of 1KHz.
8. Stopped the simulation by clicking on *run option* through *execute button* in the *tool bar*.
9. Repeated the same procedure from points 5 to 8 for the corresponding frequency values by setting in the function generator for the following steps,
20Hz, 100Hz., 200Hz., 500Hz, 1KHz, 200KHz, 400KHz, 600KHz, 1180KHz, 1MHz, 100MHz, 500MHz. in
the function generator.
10. Observed the graph for *frequency Vs amplitude* through the *AC Analysis*.
11. Finally shut down the system safely.
12. We have observed that, the graph which is drawn by manually is same to the graph which is obtained from the *AC Analysis*.

13. Now calculated and noted down the values of *voltage gain* (A_V) and *gain in dB* to the corresponding values of *output voltage* (V_O) & *input voltage* (V_i) by using the formulas given below,

$$\text{Voltage gain } (A_V) = V_O / V_i \quad \text{and} \quad \text{Gain in dB} = 20 \log_{10}(A_V).$$

PROCEDURE – HARDWARE :

- 1). Connected the circuit as per the circuit diagram.
- 2). Removed the probe of *CRO* from output (O/P) side and connected it at input (I/P) side to set the input signal
i.e. sine wave having the value of $20\text{mV}_{\text{p-p}}$ & 1KHz.
- 3). Then switched ON the *function generator* and *CRO*; but don't switched ON the *RPS*.
- 4). Now Kept the *AC/GND/DC* switch is at *AC* position.
- 5). Now applied the input signal i.e. sine wave by pressing the sine wave function key in the *function generator*.
- 6). Initially kept the peak to peak amplitude of a sine wave is of $20\text{mV}_{\text{p-p}}$, 1KHz. frequency by varying the amplitude and frequency controls in the *function generator* through observing in the *CRO*.
- 7). Kept this value of input signal as constant up to the completion of the experiment Otherwise the wrong output would occurred.
- 8). Then removed the probe of *CRO* from the input side and connected it across the output side.
- 9). Now switched ON the *RPS* and set the 10V in it i.e. $V_{CC} = 12\text{V}$.
- 10). Varied the different frequency steps of 20Hz, 100Hz., 200Hz., 500Hz, 1KHz, 100KHz, 200KHz, 400KHz, 600KHz, 1180KHz, 1MHz. by adjusted the frequency control in the *function generator* and noted down the corresponding values of output signal i.e. peak to peak amplitude (voltage) of sine wave by observing in the *CRO*.
- 11). Now switched OFF the *RPS*, *function generator* and *CRO*.
- 12). Then calculated the *voltage gain* $A_V = V_O / V_i$ & *gain in dB* $= 20 \log_{10}(A_V)$ and noted down the values in the specified columns of the tabular column.

Notes:

1. Amplifier means which amplifies the sinusoidal and non-sinusoidal wave forms with out change in frequency. In voltage series feedback amplifier, network is in parallel with the the output of the amplifier.
2. A fraction of the output voltage through the feedback network is applied in series with in the input voltage of the amplifier.
3. The series connections at the input, increase the input resistance. In this case the amplifier is a true voltage amplifier.
4. The common collector or emitter follower is an example of voltage series feedback amplifier. Since the voltage developed in the output is in series with the input voltage as for as the base – emitter junction is connected.

TABULAR COLUMN :

Input Voltage (V_i) = 20 mV_{P-P} (0.02V) is constant for all readings.

For Software :

For Hardware :

Sl.No.	Frequency In Hz/KHz.	Output Voltage (V_o) In mVolts.	Voltage gain $A_v = V_o/V_i$	Gain in dB = $20\log_{10}(A_v)$		Frequency In Hz/KHz.	Output Voltage (V_o) In mVolts.	Voltage gain $A_v = V_o/V_i$	Gain in dB = $20\log_{10}(A_v)$
1	20 Hz.								
2	100 Hz.								
3	200 Hz.								
4	1 KHz.								
5	200KHz.								
6	400KHz.								
7	600KHz.								
8	1180KHz.								
9	1 MHz.								
10	100 MHz					-----	-----	-----	-----
11	500MHz.					-----	-----	-----	-----

EXPECTED GRAPH – SOFTWARE & HARDWARE :

Note : We could not draw frequency response curve and not to be calculate band width for this experiment, because there is no any amplification. It is just working as *buffer*.

RESULT – SOFTWARE & HARDWARE :

We have obtained the Voltage gain & Gain in db of a given amplifier.

VIVA VOICE QUESTIONS:

1. What is feedback?

2. What are the advantages of negative feedback?

3. What are the feedback topologies?

4. Example for voltage series feedback amplifier.

5. What are the CC Amplifier characteristics?

6. What are the Applications of Multi stage amplifiers?

7. Example for voltage series feedback amplifier.

8. CC Amplifier characteristics?

9. What is Band Width?

10. Explain the transistor operation with the help of four regions

Experiment No. : 7 **Date :**
Name of the Experiment : CURRENT SHUNT FEEDBACK AMPLIFIER

AIM :

- 1). To plot the graph for frequency response curve of a *Current shunt feedback Amplifier with feedback and without feedback* using software and hardware
- 2). To find the bandwidth of *Current shunt feedback Amplifier*.

APPARATUS :

Software :

1. System 1 No.
2. Multisim software

Hardware :

- 1). Function generator(*FG*) ----- 1 No.
- 2). Cathode Ray Oscilloscope(*CRO*) 3). (0-30)V, 1A Dual channel ----- 1 No.
- Regulated Power Supply (*RPS*) ----- 1 No.
- :4). Probes
- 5). Bread board ----- 1 No.
- 6). Connecting wires : ----- A few Nos.

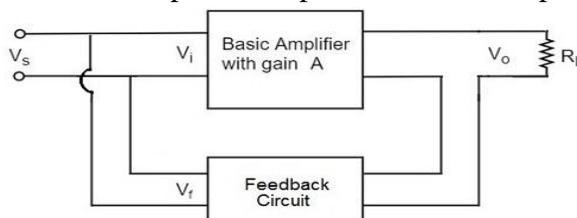
COMPONENTS :

- 1). Transistor BC 547 ----- 2 No.
- 2) Carbon fixed resistors a). 47KΩ, 10KΩ , 4.7 KΩ , 1 KΩ ----- Each 2 No.
 b). 100 Ω , ½W ----- 1 No.
- 3). Capacitors a). 0.22μF ----- 4 No.
 b). 10μF ----- 1 No.
 c). 33μF ----- 1 No.

THEORY :

In the current shunt feedback circuit, a fraction of the output voltage is applied in series with the input voltage through the feedback circuit. This is also known as **series-driven shunt-fed** feedback i.e., a series-parallel circuit.

The below figure shows the block diagram of current shunt feedback, by which it is evident that the feedback circuit is placed in series with the output but in parallel with the input.



As the feedback circuit is connected in series with the output, the output impedance is increased and due to the parallel connection with the input, the input impedance is decreased.

Let us now tabulate the amplifier characteristics that get affected by different types of negative feedbacks.

As the feedback circuit is connected in series with the output, the output impedance is increased and due to the parallel connection with the input, the input impedance is decreased.

CIRCUIT DIAGRAM :

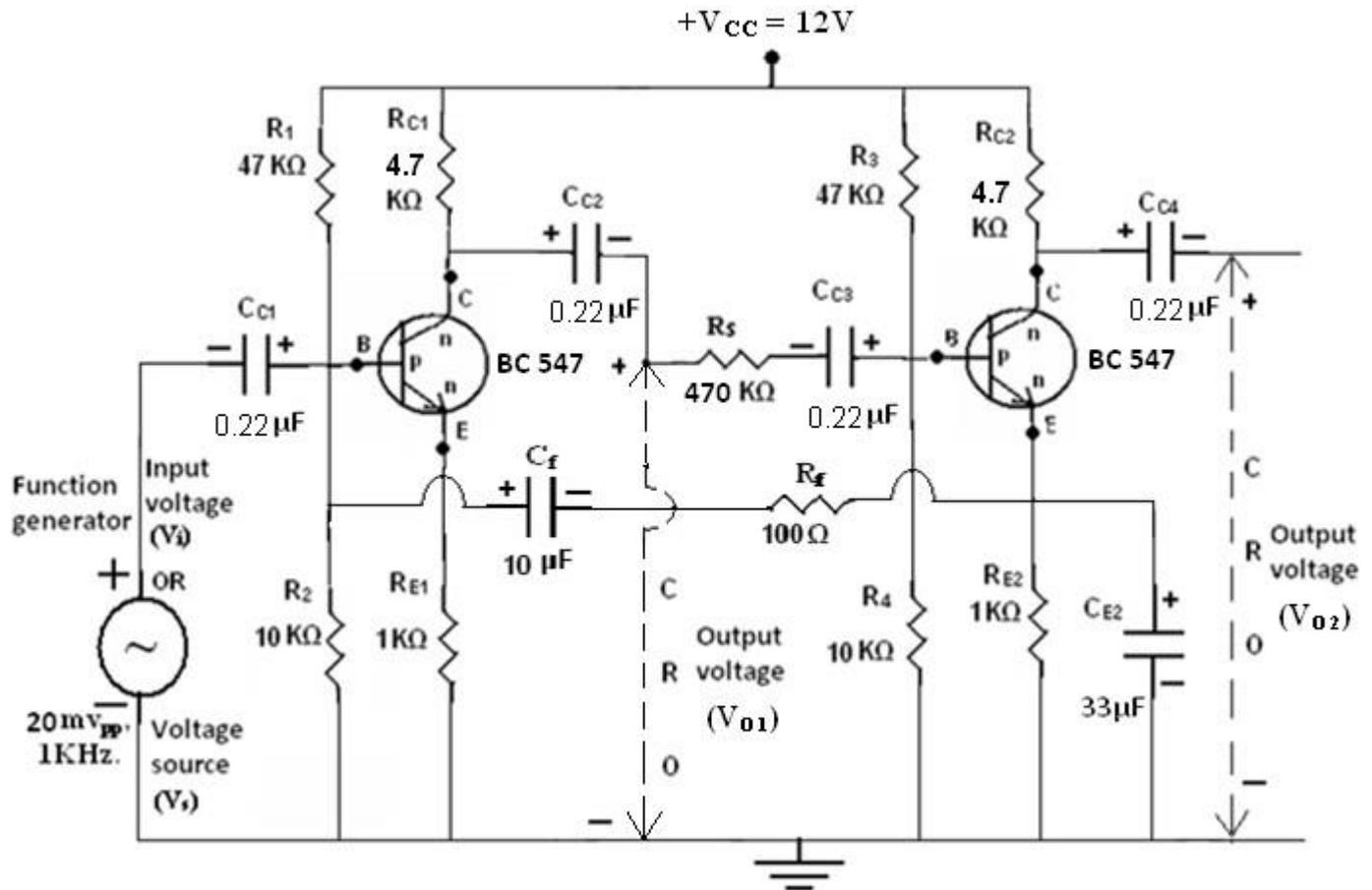


Figure : Circuit diagram Current shunt feed back amplifier

TABULAR COLUMN – WITH FEED BACK :

Input Voltage (V_i) = 20 mV_{P-P} (0.02V) is constant for all readings. For Software :For Hardware :

Sl.No.	Frequency In Hz/KHz.	Output Voltage (V_o) In mVolts.	Voltage gain $A_v = V_o/V_i$	Gain in dB = $20\log_{10}(A_v)$	Frequency In Hz/KHz.	Output Voltage (V_o) In mVolts.	Voltage gain $A_v = V_o/V_i$	Gain in dB = $20\log_{10}(A_v)$
1	20 Hz.							
2	100 Hz.							
3	200 Hz.							
5	1 KHz.							
6	200KHz.							
7	400KHz.							
8	600KHz.							
9	800KHz.							
10	1 MHz.							
11	100 MHz				-----	-----	-----	-----
12	500MHz.				-----	-----	-----	-----

TABULAR COLUMN – WITHOUT FEED BACK :

Input Voltage (V_i) = 20 mV_{P-P} (0.02V) is constant for all readings.

For Software :

For Hardware :

Sl.No.	Frequ- ency In Hz/KHz.	Output Voltage (V_o) In mVolts.	Voltage gain $A_v =$ V_o/V_i	Gain in dB = $20\log_{10}$ (A_v)		Frequ- ency In Hz/KHz.	Output Voltage (V_o) In mVolts.	Voltage gain $A_v =$ V_o/V_i	Gain in dB = $20\log_{10}$ (A_v)
1	20 Hz.								
2	100 Hz.								
3	200 Hz.								
4	1 KHz.								
5	200KHz.								
6	400KHz.								
7	600KHz.								
8	1180KHz.								
9	1 MHz.								
10	100 MHz					-----	-----	-----	-----
11	500MHz.					-----	-----	-----	-----

PROCEDURE – SOFTWARE :

1. We have picked up the components from the components bar as per above circuit.
2. Made the connections as per the above circuit diagram by using the components which we have picked up.
3. Connected the CRO across the capacitor C_4 .
4. Set the input signal as *sine wave form which is having the value 20mV_{P-P}* as constant in the function generator.
5. Initially set the input signal frequency value is 1KHz in the function generator.
6. To simulate the circuit clicked on *run option* through *execute button* in *tool bar*.
7. We have seen the *sine wave* on the **CRO** screen as o/p signal.
8. Calculated the *peak to peak voltage ($V_{o(p-p)}$)* and noted down in the column of 1 KHz. in tabular form of *with feedback* amplifier.
9. Stopped the simulation by clicked on *run option* through *execute button* in the *tool bar*.
10. Repeat the same procedure from points 6 to 9 for the corresponding frequency values by setting in the function generator for the following steps, 20Hz, 100Hz., 200Hz., 1KHz, 200KHz, 400KHz, 600KHz, 1180KHz, 1MHz, 100MHz, 500MHz in the function generator.
11. Observed the graph for *frequency Vs amplitude* through the *AC Analysis*.
12. Disconnected the C_f and R_f from the circuit and now the circuit has become as *without feedback amplifier*

13. Now taken the reading in the tabular form of *without feedback amplifier* by repeat the steps from 6 to 11
14. Finally shut down the system safely.
15. We have observed that, the graph which is drawn by manually is same to the graph which is obtained from the *AC Analysis*.
16. We have observed that the readings of *without feed back amplifier's* output voltage is greater than the *with feed back amplifier*
17. Calculated the Voltage gain by using the formula of V_o / V_i and Gain in dB by using the formula of $20\log_{10}(A_v)$ in both tabular forms of *with feed back* and *without feed back amplifiers*.
18. Drawn the graphs of both amplifiers in single graph sheet.
19. While drawing the graph taken the *frequency* on X-axis and *Gain in dB* on Y-axis.
20. Finally calculated the *bandwidth* of both amplifiers from this graph sheet as per the following formulas, i). For *Current shunt feed back amplifier (With feed back)* $(BW) = f_2 - f_1$
ii). For *Current shunt feed back amplifier (Without feed back)* $(BW) = f_4 - f_3$
21. We have noted down that the *band width* of *with feed back amplifier* is high compared to the *without feedback amplifier*.

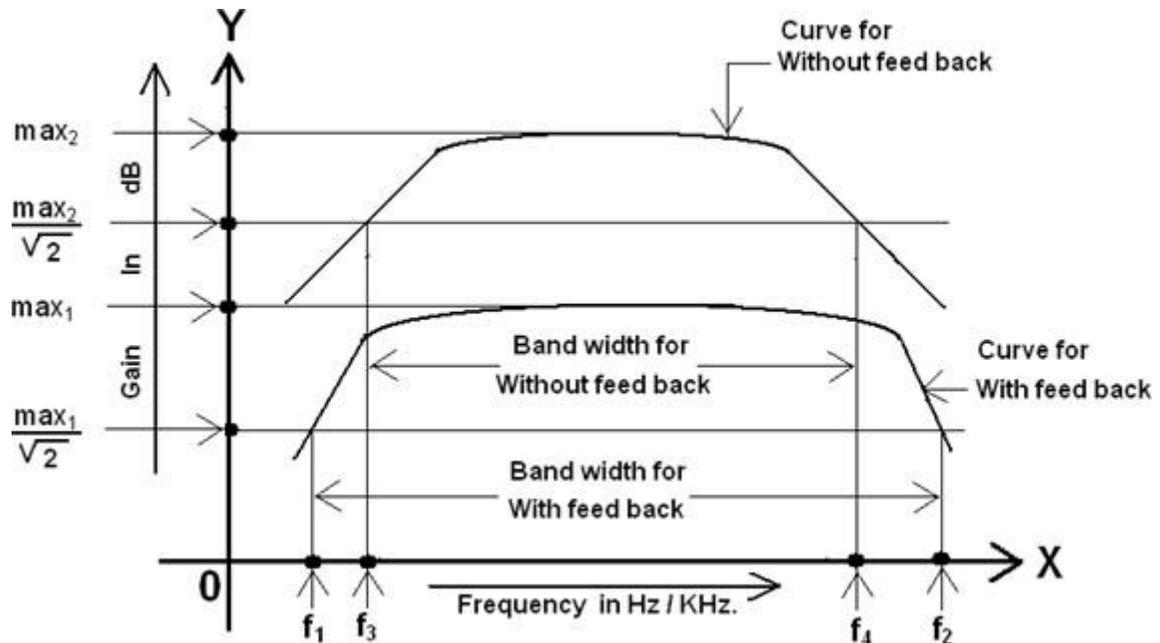
PROCEDURE – HARDWARE :

1. Connections are made as per the circuit diagram.
2. Initially connected the *CRO* across the *Function generator*.
3. Switched **ON** the Cathode ray oscilloscope (CRO) and Function generator.
4. Applied the 20 mV_{pp} , 1KHz sine wave signal to the circuit from *Function generator* by observing in the *CRO*.
5. We have kept this 20 mV_{pp} input voltage as constant for all steps of frequency while taking the readings for *Current shunt feed back amplifier* with feed back & without feed back .
6. Disconnected the CRO from the function generator, and connected it across C_{c4} to measure the peak to peak output voltage.
7. Now Connected the CRO at output side
8. Applied the $+V_{CC}$ as 10V to the circuit from the *Regulated power supply (RPS)*.
9. Later we have noted down the readings for output voltage in the tabular form of *with feed back*. from the CRO, by varying the different steps of frequency (i.e. 20Hz , 100Hz ., 200Hz ., 500Hz ., 1KHz , 200KHz , 400KHz , 600KHz , 1180KHz , 1MHz .) in function generator.
10. After this we removed the feed back capacitor (C_f) & resistor (R_f) from the circuit completely then the circuit is became as the *without feed back amplifier*.
11. Again we have noted down the readings for output voltage in the tabular form of *without feed back* from the CRO, by varying the different steps of frequency (i.e. 10Hz , 500Hz , 1KHz , 100KHz , 200KHz , 400KHz , 600KHz , 1180KHz , 1MHz .) in function generator.
12. We have observed that the readings of *without feed back amplifier's* output voltage is greater than the *with feed back amplifier*.
13. Finally we switched **OFF** the function generator, cathode ray oscilloscope and regulated power supply.
14. Calculated the Voltage gain by using the formula of V_o / V_i and Gain in dB by using the formula of $20\log_{10}(A_v)$ in both tabular forms of *with feed back* and *without feed back amplifiers*.
15. Drawn the graphs of both amplifiers in single graph sheet.

16. While drawing the graph taken the *frequency* on X-axis and *Gain in dB* on Y-axis.
17. Finally calculated the *bandwidth* of both amplifiers from this graph sheet as per the following formulas, i). For *Current shunt feed back amplifier (With feed back)* $(BW) = f_2 - f_1$
 ii). For *Current shunt feed back amplifier (Without feed back)* $(BW) = f_4 - f_3$
18. We have noted down that the *band width of with feed back amplifier* is high as compared to the *without feed back amplifier*.

EXPECTED GRAPH – SOFTWARE & HARDWARE :

The following graph shows for *Current shunt feed back amplifier with feed back* and *without Feedback amplifier* for software as well as hardware.



RESULT – SOFTWARE & HARDWARE :

We drawn the graph for frequency response of a *Current shunt feedback amplifier* for both *with feedback* and *without feedback*.

VIVA VOICE QUESTIONS:

1. What is feedback?
2. What are the input and output impedances for current shunt feedback Amplifier.
3. Applications of current shunt feedback Amplifier.
4. Mention Applications of single Tuned Amplifier.
5. What are the feedback topologies?
6. Example for voltage series feedback amplifier.
7. CC Amplifier characteristics?
8. What is Band Width?
9. What is Frequency Response?
10. Explain the transistor operation with the help of four regions

Experiment No. : 8**Date :****Name of the Experiment : SINGLE TUNED VOLTAGE AMPLIFIER****AIM :**

To obtain the frequency response curve of *Single tuned voltage amplifier* using software and Hardware

APPARATUS :

1. System with Multisim software	-----	1 No.
2. Regulated power supply (RPS)	-----	1 No.
3. Cathode Ray Oscilloscope (CRO)	-----	1 No.
4. Function generator	-----	1 No.
5. Decade Inductance box (DIB)	-----	1 No.
6. Decade capacitance box (DCB)	-----	1 No.
7. Probes	-----	1 No.
8. Breadboard	-----	1 No.
9. Connecting wires	-----	1 No.

COMPONENTS :

1. Transistor	BC 547	-----	1 No.
2. Resistors	47K Ω , 10K Ω , 1 K Ω ,	-----	Each 1No.
3. Capacitors	0.22 μ F,	-----	2 No.
	33 μ F	-----	1 No.
4. Resistors	47K Ω , 10 K Ω , 1K Ω	-----	Each 1 No.

THEORY :

Tuned amplifiers are mainly preferred to amplify the high-frequency signals in wireless communication. The tuned amplification works based on the tuning circuit implied as load. The range of the frequencies defined for a particular amplification circuit can be fixed or dynamic based on applications. The tuning circuit present at the load consists of an inductor and capacitor. For dynamic frequencies, the values of capacitance should be varied. These amplifiers are very advantageous due to its appealing large bandwidths. The increment in bandwidth is based on the number of tuning circuits present at the load. There are three types of most frequently used tuned amplifiers they are single tuned amplifier, double-tuned amplifier and stagger tuned amplifier.

Definition: A tuned amplifier consists of a single tuning circuit at the load can be defined as a single tuned amplifier. It is a multi-stage amplifier, where each stage of this amplifier must be tuned with the same frequencies. For example, tuning a radio station. If the desired carrier wave is passed and matches the defined range of passband frequency, then the radio station is tuned otherwise it is blocked.

CIRCUIT DIAGRAM – SOFTWARE & HARDWARE :

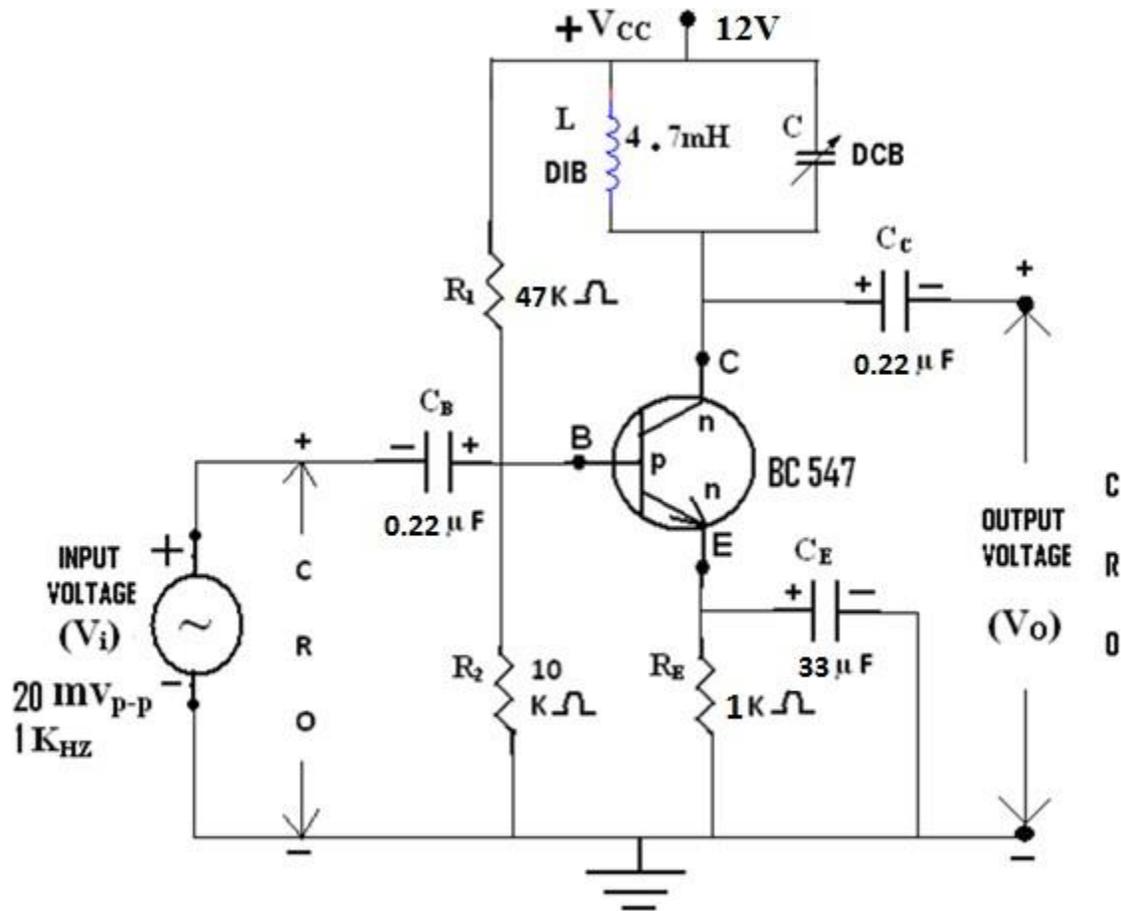


Figure : Circuit diagram of Single tuned voltage amplifier

THEORETICAL CALCULATIONS – SOFTWARE & HARDWARE :

1). When $L=4.7\text{mH}$, $f_r = 10\text{KHz}$, Then $C = ?$

$$\text{We have } f_r = \frac{1}{2\pi\sqrt{LC}} \text{ OR } C = \frac{1}{[2\pi f_r \sqrt{L}]^2}$$

$$\text{OR } C = \frac{1}{4\pi^2} \times \frac{1}{f_r^2 L} \text{ OR } C = \frac{0.0253}{f_r^2 L}$$

$$C = \frac{0.0253}{[10 \times 10^3]^2 \times 4.7 \times 10^{-3}}$$

$$= 54\text{Kpf OR } 54\text{ nF}$$

2). When $L=4.7\text{mH}$, $f = 50\text{KHz}$, Then $C = ?$

$$\text{We have } C = \frac{0.0253}{f_r^2 L}$$

$$= \frac{0.0253}{[50 \times 10^3]^2 \times 4.7 \times 10^{-3}}$$

$$= 2.16\text{Kpf OR } 2.16\text{nF}$$

PROCEDURE – SOFTWARE :

1. We have picked up the components from the components bar as per above circuit.
2. Made the connections as per the above circuit diagram by using the components which we have picked up.
3. Connected the CRO across the Collector capacitor to ground..
4. Kept the $L=4.7\text{mH}$ to take readings in tabular form-1
5. Set the input signal as sine wave form at $20\text{mV}_{\text{P-P}}$, 10KHz . as constant in the function Generator until the experiment would be completed.

6. To simulate the circuit clicked on *run option* through *execute button* in *tool bar*.
7. We could see the *sine wave* on the screen of *CRO* as o/p signal.
8. Varied the capacitance values 100nF, 118nF, 54nF, 40nF, 20nF, 10nF and noted down the peak to peak voltage values of sine wave from the *CRO* connected at o/p side. These values are noted in corresponding columns of the tabular form-1
9. Stopped the simulation by clicking on *run option* through *execute button* in the *tool bar*.
10. Observed the graph for *frequency Vs amplitude* through the *AC Analysis*.
Here we have transmitted the signal at 10KHz. So we could get the max. peak to peak voltage at 54nF, because this ckt. Was tuned at 54nF which we got the resonant frequency 9.99KHz. by using the
$$1/(2\pi\sqrt{LC})$$

 $F_r =$
11. Set the input signal as *sine wave form* at 20mV_{P-P}, 50KHz. as constant in the function Generator until the experiment would be completed.
12. After that again varied the capacitance values 20nF, 10nF, 5nF, 2.16nF, 1nF, 0.5nF and noted down the peak to peak voltage values of sine wave from the *CRO* connected at o/p side. These values are noted in corresponding columns of the tabular form-2
13. Stopped the simulation by clicking on *run option* through *execute button* in the *tool bar*.
14. Observed the graph for *frequency Vs amplitude* through the *AC Analysis*.
15. Here we have transmitted the signal at 50KHz. So we could get the max. peak to peak voltage at 2.16F, because this ckt. Was tuned at 2.16nF which we got the resonant frequency 49.98KHz. by using
$$1/(2\pi\sqrt{LC})$$

 $F_r =$
16. Finally shut down the system safely.
17. Now calculated and noted down the values of *voltage gain*(A_v) and *gain in dB* to the corresponding values of *output voltage*(V_o) & *input voltage*(V_i) by using the formulas given below,
$$\text{Voltage gain } (A_v) = V_o / V_i \quad \text{and} \quad \text{Gain in dB} = 20\log_{10}(A_v).$$
 These values have been noted in the both tabular forms.
18. Plotted the graphs for both tabular forms (frequency response curves) as per given below, a). frequency on X-axis & gain in dB on Y-axis.
b). frequency on X-axis & voltage gain on Y-axis.
19. Calculated and noted the *band width* & *resonant frequency* from both frequency response curves by using the formula, $\text{Band width} = f_2 - f_1$.
20. We have observed that, the graph which is drawn manually is same to the graph which is obtained from the *AC Analysis*.

PROCEDURE – HARDWARE :

1. We have connected the circuit as per the circuit diagram which is shown above. Initially connected the *CRO* across the function generator as per shown in the circuit diagram to set the input signal.
2. Switched *ON* the *CRO* and *function generator*.
3. Applied the input signal as *sine wave form* as 20m_{P-P}, 10KHz. from the function generator by observing in the *CRO*.
4. Later removed the *CRO* and connected it across the capacitor C_C to observe the peak to peak output voltage.

5. Kept the $L=4.7\text{mH}$ in the DIB.
6. Switched *ON* the *RPS* and kept the 12V as V_{CC} .
7. Now Noted down the peak to peak voltage of o/p signal by varied capacitance values- 100nF , 118nF , 54nF , 40nF , 20nF , 10nF from the DCB in the corresponding column of the tabular form-3
8. If we observed, for $C=54\text{nF}$ only we got max. peak to peak voltage, because this ckt. Was tuned at frequency of 10KHz .
9. Again, Applied the input signal as *sine wave form* as $20\text{m}_{\text{p-p}}$, 50KHz . from the function generator by observing in the CRO, and $L=4.7\text{mH}$.
10. Now Noted down the peak to peak voltage of o/p signal by varied capacitance values- 20nF , 10nF , 5nF , 2.16nF , 1nF , 0.5nF from the DCB. In corresponding column of the tabular form-4
11. If we observed, for $C=2.16\text{nF}$ only we got max. peak to peak voltage, because this ckt. Was tuned at frequency of 50KHz .
12. Now calculated and noted down the values of *voltage gain* (A_v) and *gain in dB* to the corresponding values of *output voltage* (V_o) & *input voltage* (V_i) by using the formulas given below,
 - i. *Voltage gain* (A_v) = V_o / V_i and *Gain in dB* = $20\log_{10}(A_v)$. These values has been noted in the both tabular forms-3 & 4.
13. Plotted the graphs for both tabular forms – 3&4 (frequency response curves) as per given below,
 - a). frequency on X-axis & gain in dB on Y-axis.
 - b). frequency on X-axis & voltage gain on Y-axis.
14. Calculated and noted the *band width* & *resonant frequency* from both frequency response curves by using the formula, *Band width* = $f_2 - f_1$.

TABULAR FORM – 1 – SOFTWARE :

Input voltage (V_i) = $20\text{mV}_{\text{p-p}}$ (0.02v) is constant for all readings. When $f_r = 10\text{KHz}$, $C = 54\text{nF}$, $L = 4.7\text{mH}$,							
Sl. No.	Inductance (L) In mH	Capacitance (C) In nF	Capacitance Setting in % $= \frac{\text{Req.Cap.(C)} \times 100}{\text{Max.Cap.}}$	Resonant Frequency $F_r = 1/(2\pi\sqrt{LC})$ in KHz.	Output Voltage (V_o) In Volts.	Voltage gain $A_v = V_o/V_i$	Gain in dB = $20\log_{10}(A_v)$
1	4.7	100	100	7.34			
2	4.7	118	118	8.21			
3	4.7	54	54	9.99			
4	4.7	40	40	11.61			
5	4.7	20	20	16.42			
6	4.7	10	10	23.22			

TABULAR FORM – 2 – SOFTWARE :

Input Voltage (V_i) = 20 mV _{P-P} (0.02V) is constant for all readings. When $f_r = 50\text{Khz.}$, $C = 2.16\text{ nF}$, $L = 4.7\text{mH}$,							
Sl. No.	Inductance (L) In mH	Capacitance (C) In nF	Capacitance Setting in % $= \frac{\text{Req.Cap.(C)} \times 100}{\text{Max.Cap.}}$	Resonant Frequency $F_r = 1/(2\pi\sqrt{LC})$ in KHz.	Output Voltage (V_o) In Volts.	Volt-age gain $A_v = V_o/V_i$	Gain in dB = $20\log_{10}(A_v)$
1	4.7	20	20	16.42			
2	4.7	10	10	23.23			
3	4.7	5	5	32.85			
4	4.7	2.16	2.16	49.98			
5	4.7	1	1	73.45			
6	4.7	0.5	0.5	103.87			

TABULAR FORM – 3 – HARDWARE :

Input Voltage (V_i) = 20 mV _{P-P} (0.02V) is constant for all readings. When $f_r = 10\text{Khz.}$, $C = 54\text{ nF}$, $L = 4.7\text{mH}$,						
Sl. No.	Inductance (L) In mH	Capacitance (C) In nF	Resonant Frequency $F_r = 1/(2\pi\sqrt{LC})$ in KHz.	Output Voltage (V_o) In Volts.	Volt-age gain $A_v = V_o/V_i$	Gain in dB = $20\log_{10}(A_v)$
1	4.7	100	7.34			
2	4.7	118	8.21			
3	4.7	54	9.99			
4	4.7	40	11.61			
5	4.7	20	16.42			
6	4.7	10	23.22			

TABULAR FORM – 4 – HARDWARE :

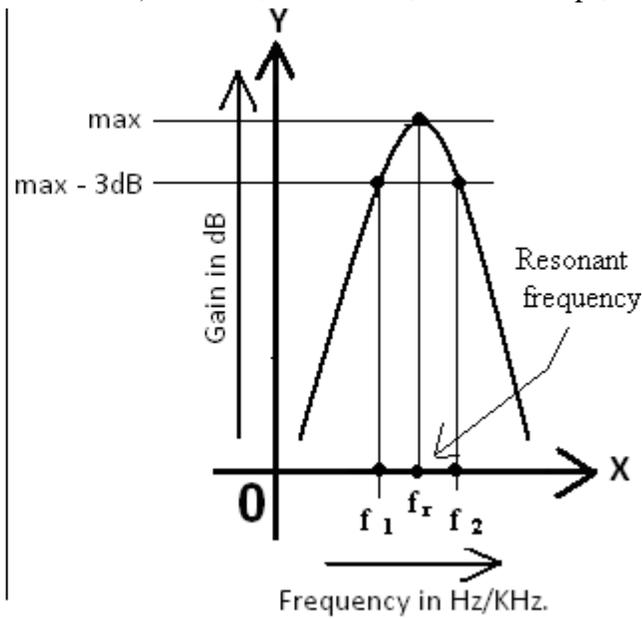
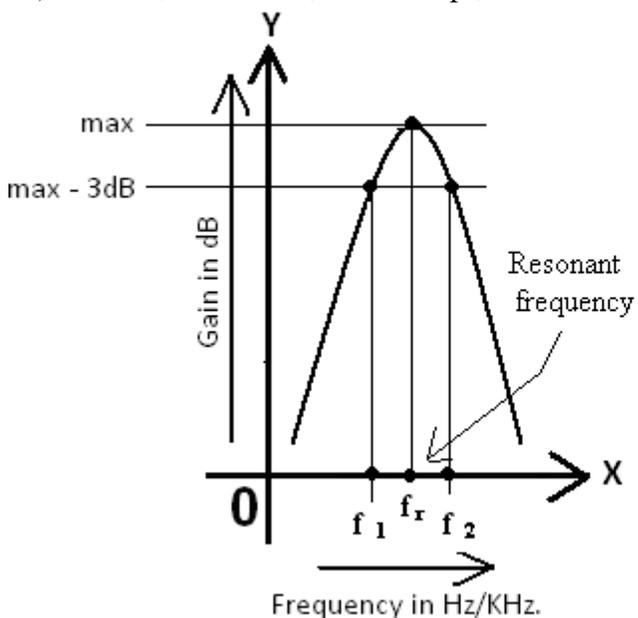
Input Voltage (V_i) = 20 mV _{P-P} (0.02V) is constant for all readings. When $f_r = 50\text{Khz.}$, $C = 2.16 \text{ nF}$, $L = 4.7\text{mH}$,						
Sl. No.	Inductance (L) In mH	Capacitance (C) In nF	Resonant Frequency $F_r = 1/(2\pi\sqrt{LC})$ in KHz.	Output Voltage (V _o) In Volts.	Voltage gain $A_v = V_o/V_i$	Gain in dB = $20\log_{10}(A_v)$
1	4.7	20	16.42			
2	4.7	10	23.23			
3	4.7	5	32.85			
4	4.7	2.16	49.98			
5	4.7	1	73.45			
6	4.7	0.5	103.87			

EXPECTED GRAPH – SOFTWARE & HARDWARE :

The following graphs shows the frequency response curve for *single tuned voltage amplifie*

A). When $f_r = 10 \text{ KHz.}$, $C = 54 \text{ Kpf}$, $L = 4.7\text{mH}$

B). When $f_r = 50 \text{ KHz.}$, $C = 2.16 \text{ Kpf}$, $L = 4.7\text{mH}$



PRACTICAL CALCULATIONS – SOFTWARE & HARDWARE :

When $f_r = 10 \text{ KHz.}$, $C = 54 \text{ Kpf}$, $L = 4.7\text{mH}$	When $f_r = 50 \text{ KHz.}$, $C = 2.16 \text{ Kpf}$, $L = 4.7\text{mH}$
1). Band width = $f_2 - f_1$ =	1). Band width = $f_2 - f_1$ =
2). Resonant frequency (f_r) =	2). Resonant frequency (f_r) =

TABULAR FORM - 5 – SOFTWARE & HARDWARE :

The following tabular form shows the comparison between the theoretical and practical resonant frequency values.

Sl. No.	Inductr (L) (Note down From the theoreti calcalcul ations)	Capacitor (C) (Note down From the theoretical calculate ions)	Theoretica l Resonant frequency (f_r) (Note down From the theoretical calculation s)	Software		Hardware	
				Practical Resonant frequency (f_r)(Note down from the graph)	Max. voltge gain in dB at resonant frequency. (Note down from the graph)	Practical Resonant frequency (f_r) (Note down from the graph)	Max. voltage gain in dB at resonant frequency. (Note down from the graph)
1.	4.7mH	54Kpf	10KHz.				
2.	4.7mH	2.16Kpf	50KHz.				

CONCLUSSION – SOFTWARE & HARDWARE :

If I observed in the tabular form-5 the voltage gain of the output signal is maximum when the practical resonant frequency value is approximately equal to the theoretical resonant frequency value.

APPLICATIONS – SOFTWARE & HARDWARE :

Mainly uses in the radio receivers to tuned the appropriate signal / station which is transmitted in relay station.

RESULT – SOFTWARE & HARDWARE :

I have drawn the frequency response curve and calculated the values of band width, and resonant frequency of a *single tuned voltage amplifier*.

VIVA VOICE Questions:

1. What is single Tuned Amplifier?
2. What is Q factor?
3. What is tank circuit?
4. Mention Applications of single Tuned Amplifier.
5. What is the resonant frequency of single tuned Amplifier?
6. Tuned Amplifier is Narrow or Wide BW Amplifier?
7. Difference between single tuned and double tuned Amplifier?
8. What is stagger tuned Amplifier?
9. Effect of cascading of single tuned Amplifier on BW?
10. What is frequency response?

Experiment No. 9**Date :****Name of the Experiment : CLASS A POWER AMPLIFIER****AIM :**

1. To verify / plot the output signal (sine wave form) of a given *Class-A Series-fed Power Amplifier* by using software & hardware .
2. To calculate the conversion efficiency of a given amplifier.

APPARATUS :

1. System with Multisim software	-----	1 No.
2. Regulated Power Supply	(0-30)V, 1A -----	1 No.
3. Function generator	1MHz. -----	1 No.
4. Probes	----- -----	1 No.
5. Bread board	----- -----	1 No.
6. Connecting wires	----- -----	A few Nos.
7. Ammeters	(0-10)mA DC Type	1 No.

COMPONENTS :

1. Transistors	BC 547 -----	1 No.
2. Resistors	1K Ω , 10 K Ω , 47 K Ω	Each 1 No.
3. Capacitors	0.22 μ F -----	2 No.
	33 μ F -----	1 No.

THEORY :

Class A power amplifier is a type of power amplifier where the output transistor is ON full time and the output current flows for the entire cycle of the input wave form. Class A power amplifier is the simplest of all power amplifier configurations. They have high fidelity and are totally immune to crossover distortion. Even though the class A power amplifier have a handful of good feature, they are not the prime choice because of their poor efficiency. Since the active elements (transistors) are forward biased full time, some current will flow through them even though there is no input signal and this is the main reason for the inefficiency

The theoretical maximum efficiency of a Class A power amplifier is 50%. In practical scenario, with capacitive coupling and inductive loads (loud speakers), the efficiency can come down as low as 25%. This means 75% of power drawn by the amplifier from the supply line is wasted. Majority of the power wasted is lost as heat on the active elements (transistor).As a result, even a moderately powered Class A power amplifier require a large power supply and a large heatsink.

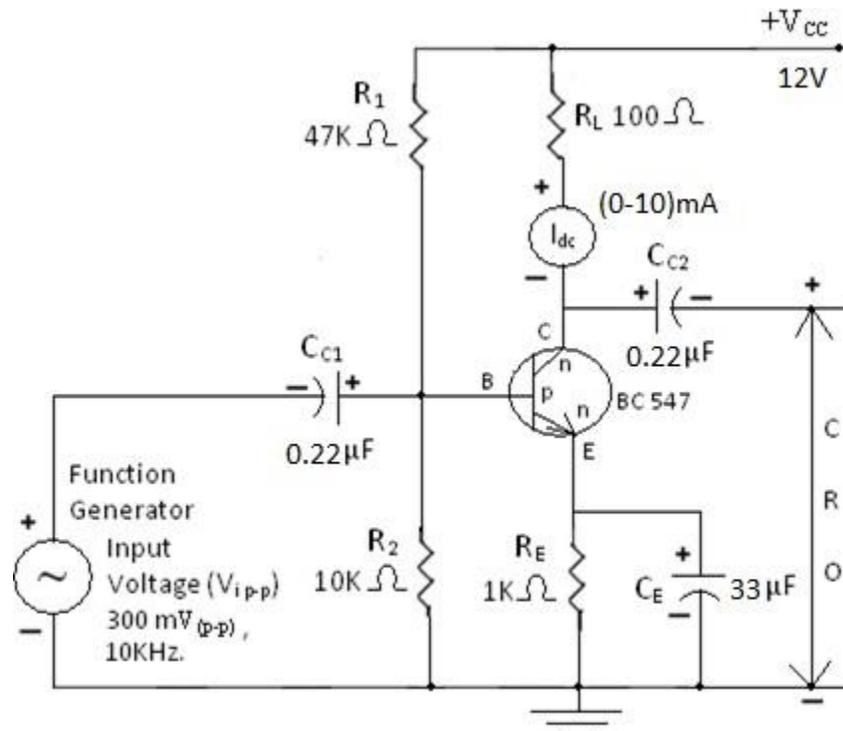
CIRCUIT DIAGRAM :

Figure : Circuit diagram of Class-A Power amplifier

PROCEDURE – SOFTWARE :

1. Picked up the components from *components bar* in multisim software as per the circuit diagram.
2. Made the connections as per the circuit diagram.
3. Set the 300 mV_{p-p} (as input voltage) , 10 KHz (as input frequency) *sine wave* signal to the circuit from the *Function generator* .
4. Noted down the *Input voltage*(v_i) , *Input frequency* against the corresponding columns of the tabular form of *practical calculations*.
5. Set the *supply voltage* 12V as V_{CC} to the circuit as shown in the circuit diagram.
6. To simulate this circuit click on *Run* button in *tool bar*.
7. Observed the *sine wave* signal in CRO and drawn this signal on the graph sheet.
8. Calculated the *output voltage* ($V_{O_{p-p}}$) , *time period* (T) , *frequency* (f) from the graph, and noted down these values against the corresponding columns in the tabular form of *practical calculations*.
9. Noted down the *supply voltage* (V_{CC}) and *collector dc current* I_{dc} at *Quiescent* condition i.e. when no signal is applied i.e. by disconnected the *function generator* from the circuit against the corresponding columns of the tabular form of *practical calculations*.
10. Stop the simulation by click on *Run* button in *tool bar*.
11. Shut down the system safely.
12. Later calculated and noted the input *dc power* $P_i(dc)$, *output ac power* $P_o(ac)$ and % of *efficiency* (η) by using the formulas which are mentioned in the corresponding columns of the tabular form of *practical calculations*.
13. Noted that the practical value should be less than the *Typical Max. efficiency value* i.e. **25.4%**.

PROCEDURE – HARDWARE :

1. Connections are made as per the circuit diagram.
2. Initially connected the *CRO* across the *Function generator*.
3. Switched **ON** the Cathode ray oscilloscope (*CRO*) and *Function generator*.
4. Applied the 300 mV_{p-p} , 10 Khz *sine wave* signal to the circuit from the *Function generator* by observing on the *crt* of the *CRO*.
5. Later connected the *CRO* across R_L i.e at output side.
6. Now switched **ON** the Regulated Power Supply (*RPS*) and apply the *supply voltage* 12V as V_{CC} to the circuit as per shown in the figure.
7. Observed the *sine wave* signal on the *CRT* of the *CRO* and draw this signal on the graph sheet.
8. Now noted down the *collector dc current* I_{dc} at *Quiescent* condition i.e. when no signal is applied and *supply voltage* (V_{CC}) by disconnected the *function generator* from the circuit against the corresponding columns in the tabular form of *practical calculations*.
9. Switched **OFF** the *function generator*, *RPS*, *CRO*.
10. Noted down the *Input voltage* (V_i), *Input frequency* against the corresponding columns in the tabular form.
11. Calculated the *output voltage* ($V_{O(p-p)}$), *time period* (T), *frequency* (f) from the graph, and noted down these values against the corresponding columns in the tabular form.
12. Later calculated the *Input dc power* $P_i(dc)$, *output ac power* $P_o(ac)$ and % of *efficiency* (η) by using the formulas which are mentioned in the corresponding columns in the tabular form.
13. Noted that The practical value should be less than the *Typical Max. efficiency value* i.e. **25.4%**.

PRACTICAL CALCULATIONS – SOFTWARE & HARDWARE :

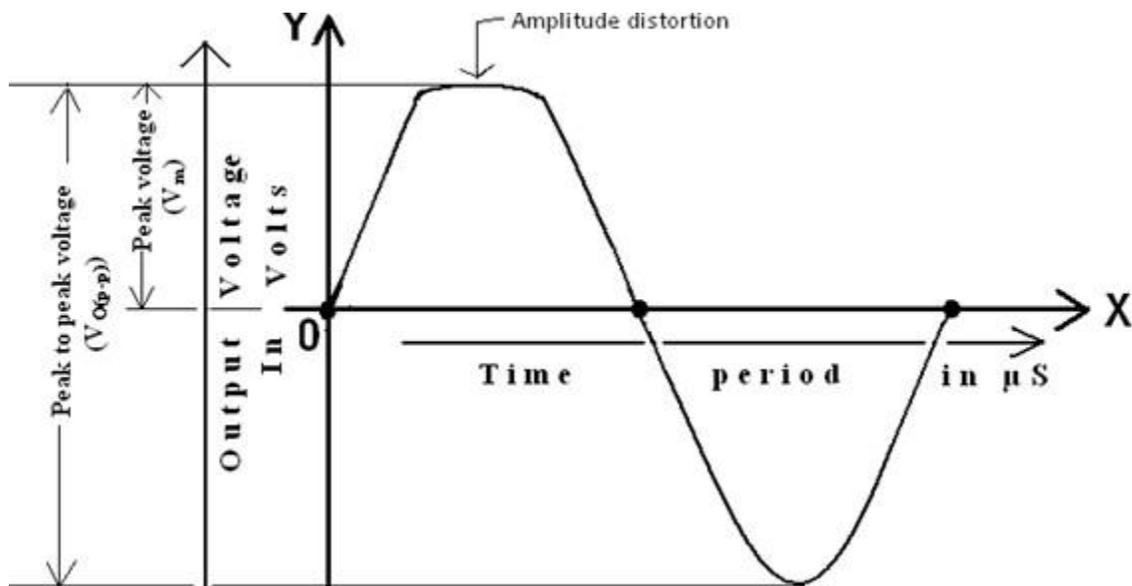
The practical calculations for the parameters are shown in the following tabular form.

Sl.No.	Name of the parameter	Value from Software	Value From Hardware
01.	Input Voltage (V_i) _{p-p} (In mV).	300	
02	Input frequency (In Khz.).	10	
03	Supply DC Voltage (V_{CC}) (in Volts.)	10	
04	Output voltage $V_{O(p-p)}$ (In volts.).		
05	Time period (T) for output signal (In ms)		
06	Fequency for output signal = $1/T$ (In Khz.)		
07	Collector dc current (I_{dc}) (At quesient condition i.e. When no input signal is applied) (In mA.).		

Sl.No.	Name of the parameter	Value from Software	Value From Hardware
08	Collector DC current when sine wave (AC) signal is applied as input signal (I_{ac})		
09	Input DC power $P_i(dc) = I_{dc} \times V_{CC}$ (In Watts).		
10	Output ac power $P_o(ac) = \frac{V_{O(P-P)}^2}{8R_L}$ (In Watts)		
11	% of efficiency (η) = $[P_o(ac) / P_i(dc)] \times 100 =$		
12	Typical Max. efficiency (η) =	25.40%	

EXPECTED WAVEFORM – SOFTWARE & HARDWARE :

The following waveform shows the output signal of *Class A Series-fed Power Amplifier* .



RESULT – SOFTWARE & HARDWARE :

I have verified / drawn the output signal and calculated the conversion efficiency of given *Class-A Series-fed Power amplifier*.

VIVA VOICE Questions:

1. What is Power Amplifier?
2. Classifications of power Amplifiers.
3. Efficiency of class A power Amplifier.
4. Difference between Direct coupled and Transformer coupled class A power Amplifier?
5. What is the amplitude (Harmonic) Distortion?
6. Where is the Q point in class A power Amplifier?
7. Applications of class A power Amplifier.
8. What are the disadvantages of class A power Amplifier.
9. Mention the conduction angle of class A power Amplifier.
10. What are the disadvantages of class A power Amplifier.

Experiment No. 10**Date :**

**Name of the Experiment : COMPLEMENTARY SYMMETRY PUSH PULL
CLASS B POWER AMPLIFIER**

AIM :

1. To verify / plot the output signal (sine wave form) of a given *Class-B Power Amplifier* by using software & hardware .
2. To calculate the conversion efficiency of a given amplifier.

APPARATUS :

1. System with Multisim software	-----	1 No.
2. Regulated Power Supply	(0-30)V, 1A -----	1 No.
3. Function generator	1MHz. -----	1 No.
4. Probes	-----	1 No.
5. Bread board	-----	1 No.
6. Connecting wires	-----	A few Nos.
7. Ammeters	(0-10)mA DC Type -----	1 No.

COMPONENTS :

1. Transistors	BC 547, Bc557-----	Each 1 No.
2. Resistors	220K Ω , 18 K Ω	Each 2 No.
	1K Ω -----	1 No.
	10 Ω -----	3 No.
3. Capacitors	10 μ F -----	2 No.

THEORY :

Class B amplifier is a type of power amplifier where the active device (transistor) conducts only for one half cycle of the input signal. That means the conduction angle is 190° for a Class B amplifier. Since the active device is switched off for half the input cycle, the active device dissipates less power and hence the efficiency is improved. Theoretical maximum efficiency of Class B power amplifier is 78.5%.

Class-B or Push-pull amplifiers use two “complementary” or matching transistors, one being an NPN-type and the other being a PNP-type with both power transistors receiving the same input signal together that is equal in magnitude, but in opposite phase to each other. This results in one transistor only amplifying one half or 190° of the input waveform cycle while the other transistor amplifies the other half or remaining 190° of the input waveform cycle with the resulting “two-halves” being put back together again at the output terminal. Then the conduction angle for this type of amplifier circuit is only 190° or 50% of the input signal. This pushing and pulling effect of the alternating half cycles by the transistors gives this type of circuit its amusing “push-pull” name, but are more generally known as the **Class B Amplifier**

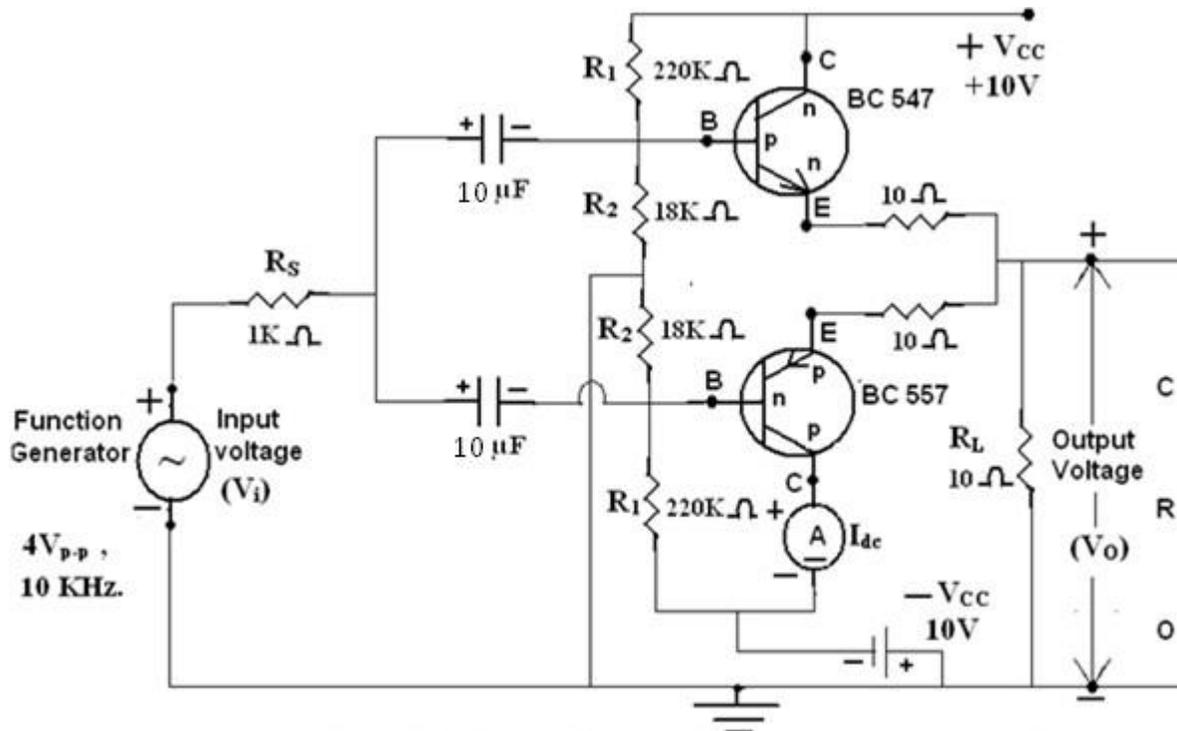
CIRCUIT DIAGRAM :

Figure: Circuit diagram of Class-B complimentary symmetry power amplifier.

PROCEDURE – SOFTWARE :

1. Picked up the components from *components bar* in multisim software as per the circuit diagram.
2. Made the connections as per the circuit diagram.
3. Set the $4 V_{p-p}$ (as input voltage) , $10 KHz$ (as input frequency) *sine wave* signal to the circuit from the *Function generator*.
4. Noted down the *Input voltage*(V_i) , *Input frequency* against the corresponding columns of the tabular form of *practical calculations*.
5. Set the *supply voltage* $20V$ as V_{CC} to the circuit as shown in the circuit..
6. To simulate this circuit click on *Run* button in *tool bar*.
7. Observed the *sine wave* signal in CRO_2 and drawn this signal on the graph sheet.
8. Calculated the *output voltage* ($V_{O_{p-p}}$) , *time period* (T) , *frequency* (f) from the graph, and noted down these values against the corresponding columns of the tabular form of practical calculations.
9. Noted down the *supply voltage* (V_{CC}) and *collector dc current* I_{dc} at *Quiescent* condition i.e. when no signal is applied i.e. by disconnected the *function generator* from the circuit against the corresponding columns of the tabular form of *practical calculations*.
10. Stop the simulation by click on *Run* button in *tool bar*.
11. Shut down the system safely.
12. Later calculated and noted the *Input dc power* $P_i(dc)$, *output ac power* $P_o(ac)$ and % of *efficiency* (η) by using the formulas which are mentioned in the corresponding columns of the tabular form of practical calculations.
13. Noted that The practical value should be less than the *Typical Max. efficiency value* i.e. **78.5%**.

PROCEDURE – HARDWARE :

1. Connections are made as per the circuit diagram.
2. Initially connected the *CRO* across the *Function generator*.
3. Switched **ON** the Cathode ray oscilloscope (CRO) and Function generator.
4. Applied the $4V_{p-p}$, 10 KHz sine wave signal to the circuit from the *Function generator* by observing on the *crt* of the *CRO*.
5. Later connected the *CRO* across R_L i.e. at output side.
6. Now switched **ON** the Regulated Power Supply (RPS) and apply the *supply voltage* +10V from one channel (+ V_{CC}) and -10V from another ($-V_{CC}$) to the circuit as per shown in the figure.
7. Observed the *sine wave* signal on the CRT of the CRO and draw this signal on the graph sheet.
8. Now noted down the *collector dc current* I_{dc} at *Quiescent* condition i.e. when no signal is applied by disconnected the *function generator* from the circuit and *supply voltage* (V_{CC}) against the corresponding columns of the tabular form of *practical calculations*.
9. Noted down the *Input voltage* (V_i), *Input frequency* against the corresponding columns of the tabular form of *practical calculations*.
10. Switched **OFF** the *function generator*, *RPS*, *CRO*.
11. Calculated the peak to peak voltage ($V_{O(p-p)}$), *peak voltage* (V_m), *time period* (T), *frequency* (f) from the graph, and noted down these values against the corresponding columns of the tabular form of practical calculations.
12. Later calculated the *Input dc power* $P_i(dc)$, *output ac power* $P_o(ac)$ and % of *efficiency* (η) by using the formulas which are mentioned in the corresponding columns of the tabular form of practical calculations.
13. Noted that The practical value should be less than the *Typical Max. efficiency value* i.e. **78.5%**.

PRACTICAL CALCULATIONS – SOFTWARE & HARDWARE :

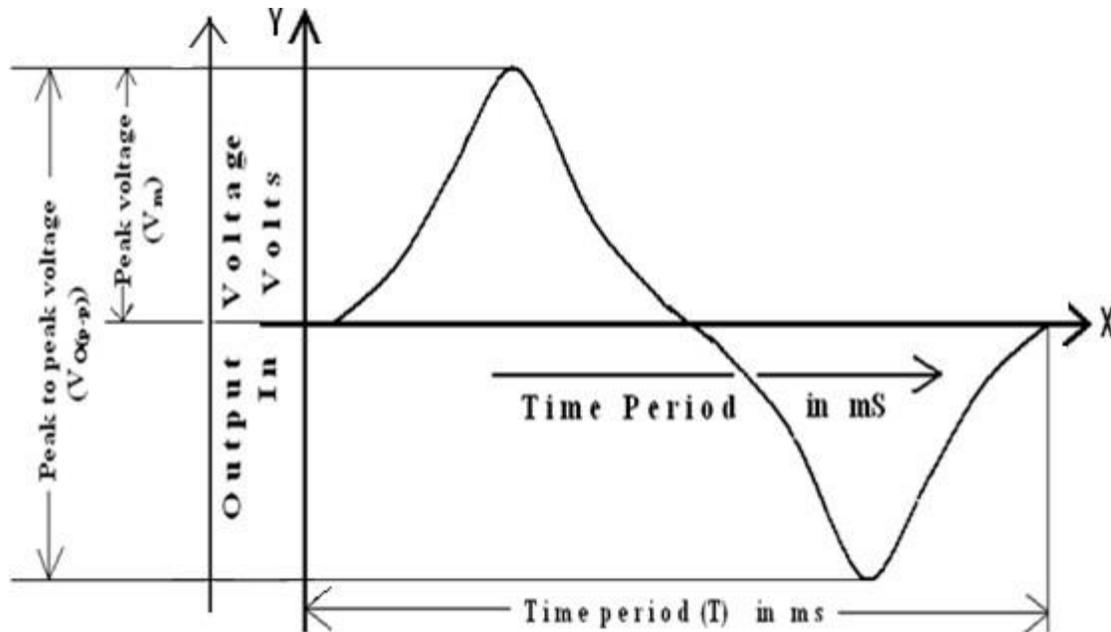
The practical calculations for the parameters are shown in the following tabular form,

Sl. No.	Name of the parameter	Value from software	Value from hardware
01.	Input peak to peak voltage (V_i) (In Volts).	4	4
02	Input frequency (In Khz.).	10	10
03	Positive supply DC Voltage ($+V_{CC}$) (in Volts.)	10	10
	Negative supply DC Voltage ($-V_{CC}$) (in Volts.)	10	10
04	Peak to peak voltage of output $V_{O(p-p)}$ (In volts.).		
05	Peak voltage of output (V_m) = $V_{O(p-p)}/2$ (In volts).		
06	Time period (T) for output signal (In ms)		
07	Frequency for output signal = $1/T$ (In Khz.)		
08	Collector dc current (I_{dc}) (At quiescent condition i.e. When no input signal is applied) (In mA.).		
09	Collector DC current when sine wave (AC) signal is applied as input signal (I_{ac})		

Sl. No.	Name of the parameter	Value from software	Value from hardware
10	Input DC power $P_i(dc) = I_{dc} \times V_{CC}$ (In mWatts).		
11	Output ac power $P_o(ac) = \frac{V_m^2}{2R_L}$ (In mWatts) =		
12	% of efficiency $(\eta) = \frac{P_o(ac)}{P_i(dc)} \times 100$		
13	Typical Max. efficiency $(\eta) =$	78.50 %	

EXPECTED GRAPH - SOFTWARE & HARDWARE :

The following graph shows for *Class B complementary symmetry power amplifier*.



RESULT - SOFTWARE & HARDWARE :

I have drawn the graph for output signal and calculated the conversion efficiency of given complementary symmetry Class-B push-pull power amplifier.

VIVA VOICE Questions:

1. What is Power Amplifier?
2. Classifications of power Amplifiers.
3. Efficiency of class B power Amplifier.
4. Difference between Transformer coupled and Complementary symmetry class B power Amplifier?
5. What is the Crossover Distortion?
6. Where is the Q point in class B power Amplifier?
7. Applications of class B power Amplifier.
8. What are the disadvantages of class B power Amplifier.
9. Mention the conduction angle of class B power Amplifier.
10. What are the disadvantages of class B power Amplifier.

Experiment No. : 11.A

Date :

Name of the Experiment : RC PHASE SHIFT OSCILLATOR

AIM :

To verify the *sine wave form* and to calculate its frequency values of a given *RC Phase shift Oscillator*

by using software and hardware.

APPARATUS :

- | | |
|--|------------|
| 1. System with Multisim Software ----- | 1 No. |
| 2. Regulated power supply (RPS)----- | 1 No. |
| 3. Cathode ray oscilloscope ----- | 1 No. |
| 4. Decade Resistance Box (DRB)----- | 1 No. |
| 5. Decade Capacitance Box (DCB)----- | 1 No. |
| 6. Bread board ----- | 1 No. |
| 7. Connecting wires ----- | A few Nos. |

COMPONENTS :

- | | |
|-------------------------------------|-------|
| 1. Resistors : 1K Ω | 1 No. |
| 4.7 K Ω | 1 No. |
| 47 K Ω | 1 No. |
| 10 K Ω | 3 No. |
| 2. Capacitors : 0.047 μ F ----- | 1 No. |
| 1000 μ F | 1 No. |
| 3. Transistor : BC547 ----- | 1 No. |

THEORY :

A phase shift oscillator can be defined as; it is one kind of linear oscillator which is used to generate a sine wave output. It comprises of an inverting amplifier component like operational amplifier otherwise a transistor. The output of this amplifier can be given as input with the help of the phase shifting network. This network can be built with resistors as well as capacitors in the form of a ladder network. The phase of the amplifier can be shifted to 180° at the oscillation frequency by using a feedback network to provide a positive response. These types of oscillators are frequently used as audio oscillators on audio frequency. This article discusses an overview of RC phase shift oscillator.

RC phase-shift oscillator circuit can be built with a resistor as well as a capacitor. This circuit offers the required phase shift with the feedback signal. They have outstanding frequency strength and can give a clean sine wave for an extensive range of loads. Preferably an easy RC network can be expected to include an o/p which directs the input with 90°.

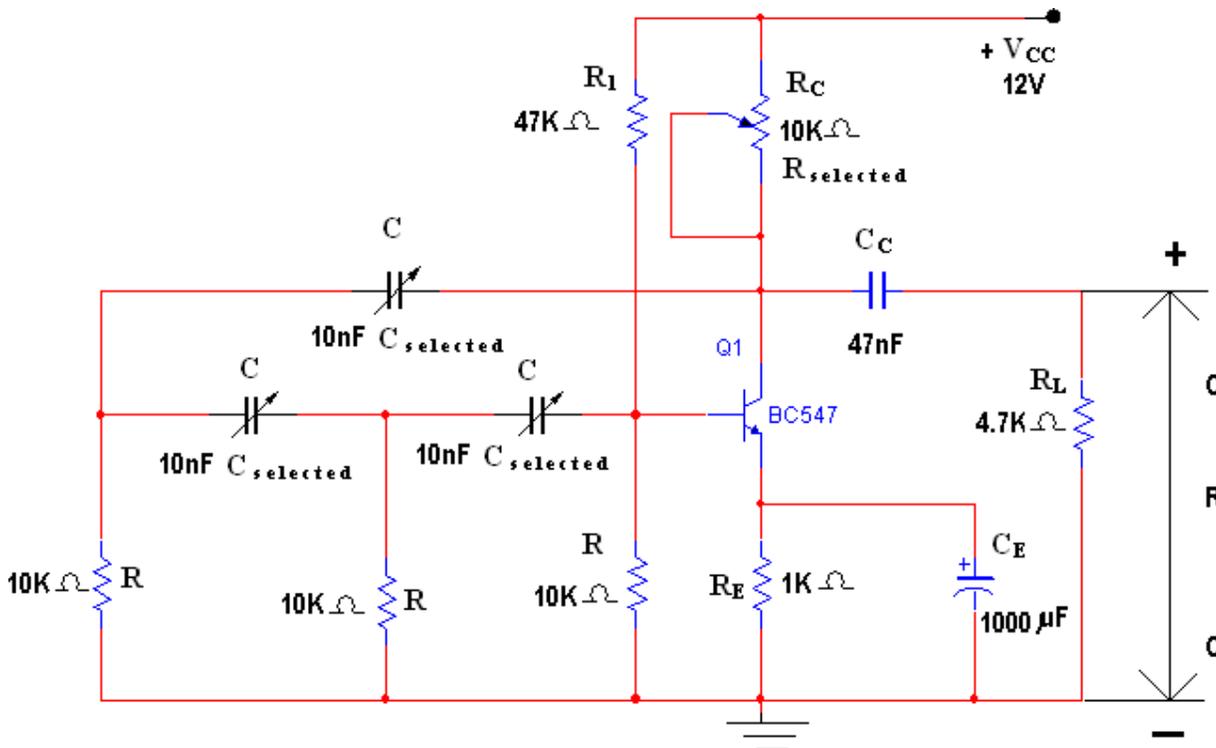
CIRCUIT DIAGRAM – SOFTWARE & HARDWARE :

Figure: Circuit diagram of RC Phase shift oscillator.

PROCEDURE – SOFTWARE :

1. Picked up the components from *components bar* in multisim software as per the circuit diagram.
2. Made the connections as per the circuit diagram.
3. Set the V_{CC} value as 12V.
4. Initially set the *Capacitor C* values as 1nF (0.001μF or 1Kpf) by varied the *capacitor C* Value in % by using the following formula,

$$\text{Setting in \%} = \frac{C_{\text{required}} \times 100}{C_{\text{selected}}}$$

5. To start the simulation clicked on *Run button*.
6. Varied the R_C value until we get *sine wave form* which is consist the $V_{O(p-p)}$ is approximately 6V because this circuit is designed to get the output voltage as $6V_{(p-p)}$ in the *CRO*.
7. We observed *Sine wave form* as a output signal in the *CRO*.
8. Drawn the *sine wave form* on the graph by taking the *time period* on X-axis and *Amplitude ($V_{O(p-p)}$)* on Y-axis.
9. Calculated and noted the *collector resistor (R_C)* and *theoretical frequency (f_o)* value for corresponding *capacitor C* values in the tabular form by using the formulas which are given below,

$$R_c = R_{\text{selected}} - \left[\frac{\text{Setting in \%} \times R_{\text{selected}}}{100} \right]$$

$$f_o = \frac{1}{2\pi RC \sqrt{6+4(R_c/R)}}$$

11. Calculated the frequency and output voltage ($V_{O(p-p)}$) values from the graph then noted in the Columns of *practical frequency* and *output voltage* for corresponding *capacitor C* values in the tabular form respectively.
12. Stopped the simulation by click on *Run option* through *Execute button*.
13. Repeat the same procedure from points 4 to 11 for corresponding *C* values which are given below,
 - a). 2.2 nF (0.0022 μ F or 2.2Kpf).
 - b). 3.3 nF (0.0033 μ F or 3.3Kpf).
 - c). 10.0 nF (0.01 μ F or 10Kpf).
14. Shut down the system safely.
15. We compared that *theoretical frequency* value (f_o) and *practical frequency* value are same approximately.

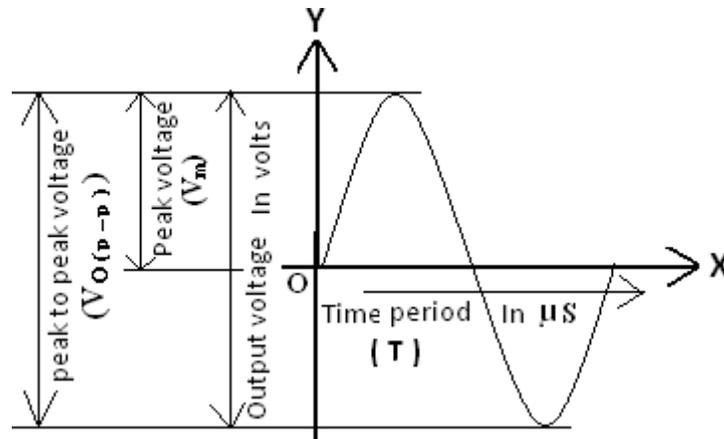
PROCEDURE – HARDWARE :

1. Made the connections as per the circuit diagram.
2. Kept the V_{CC} value as 12V.
3. Kept the *Capacitor C* values as 1nF (0.001 μ F or 1Kpf) in DCB.
4. Varied the R_C (i.e. Appx. 4.3K Ω) until we get *sine wave form* which is consist the $V_{O(p-p)}$ is approximately 6V because this circuit is designed to get the output voltage as 6V_(p-p) in the *CRO*.
5. Now noted the value of R_C to the corresponding *C* value in tabular form.
6. We observed the *Sine wave form* as a output signal in the *CRO*.
7. Now calculated and noted the *theoretical frequency value* (f_o) to the corresponding *C* value in the tabular form by using the formula given below,

$$\frac{1}{2\pi RC \sqrt{6+4(R_c/R)}}$$
8. Drawn the *sine wave form* on the graph by taking the *time period* on X-axis and *amplitude* ($V_{O(p-p)}$) on Y-axis.
9. Calculated the frequency and output voltage ($V_{O(p-p)}$) values from the graph then noted in the Columns of *practical frequency* and *output voltage* in the tabular form respectively.
10. Repeat the same procedure from points 4 to 9 for corresponding *C* values which are given below,
 - a). 2.2 nF (0.0022 μ F or 2.2Kpf).
 - b). 3.3 nF (0.0033 μ F or 3.3Kpf).
 - c). 10.0 nF (0.01 μ F or 10Kpf).
11. Switch *OFF* the RPS and *CRO*.
12. We compared that *theoretical frequency* value (f_o) and *practical frequency* values are approximately same.

EXPECTED WAVEFORM – SOFTWARE & HARDWARE :

The following waveform shows the output signal for different capacitor values of *RC phase shift Oscillator*,

**CALCULATIONS – SOFTWARE :**

Sl. No.	Resistor (R) In K Ω	Capacitor (C) In Kpf	$R_C = R_{\text{selected}} - \left[\frac{\text{Setting in \%} \times R_{\text{selected}}}{100} \right]$ In K Ω	Theoretical frequency (f_0) $\frac{1}{2\pi RC \sqrt{6+4(R_C/R)}}$ In Hz/KHz.	Practical Time Period (In μ S)	Practical Frequency In Hz/KHz.	Output Voltage ($V_{O(p-p)}$) In Volts
1	10	1					
2	10	2.2					
3	10	3.3					
4	10	10					

CALCULATIONS – HARDWARE :

Sl. No.	Resistor (R) In K Ω	Capacitor (C) In Kpf	R_C In K Ω	Theoretical frequency (f_0) $\frac{1}{2\pi RC \sqrt{6+4(R_C/R)}}$ In Hz/KHz.	Practical Time Period (In μ S)	Practical Frequency In Hz/KHz.	Output Voltage ($V_{O(p-p)}$) In Volts
1	10	1					
2	10	2.2					
3	10	3.3					
4	10	10					

RESULT : I have verified / drawn the output signal and calculated the frequency values of a given *RC phase shift oscillator*.

Experiment No. : 11.B**Date :****Name of the Experiment : WEIN BRIDGE OSCILLATOR****AIM :**

To draw the *sine wave form* and to calculate its frequency values of a given *Wein bridge oscillator*

using software & hardware

APPARATUS :**Software :**

1. System ----- 1 No.
2. Multisim software

Hardware :

1. Regulated power supply (RPS)----- 1 No.
2. Cathode ray oscilloscope ----- 1 No.
3. Decade Resistance Box (DRB) ----- 1 No.
4. Decade Capacitance Box (DCB) ----- 1 No.
5. Probes ----- 1 No.
6. Connecting wires ----- 1 No.

COMPONENTS :

1. Transistor BC 547 ----- 2 No.
2. Resistors 4.7K Ω , 10K Ω , 47K Ω ----- 1 No.
- 1K Ω , 8.2K Ω , ----- 3 No.
- 8.2K Ω , ----- 2 No.
2. Capacitors 4.7 μ F ----- 1 No.

THEORY :

The simplest sine wave oscillators which uses a RC network in place of the conventional LC tuned tank circuit to produce a sinusoidal output waveform, is called a **Wien Bridge Oscillator**.

The **Wien Bridge Oscillator** is so called because the circuit is based on a frequency-selective form of the Wheatstone bridge circuit. The Wien Bridge oscillator is a two-stage RC coupled amplifier circuit that has good stability at its resonant frequency, low distortion and is very easy to tune making it a popular circuit as an audio frequency oscillator but the phase shift of the output signal is considerably different from the previous phase shift **RC Oscillator**.

Wien Bridge Oscillator Frequency

$$f_r = \frac{1}{2\pi RC}$$

Where:

f_r is the Resonant Frequency in Hertz

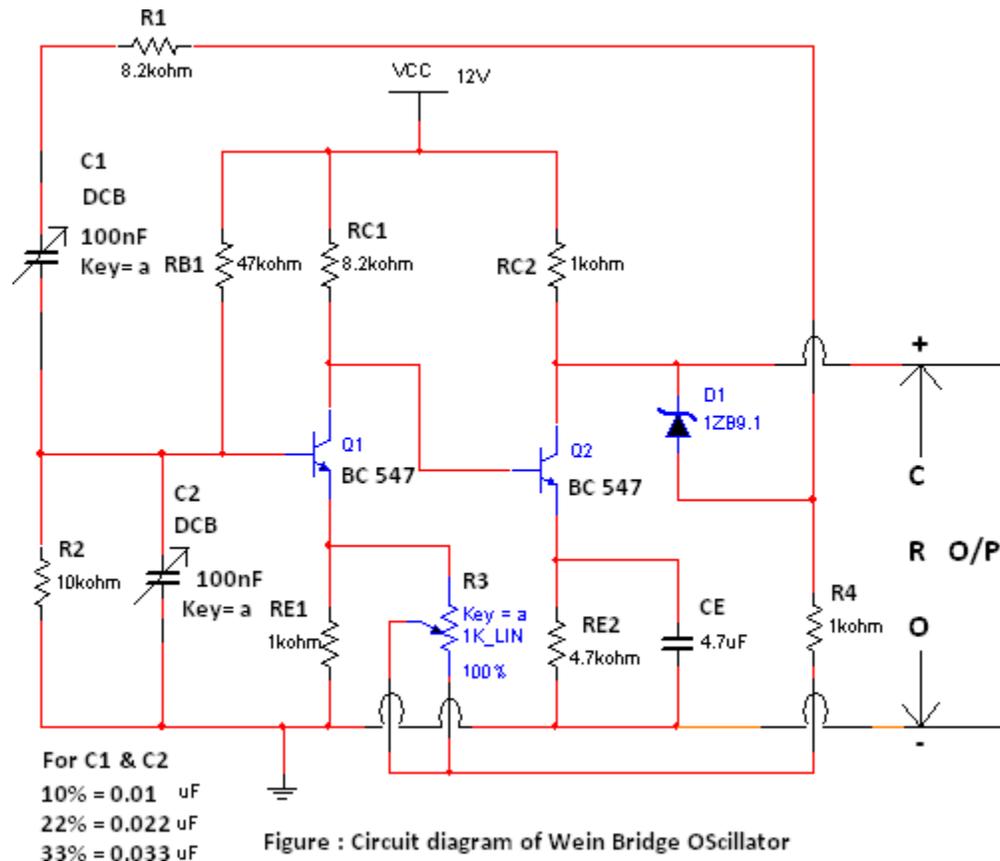
R is the Resistance in Ohms

C is the Capacitance in Farads

Some of the applications of the Wein bridge oscillator as given below. These are highly used for audio testing. Clock signals for testing filter circuits can be generated by this oscillator. Used in distortion testing of power amplifiers.

Due to the advantages like good frequency stability, very low distortion and ease of tuning, a Wien bridge oscillator becomes the most popular audio frequency range signal generator circuit.

CIRCUIT DIAGRAM :



PROCEDURE – SOFTWARE :

1. Picked up the components from *components bar* in multisim software as per the circuit diagram.
2. Made the connections as per the circuit diagram.
3. Set the V_{CC} value as 12V.
4. Initially set the *Capacitor C* values as 1nF (0.001 μ F or 1Kpf) by varied the *capacitor C* Value in % by using the following formula,

$$\text{Setting in \%} = \frac{C_{\text{required}} \times 100}{C_{\text{selected}}}$$

5. To start the simulation clicked on *Run button*.
6. Varied the R_3 value until we get *sine wave form* which is consist the $V_{O(p-p)}$ is approximately 2.4V because this circuit is designed to get the output voltage as $2.4V_{(p-p)}$ in the *CRO*.
- 7.. We observed *Sine wave form* as a output signal in the *CRO*.
8. Drawn the *sine wave form* on the graph by taking the *time period* on X-axis and *Amplitude ($V_{O(p-p)}$)* on Y-axis.

TABULAR FORM / CALCULATIONS - SOFTWARE :

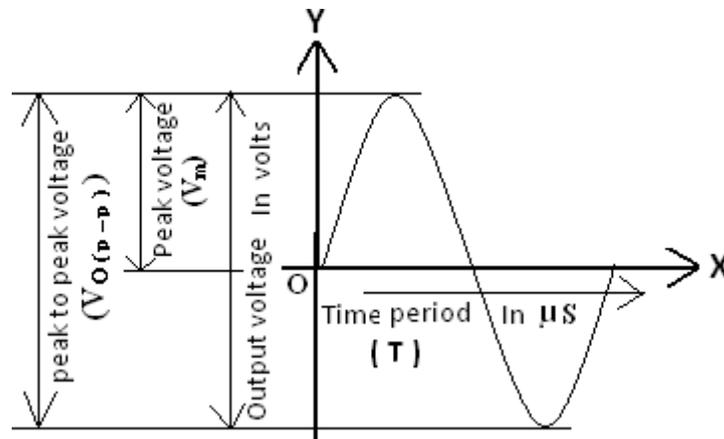
Sl. No.	$R = R_{\text{selected}} - \left[\frac{\text{Setting in \%} \times R_{\text{selected}}}{100} \right]$ In $K\Omega$	Capacitor (C) In Kpf	Theoretical frequency (f_0) $f_0 = \frac{1}{2\pi RC}$ In Hz/KHz.	Practical Time Period In $\mu\text{S}/\text{mS}$	Practical Frequency In Hz/KHz.	Output Voltage ($V_{O(p-p)}$) In Volts
1		10		0.52mS		
2		22		1mS		
3		33		1.6mS		

TABULAR FORM / CALCULATIONS – HARDWARE :

Sl.No.	Resistor (R) In $K\Omega$	Capacitor (C) In Kpf/nF	Theoretical Frequency (f_0) $f_0 = \frac{1}{2\pi RC}$ In Hz / KHz.	Practical Time Period In $\mu\text{S}/\text{mS}$	Practical Frequency In Hz./KHz.	Practical Voltage ($V_{O(p-p)}$) In Volts
1		10		0.52mS		
2		22		1mS		
3		33		1.6mS		

EXPECTED WAVEFORM :

The following waveform shows the output signal for *Hartley Oscillator*,



RESULT : I have drawn the output signal and calculated the frequency values of a given *Wein bridge Oscillator* using software & hardware.

VIVA VOICE Questions:

1. What is positive feedback Amplifier?
2. What are the conditions for oscillations?
3. What are the classifications of oscillators?
4. What are the types of RC oscillators?
5. What is the frequency of RC phase shift oscillator?
6. What is the frequency of Wien Bridge oscillator?
7. Applications of RC oscillators?
8. In RC phase shift oscillator, each RC section gives how much phase shift?
9. In Wien Bridge oscillator, feedback circuit gives how much phase shift?
10. In AF oscillators which oscillators are used?

Experiment No. : 12.A**Date :****Name of the Experiment : COLPITTS OSCILLATOR****AIM :**

To draw the *sine wave form* and to calculate its frequency values of a given *Colpitts Oscillator* using software & hardware

APPARATUS :**Software :**

- | | | |
|----------------------|-------|-------|
| 1. System | ----- | 1 No. |
| 2. Multisim software | | |

Hardware :

- | | | |
|-----------------------------------|-------|-------|
| 1. Regulated power supply (RPS) | ----- | 1 No. |
| 2. Cathode ray oscilloscope | ----- | 1 No. |
| 3. Decade Inductance Box (DIB) | ----- | 1 No. |
| 4. Decade Capacitance (DCB) | ----- | 1 No. |
| 5. Bread board | ----- | 1 No. |
| 6. Probes | ----- | 1 No. |
| 7. Connecting wires | ----- | 1 No. |

COMPONENTS :

- | | | | |
|---------------------------------------|-------|-------|------------|
| 1. Transistor | BC547 | ----- | 1 No. |
| 2. Resistors : 1KΩ, 1.5KΩ, 10KΩ, 47KΩ | | ----- | Each 1 No. |
| 3. 0.1μF, 0.01μF | | ----- | Each 1 No. |

THEORY :

The basic configuration of the **Colpitts Oscillator** resembles that of the *Hartley Oscillator* but the difference this time is that the centre tapping of the tank sub-circuit is now made at the junction of a “capacitive voltage divider” network instead of a tapped autotransformer type inductor as in the Hartley oscillator.

the resonant frequency of the LC tank circuit and is given as:

$$f_r = \frac{1}{2\pi\sqrt{L C_T}}$$

where C_T is the capacitance of C_1 and C_2 connected in series and is given as:

$$\frac{1}{C_T} = \frac{1}{C_1} + \frac{1}{C_2} \quad \text{or} \quad C_T = \frac{C_1 \times C_2}{C_1 + C_2}$$

The configuration of the transistor amplifier is of a *Common Emitter Amplifier* with the output signal 180° out of phase with regards to the input signal. The additional 180° phase shift require for oscillation is achieved by the fact that the two capacitors are connected together in series but in parallel with the inductive coil resulting in overall phase shift of the circuit being zero or 360°.

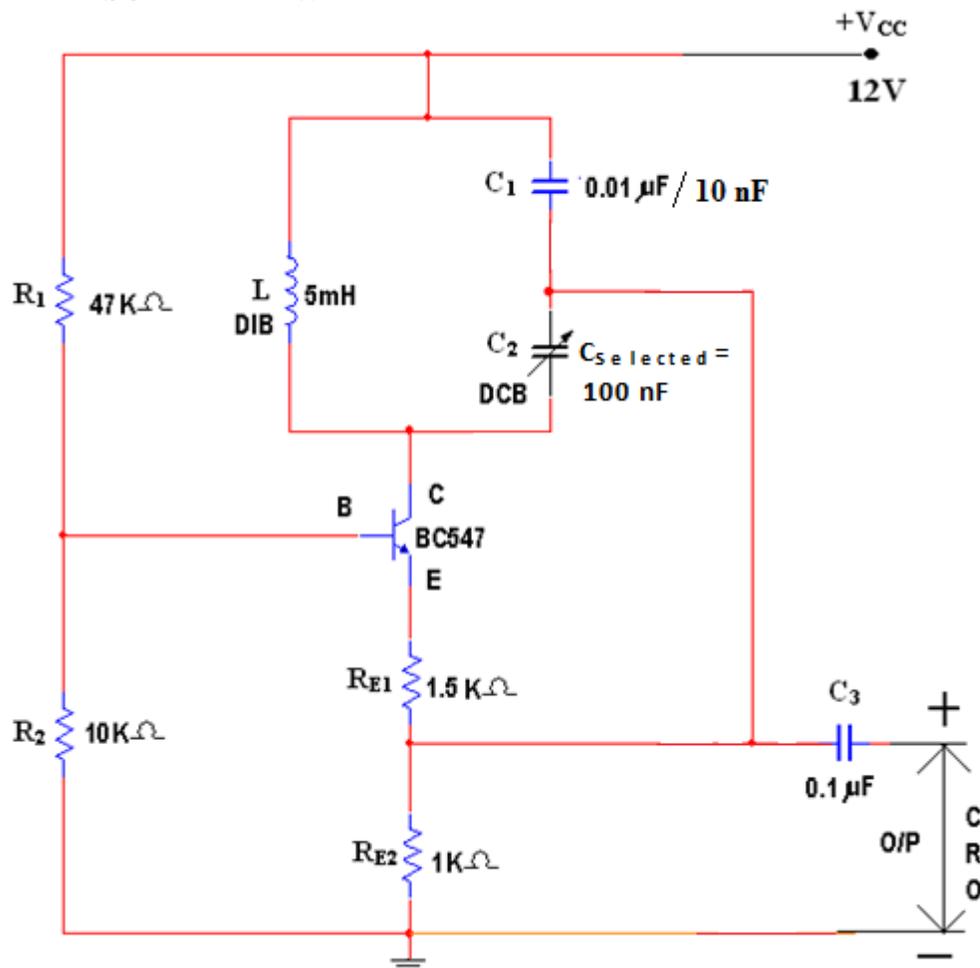
CIRCUIT DIAGRAM – SOFTWARE & HARDWARE :

Figure : Circuit diagram of Colpitt's Oscillator

PROCEDURE - SOFTWARE :

1. First calculated the theoretical frequency for all capacitor C_2 values by using the formula which is available in the tabular form.
2. Picked up the components from *components bar* in multisim software as per the circuit diagram.
3. Made the connections as per the circuit diagram.
4. Set the V_{CC} value as 12V.
5. Set the *inductance(L)* value as 5mH in DIB for all readings .
6. Initially set the *Capacitor C* values as 1nF (0.001μF/1Kpf/1nF) by varied the *capacitor C* Value in % by using the following formula,

$$\text{Setting in \%} = \frac{C_{\text{required}} \times 100}{C_{\text{selected}}}$$

7. To start the simulation clicked on *Run button* .
8. We observed *Sine wave form* as a output signal in the *CRO*.
9. Drawn the *sine wave form* on the graph by taking the *time period* on X-axis and *Amplitude ($V_{O(p-p)}$)* on Y-axis.
10. Calculated the time period and output voltage ($V_{O(p-p)}$) values from the graph then noted in the Columns of *practical time period* and *output voltage* for corresponding *capacitor C_2* values in the tabular form respectively.

11. Stopped the simulation by click on *Run option* through *Execute button*.
12. Repeated the same procedure from points 6 to 10 for corresponding C_2 values which are given below,
 - a). 2.2 nF/2.2Kpf.
 - b). 3.3 nF/3.3Kpf.
 - c). 10.0 nF/10Kpf.
13. Shut down the system safely.
14. Now calculated the practical frequency by using formula $1/T$ and noted it in corresponding columns of C_2
15. We compared that *theoretical frequency* value (f_o) and *practical frequency* value are same approximately.

PROCEDURE - HARDWARE :

1. Made the connections as per the circuit diagram.
2. Switched **ON** the *RPS* and *CRO*.
3. Set the V_{CC} value as 12V in *RPS*.
4. Set the *inductance(L)* value as 5mH in *DIB* .
5. Set the *Capacitor C₂* value as 1nF (0.001 μ F or 1Kpf) in *DCB*.6..
6. We observed *Sine wave form* as a output signal in the *CRO*.
7. Drawn the *sine wave form* on the graph by taking the *time period* on X-axis and *amplitude($V_{O(p-p)}$)* on Y-axis.
8. Calculated the frequency and output voltage ($V_{O(p-p)}$) values from the graph then noted in the columns of *practical frequency* and *output voltage* in the tabular form respectively.
9. Repeat the same procedure from points 5 to 7 for corresponding C_2 values which are given below,
 - a). 2.2 nF (0.0022 μ F or 2.2Kpf).
 - b). 3.3 nF (0.0033 μ F or 3.3Kpf).
7. Switch **OFF** the *RPS* and *CRO*.
8. Finally calculated and noted down the *theoretical frequency value (F_o)* by using the formula, $1 / (2\pi\sqrt{LC_T})$ in the tabular form.
9. I compared that *theoretical frequency* value (F_o) and *practical frequency* values are approximately same.

TABULAR FORM / CALCULATIONS – SOFTWARE :

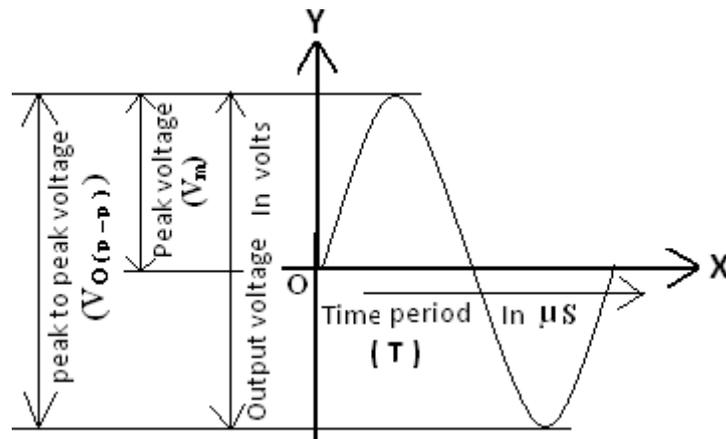
Sl No.	Capa Citor (C_1)	Capa Citor (C_2)	Inductor. (L) In mH	Total Capacitance (C_T) $= \frac{C_1 C_2}{C_1 + C_2}$ In nF	Theoretical Frequency (f_o) = $\frac{1}{2\pi\sqrt{LC_T}}$ In KHz	Practical – Time-Period. In μ S	Practical frequency In KHz.	Output voltage ($V_{O(p-p)}$) In Volts.
1.	10Kpf	1Kpf	5					
2.	10Kpf	2.2Kpf	5					
3.	10Kpf	3.3Kpf	5					

TABULAR FORM / CALCULATIONS – HARDWARE :

Sl No.	Capa Citor (C ₁)	Capa Citor (C ₂)	Indu-ctor. (L) In mH	Total Capacitance (C _T) $C_1 C_2 / (C_1 + C_2)$ In nF	Theoretical Frequency (f _o) = $\frac{1}{2\pi\sqrt{LC_T}}$ In KHz	Pract-ical – Time-Period. In μS	Pract-ical frequency In KHz.	Output voltage (V _{O p-p}) In Volts.
1.	10Kpf	1Kpf	5					
2.	10Kpf	2.2Kpf	5					
3.	10Kpf	3.3Kpf	5					

EXPECTED WAVEFORM – SOFTWARE & HARDWARE :

The following waveform shows the output signal for *Colpitts Oscillator*



RESULT : I have drawn the output signal and calculated the frequency values of a given *Colpitts Oscillator*.

VIVA VOICE Questions:

1. What are LC oscillators?
2. What is the frequency of Colpitts oscillator?
3. What is the condition for sustained oscillation in Colpitts oscillator?
4. Applications of LC oscillators?
5. In Colpitts oscillator, feedback circuit consists of how many Inductors and capacitors?
6. Which type of feedback is used for Colpitts oscillator?
7. What is Q in Colpitts oscillator?
8. What is the advantage of Colpitts oscillator?
9. How does Colpitts oscillator calculate frequency?
10. What are the advantages and disadvantages of LC oscillator?

Experiment No. : 12.B

Date :

Name of the Experiment : HARTLEY OSCILLATOR

AIM :

To draw the *sine wave form* and to calculate its frequency values of a given *Hartley Oscillator* using software and hardware

APPARATUS :

1. System ----- 1 No.
2. Multisim software

Hardware :

1. Regulated power supply (RPS)----- 1 No.
2. Cathode ray oscilloscope ----- 1 No.
3. Decade Inductance Box (DIB)----- 2 No.
4. Decade Capacitance Box (DCB) ----- 1 No.
5. Bread board ----- 1 No.
6. Probes ----- 1 No.
7. Connecting wires ----- A few Nos.

COMPONENTS :

1. Transistor BC547 ----- 1 No.
2. Resistors : 1K Ω , 10K Ω , 47K Ω ----- Each 1 No.
3. Capacitors 0.22 μ F ----- 2 No.

THEORY :

A transistor consists of inter element capacitance, i.e., the collector to emitter capacitor. If the value of this capacitor changes the oscillations frequency is also changing, hence the stability of the oscillator. This effect can be neutralized by placing swamping capacitor across the offending elements.

The Hartley oscillator is an electronic oscillator circuit in which the oscillation frequency is determined by a tuned circuit consisting of capacitors and inductors, that is, an LC oscillator. The circuit was invented in 1915 by American engineer Ralph Hartley.

The Hartley oscillator is used as a local oscillator in radio receivers. Due to the reason for a wide range of frequencies, it is a popular oscillator. This oscillator is suitable for oscillations in Radio Frequency (RF) range up to 30MHz.

The feedback used in Hartley oscillator is Voltage series feedback. It cannot be used as a low-frequency oscillator since the value of inductors becomes large and the size of the inductors becomes large. The harmonic content in the output of this oscillator is very high and hence it is not suitable for the applications which require a pure sine wave.

In Hartley oscillator frequency based on the formula, **frequency= $\frac{1}{2\pi\sqrt{LtC}}$** , where C is the value of the capacitor and LT is the equivalent inductance of the inductors in series. The equivalent inductance in series is equal to the sum of both inductors together.

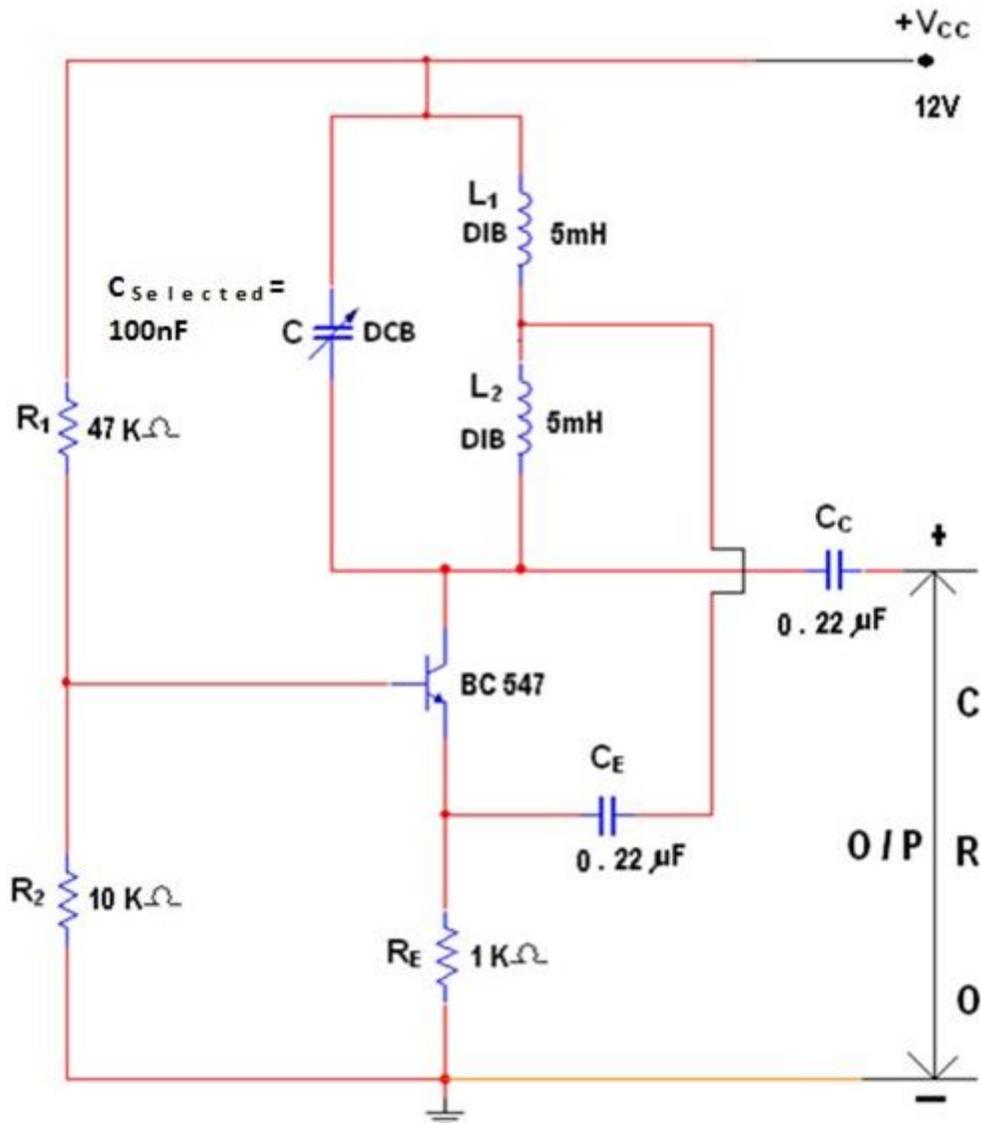
CIRCUIT DIAGRAM :

Figure : Circuit diagram of Hartley Oscillator

PROCEDURE - SOFTWARE :

1. First calculated the theoretical frequency for all capacitor C values by using the formula which is available in the tabular form.
2. Picked up the components from *components bar* in multisim software as per the circuit diagram.
3. Made the connections as per the circuit diagram.
4. Set the V_{CC} value as 12V.
5. Set the inductance L_1 & L_2 values as 5mH in both DIB's for all readings .
6. Initially set the Capacitor C values as 10nF (0.01μF/10Kpf/10nF) by varied the capacitor C Value in % by using the following formula,

$$\text{Setting in \%} = \frac{C_{\text{required}} \times 100}{C_{\text{selected}}}$$

7. To start the simulation clicked on *Run button* .
8. We observed *Sine wave form* as a output signal in the *CRO*.
9. Drawn the *sine wave form* on the graph by taking the *time period* on X-axis and *Amplitude* ($V_{O(p-p)}$) on Y-axis.

10. Calculated the time period and output voltage ($V_{O(p-p)}$) values from the graph then noted in the Columns of *practical time period* and *output voltage* for corresponding *capacitor C* values in the tabular form respectively.
11. Stopped the simulation by click on *Run option* through *Execute button*.
12. Repeated the same procedure from points 6 to 10 for corresponding C_2 values which are given below, a). 33nF/33Kpf. b). 47 nF/47Kpf.
13. Shut down the system safely.
14. Now calculated the practical frequency by using formula $1/T$ and noted it in corresponding columns of C
15. We compared that *theoretical frequency* value (f_o) and *practical frequency* value are same approximately.

PROCEDURE - HARDWARE :

1. Calculated and noted the values of theoretical *frequency* (f_o) values to the corresponding *capacitor C* values in the tabular form by using the formula given below,
2. Made the connections as per the circuit diagram.
3. Switched **ON** the *RPS* and *CRO*.
4. Kept the V_{CC} value as 12V in *RPS*.
5. Kept the *Capacitor C* value as 10nF (0.01 μ F or 10Kpf) in DCB and 5mH in both Inductors and maintained the 5mH value as a constant in both DIB's up to the experiment is completed..
6. I observed *Sine wave form* as a output signal in the *CRO*.
7. Drawn the *sine wave form* on the graph by taking the *time period* on X-axis and *amplitude* ($V_{O(p-p)}$) on Y-axis.
8. Calculated the frequency and output voltage ($V_{O(p-p)}$) values from the graph then noted in the columns of *practical frequency* and *output voltage* to the corresponding values in the tabular form respectively.
9. Repeated the same procedure from points 6 to 8 for corresponding C value which are given below,
 - a). 33nF (0.01 μ F or 33Kpf).
 - b). 47nF (0.047 μ F or 47Kpf).
10. Switch **OFF** the *RPS* and *CRO*.
11. I compared that *theoretical frequency* value (F_o) and *practical frequency* values are approximately same.

TABULAR FORM / CALCULATIONS – SOFTWARE :

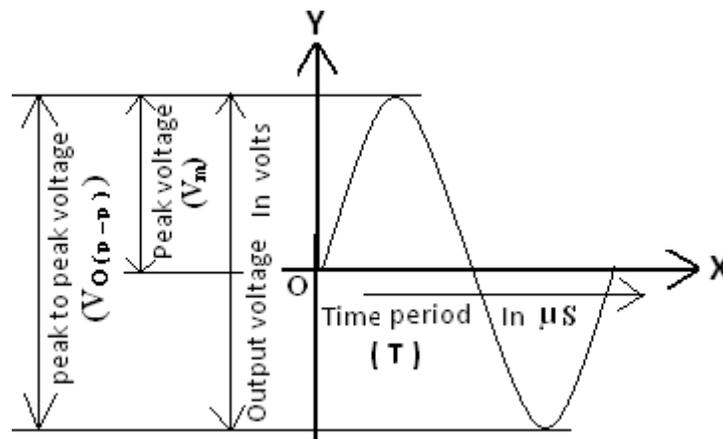
Sl No.	Capa Citor (C) In nF/Kpf	Indu ctnce (L ₁) In mH	Indu- tance. (L ₂) In mH	Total Induc tance (L _T) = L ₁ +L ₂ In mH	Theoretical Frequency (f ₀)= $\frac{1}{2\pi\sqrt{L_T C}}$ In KHz.	Practi cal Time period (T) In μS	Pract- ical- frequency (f) In KHz.	Output voltage (V _{O p-p}) In Volts.
1.	10	5	5					
2.	33	5	5					
3.	47	5	5					

TABULAR FORM / CALCULATIONS - HARDWARE :

Sl No.	Capa Citor (C) In nF/Kpf	Indu ctnce (L ₁) In mH	Indu- tance. (L ₂) In mH	Total Induc tance (L _T) = L ₁ +L ₂ In mH	Theoretical Frequency (f ₀)= $\frac{1}{2\pi\sqrt{L_T C}}$ In KHz.	Practi cal Time period (T) In μS	Pract- ical- frequency (f) In KHz.	Output voltage (V _{O p-p}) In Volts.
1.	10	5	5					
2.	33	5	5					
3.	47	5	5					

EXPECTED WAVEFORM – SOFTWARE & HARDWARE :

The following waveform shows the output signal for *Hartley Oscillator*,



RESULT : I have drawn the output signal and calculated the frequency values of a given *Hartley Oscillator*.

VIVA VOICE Questions:

1. What is positive feedback Amplifier?
2. What are the conditions for oscillations?
3. What are the classifications of oscillators?
4. What are the types of LC oscillators?
5. What is the frequency of Hartley oscillator?
6. Applications of LC oscillators?
7. In Hartley oscillator, feedback circuit consists of how many Inductors and capacitors?
8. In RF oscillators which oscillators are used?
9. Why Hartley oscillator is used?
10. Why Hartley oscillator is used?

Experiment No. : 13**Date :****Name of the Experiment : BOOTSTRAPPED EMITTER FOLLOWER
(Beyond the Syllabus-Using Software & Hardware)****AIM :**

To obtain the frequency response of *Bootstrapped Emitter Follower* using software and hardware

APPARATUS :

1. System ----- 1 No.
2. Multisim software

APPARATUS :

1. Regulated power supply (RPS) ----- 1 No.
2. Cathode Ray Oscilloscope (CRO) ----- 1 No.
3. Function generator ----- 1 No.
4. Probes ----- 1 No.
5. Bread board ----- 1 No.
6. Connecting wires ----- A few Nos.

COMPONENTS :

1. Transistor BC 547 ----- 2 No.
2. Capacitors :
 - ii). 10 μF ----- 2 No.
 - i). 22 μF ----- 1 No.
3. Resistors :
 - i). 100 $\text{K}\Omega$ ----- 2 No.
 - ii). 10 $\text{K}\Omega$ ----- 2 No.
 - iii). 100 Ω ----- 2 No.

THEORY :

Typically Bootstrapping is technique where some part of output is used at the startup. In Bootstrap amplifier, bootstrapping is used to increase the input impedance. Due to which the loading effect on the input source also decreases. The design looks similar to the Darlington pair, having a bootstrap capacitor. Bootstrap capacitor is used to provide AC signal's positive feedback to the base of the transistor. This positive feedback help in improving the effective value of the base resistance. This increment in the base resistance also determined by the voltage gain of the amplifier circuit.

High input impedance improves the amplification of the input signal and thus required in various amplifier applications. If we have low input impedance we will get low amplification. Generally, BJT (Bipolar Junction Transistor) have low input impedance (typically 1 ohm to 50 kilo ohm). So for this, bootstrapping technique is used to increase the input impedance.

The voltage across the input impedance is calculated by using the below formula:

$$V = \{(V_{in} \cdot Z_{in}) / (V_{in} + Z \cdot V_{in})\}$$

Hence, according to the formula, the input impedance is proportional to the voltage across it. If the input impedance is increased the voltage across it will also increase and vice versa.

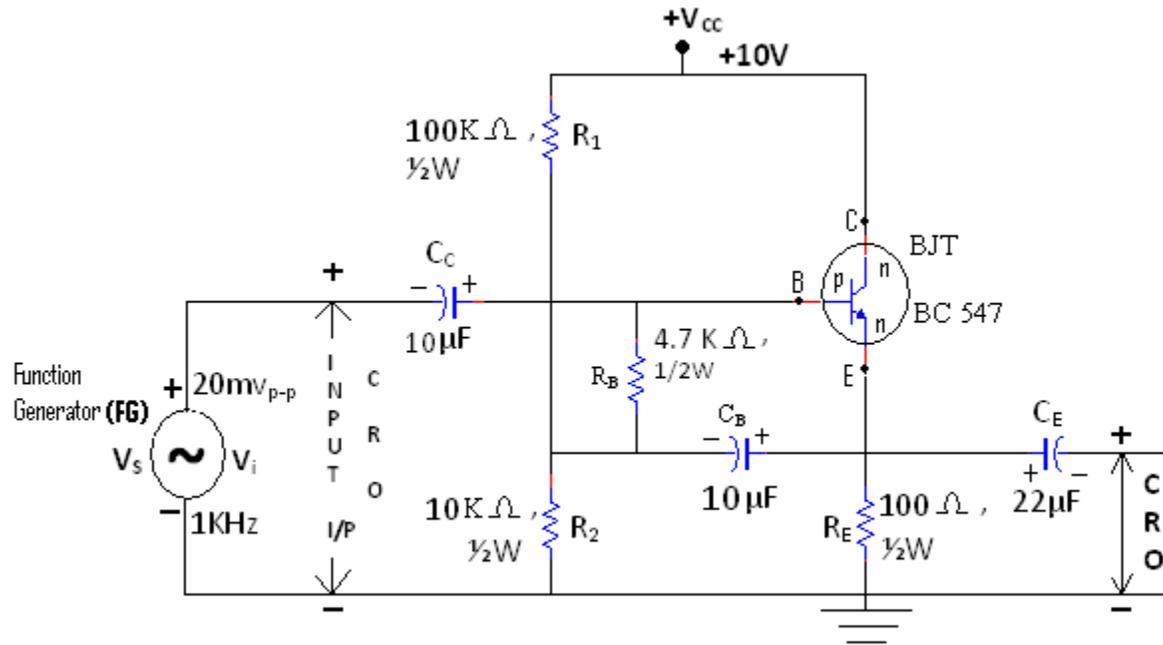
CIRCUIT DIAGRAM – SOFTWARE & HARDWARE :

Figure: Circuit diagram of Bootstrapped Emitter Follower

PROCEDURE – SOFTWARE :

1. We have picked up the components from the components bar as per above circuit.
2. Made the connections as per the above circuit diagram by using the components which we have picked up.
3. Connected the CRO across the Emitter capacitor to ground.
4. Set the input signal as *sine wave form* which is having the value $20\text{mV}_{\text{P-P}}$ as constant in the function generator.
5. Initially set the input signal frequency value is 1KHz in the function generator.
6. To simulate the circuit clicked on *run option* through *execute button* in *tool bar*.
7. We have seen the *sine wave* on the **CRO** screen as o/p signal.
8. Calculated the *peak to peak voltage* ($V_{O(p-p)}$) and noted down in the tabular form Against the column of 1KHz.
9. Stopped the simulation by clicked on *run option* through *execute button* in the *tool bar*.
10. Repeat the same procedure from points 6 to 9 for the corresponding frequency values by setting in the function generator for the steps of 10Hz, 500Hz, 1KHz, 100KHz, 200KHz, 400KHz, 600KHz, 1180KHz, 1MHz, 100MHz, 500MHz. in the function generator.
11. Observed the graph for *frequency Vs amplitude* through the *AC Analysis*.
12. Finally shut down the system safely.
13. We have observed that, the graph which is drawn by manually is same to the graph which is obtained from the *AC Analysis*.
14. Now calculated and noted down the values of *voltage gain* (A_v) and *gain in dB* to the corresponding values of *output voltage* (V_o) & *input voltage* (V_i) by using the formulas given below,

$$\text{Voltage gain } (A_v) = V_o / V_i \text{ and } \text{Gain in dB} = 20 \log_{10}(A_v).$$
15. Plotted the graphs (frequency response curves) as per below
 - a). frequency on X-axis & gain in dB on Y-axis.
 - b). frequency on X-axis & voltage gain on Y-axis.

PROCEDURE- HARDWARE :

1. We have connected the circuit as per the circuit diagram which is shown above.
2. Initially connected the probe across the function generator as per shown in the circuit diagram to set the input signal.
3. Switched *ON* the *CRO* and *function generator*.
4. Applied the input signal as *sine wave form* having the values of $20\text{m}_{\text{p-p}}$, 1KHz. from the function generator by observing in the CRO.
5. Later removed the probe from that place and connected it across the capacitor C_{C2} to observe the output.
6. Switched *ON* the *RPS* and kept the 10V as V_{CC} .
7. Kept the amplitude of the input signal as constant as $20\text{mV}_{\text{p-p}}$ for all frequency steps.
8. Noted down the values output voltage of output signal in terms of peak to peak voltages by varying the different frequency steps in the function generator which are given below,
10Hz, 500Hz, 1KHz, 100KHz, 200KHz, 400KHz, 600KHz, 1180KHz, 1MHz.
9. Repeat the same procedure for point 8 for corresponding frequency values.
10. Now calculated and noted down the values of *voltage gain*(A_V) and *gain in dB* to the corresponding values of *output voltage*(V_O) & *input voltage*(V_i) by using the formulas given below,
$$\text{Voltage gain } (A_V) = V_o / V_i \text{ and } \text{Gain in dB} = 20\log_{10}(A_V).$$
11. Plotted the graph between *frequency on X- axis* and *gain in dB on Y- axis*.

Note: Bootstrap Emitter Follower uses to increase the input impedance and to work as correct *Buffer*.

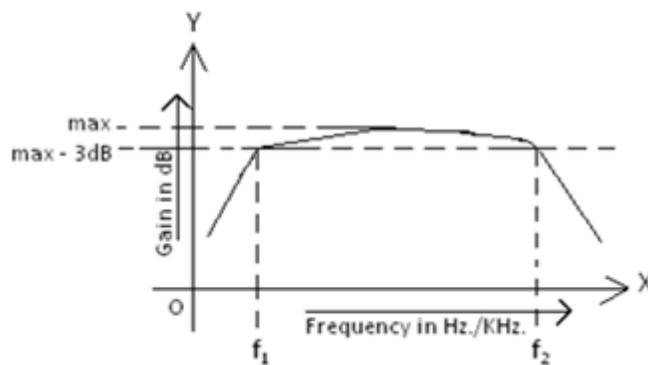
For example, The voltage gain of *voltage series feedback amplifier* is 1 it means the output voltage is equal to input voltage, then we can say that it is the correct *Buffer*. Now If you observed the output of *voltage series feedback amplifier* the output voltage is less as compared to input voltage, it means buffer is incorrect. To increase the output voltage which is equal to the input voltage here we have used the *Bootstrapped Emitter Follower*.

TABULAR FORM – SOFTWARE & HARDWARE :

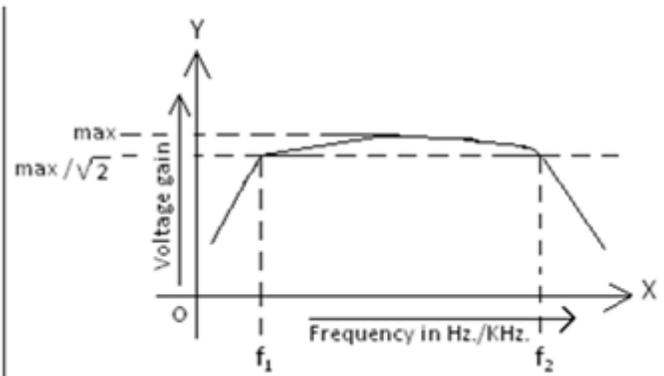
Input Voltage in Function generator $V_i = 20 \text{ mV}$ or 0.02V for all readings								
Software					Hardware			
Sl. No	Frequency In Hz/KHz.	Output Voltage (V_O) In mVolts.	Voltage gain $A_V = V_o/V_i$	Gain in dB = $20\log_{10}(A_V)$	Output Voltage (V_O) In mVolts.	Voltage gain $A_V = V_o/V_i$	Gain in dB = $20\log_{10}(A_V)$	
01	10 Hz.							
02	500 Hz.							
03	1 KHz.							
04	100 KHz.							
05	200 KHz.							
06	400 KHz.							
07	600 KHz.							
08	1180 KHz.							
09	1 MHz.							
10	100 MHz							
11	500 MHz.							

EXPECTED GRAPH – SOFTWARE & HARDWARE :

A). Frequency response curve for Bootstrapped Emitter Follower for frequency verses gain dB.



B). Frequency response curve for Bootstrapped Emitter Follower For frequency verses voltage gain.



Note : It is not possible to find the band width because there is no amplification in this amplifier. It is just working as *buffer*.

RESULT :

I have obtained the gain of *Bootstrapped Emitter Follower* for different frequencies .

Experiment No. : 14**Date :****Name of the Experiment :****ASTABLE MULTIVIBRATOR****(Beyond the Syllabus-Using Software & Hardware)****AIM :**

To conduct and verify the Astable multi vibrator and to draw the waveforms using software and hardware

APPARATUS :

- | | | |
|-----------------------------------|-------|-------|
| 1. System with Multisim software | ----- | 1 No. |
| 2. Regulated power supply (RPS) | ----- | 1 No. |
| 3. Cathode ray oscilloscope | ----- | 1 No. |
| 4. Function Generator | ----- | 1 No. |
| 5. Bread board | ----- | 1 No. |
| 6. Probes | ----- | 1 No. |
| 7. Connecting wires | ----- | 1 No. |

COMPONENTS :

- | | | |
|-------------------------------------|-------|-------|
| 1. Resistors : 1K Ω | ----- | 2 No. |
| 10 K Ω | ----- | 2 No. |
| 100 K Ω | ----- | 2 No. |
| 2. Capacitors : 0.1 μ F / 100nF | ----- | 2 No. |
| 3. Transistor : BC547 | ----- | 2 No. |

THEORY :

Astable Multivibrator is a two stage switching circuit in which the output of the first stage is fed to the input of the second stage and vice versa. The outputs of both the stages are complementary. This free running multivibrator generates square wave without any external triggering pulse.

It is also called free-running relaxation oscillator. It has no stable state but only two quasi-stable states between which it keeps oscillating continuously of its own accord without any external excitation. When one transistor is in ON state and other remains in OFF state.

As a timing oscillator or clock of a computer system. It is also used for a flashing lights, switching and power supply circuits.

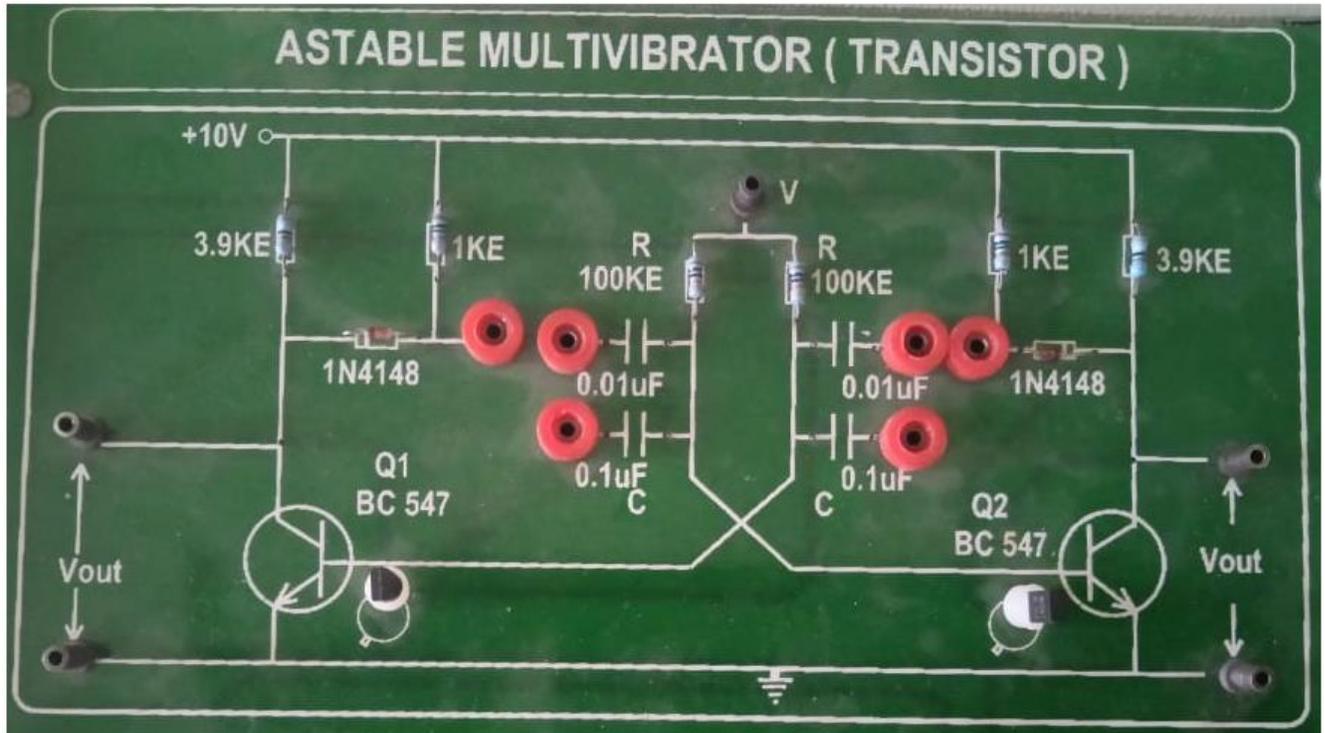
Advantages :

1. They work consistently and are not influenced by any outside forces or events.
2. They are inexpensive.
3. They are simple in design.
4. They can remain functional for an extraordinary length of time.

Disadvantages :

They do not transfer the entire output signal to the input due to several reasons like:

1. There is resistance within the circuit.
2. Absence of a completely closed loop at the output terminals.
3. One capacitor or transistor has a tendency to absorb energy at a slightly different rate than the other.
4. Even though the amplifier restores the lost energy when it amplifies the signal, the signal is too small.

CIRCUIT DIAGRAMS – SOFTWARE & HARDWARE:**Design Procedure :**

The period T is given by

$$T = T_1 + T_2 = 0.69 (R_1 C_1 + R_2 C_2)$$

For symmetrical circuit,

$$\text{with } R_1 = R_2 = R \text{ \& } C_1 = C_2 = C$$

$$T = 1.38 RC$$

Let $V_{CC} = 12V$; $h_{fe} = 51$ (for BC107), $V_{BESat} = 0.7V$; $V_{CESat} = 0.3V$ Let $C = 0.1\mu F$ & $T = 1mSec$.
 $10^{-3} = 1.38 \times R \times 0.1 \times 10^{-6}$

$R = 7.24K\Omega$ (Practically choose $10K\Omega$) i.e., R_1 and R_2 resistors = $10K\Omega$

Let $I_{Cmax} = 10mA$

$$R_C = \frac{V_{CC} - V_{cesat}}{I_{cmax}} = \frac{12 - 0.3}{0.01} = 1.17K\Omega \text{ (Practically choose } 1K\Omega \text{) i.e., } R_{c1} \text{ and } R_{c2} \text{ resistors} = 1K\Omega$$

Theoretical calculations :

$$F = 1/T = (1/1.38RC)$$

$$R = 10K\Omega \quad C = 0.1\mu F$$

PROCEDURE - SOFTWARE :

1. I have picked up the components from the components bar as per above circuit.
2. Made the connections as per the above circuit diagram by using the components which we have picked up.
3. Connected the CRO across VC1 and VC2 .
6. To simulate the circuit clicked on *run option* through *execute button* in *tool bar*.
7. I have observed the wave forms as shown under the heading of Expected graphs .
8. Observed the Base Voltage and Collector Voltages of Q1 & Q2 on CRO in DC mode and measured the frequency ($f = 1/T$).
9. Traced the waveforms at collector and base as each transistor with the help of dual trace CRO and plot the waveforms.
10. Verified the practical output frequency with theoretical values $f = 1/T$, where $T = 1.38RC$
11. Shut down the system safely.
12. Plotted the graphs for VB1 & VC1 and VB2 and VC2 by taking the Time period on X-axis and Voltage on Y-axis for all graphs as per shown in the Expected graphs heading.
13. Noted the practical Time period T values at VB1 & VC1 and VB2 and VC2 and noted down in the corresponding columns of the Tabular form.
14. Calculated the practical frequency values by using formula $1/T$ and noted down in the corresponding columns of the Tabular form.
15. I Compared the Theoretical and practical values are approximately same.

PROCEDURE - HARDWARE :

1. I have made the connections as per the circuit diagram.
2. Observed the Base Voltage and Collector Voltages of Q1 & Q2 on CRO in DC mode and measured the frequency ($f = 1/T$).
3. Traced the waveforms at collector and base as each transistor with the help of dual trace CRO and plot the waveforms.
4. Verified the practical output frequency with theoretical values $f = 1/T$, where $T = 1.38RC$
5. Switched off the RPS and CRO.
6. Plotted the graphs for VB1 & VC1 and VB2 and VC2 by taking the Time period on X-axis and Voltage on Y-axis for all graphs as per shown in the Expected graphs heading.
7. Noted the practical Time period T values at VB1 & VC1 and VB2 and VC2 and noted down in the corresponding columns of the Tabular form.
8. Calculated the practical frequency values by using formula $1/T$ and noted down in the corresponding columns of the Tabular form.
9. I Compared the Theoretical and practical values are approximately same.

TABULAR FORMS – SOFTWARE & HARDWARE :

	Software			Hardware	
	At VC1	At VC2		At VC1	At VC2
Theoretical Time period (T)					
Theoretical Frequency (f) = 1/T					
Practical Time period (T)					
Practical Frequency (f) = 1/T					

EXPECTED WAVEFORM SOFTWARE & HARDWARE :

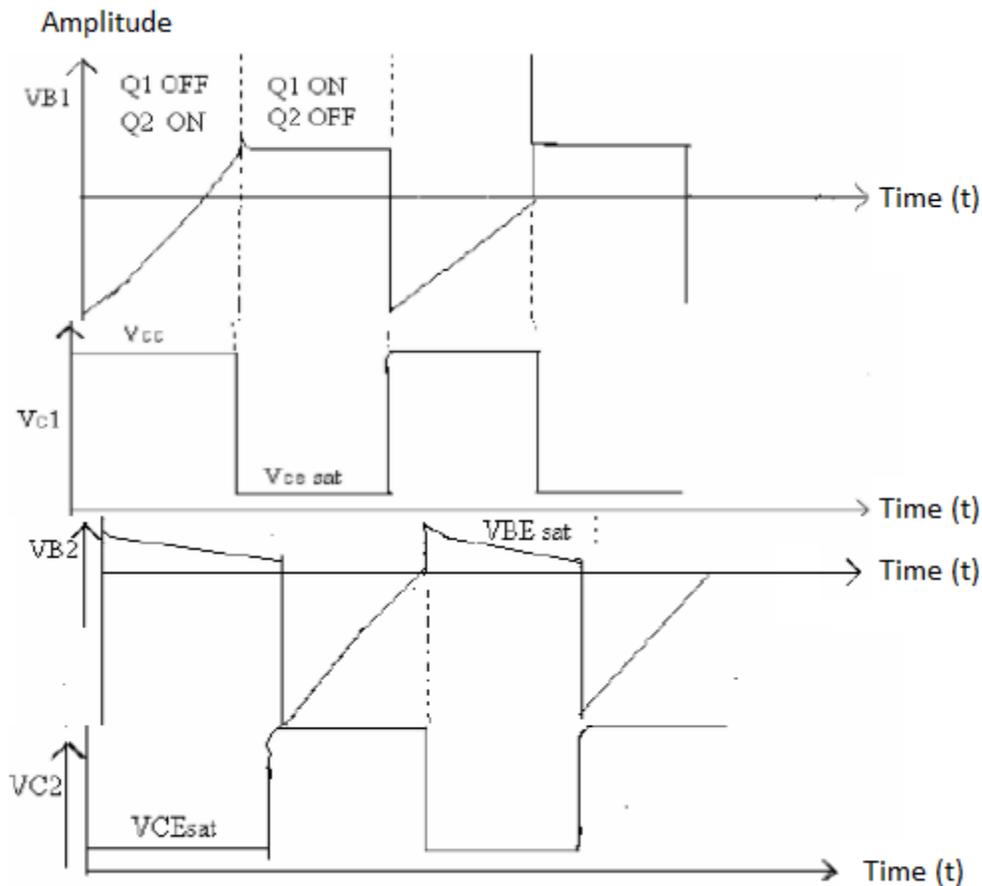


Fig : Wafeforms of Astable Multivibrator

RESULT : I have conducted and verified the Astable Multivibrator.

VIVA VOICE Questions:

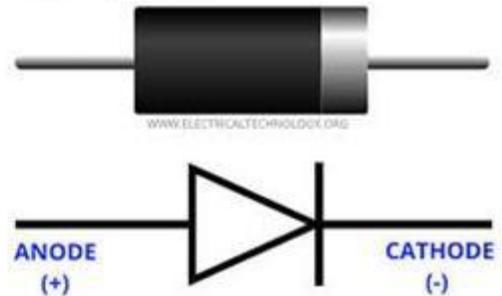
1. What is Multi-vibrator?
2. What are the types of Multi-vibrators?
3. What is Astable Multi-vibrator?
4. Mention the Applications of Astable Multi-vibrator.
5. Astable Multi-vibrator is having how many stable states?
6. Which one is Square wave oscillator?
7. Which of the multi-vibrator used in Relaxation oscillators?
11. Compare Mono-stable, Bi-stable and Astable multi-vibrators.
12. What is Quasi stable?
13. Free running Multi vibrator generates Square wave. (True or False)

A. DATA SHEETS**PN JUNCTION DIODE :****1N4001 - 1N4007 1.0A****Features**

- Diffused Junction
- High Current Capability and Low Forward Voltage Drop
- Surge Overload Rating to 30A Peak
- Low Reverse Leakage Current
- Lead Free Finish, RoHS Compliant (Note 3)

Mechanical Data

- Case: DO-41
- Case Material: Molded Plastic. UL Flammability Classification Rating 94V-0
- Moisture Sensitivity: Level 1 per J-STD-020D
- Terminals: Finish - Bright Tin. Plated Leads
Solderable per MIL-STD-202, Method 208
- Polarity: Cathode Band
- Ordering Information: See Page 2
- Marking: Type Number
- Weight: 0.30 grams (Approximate)

**Maximum Ratings and Electrical Characteristics** (@ $T_A = +25^\circ\text{C}$ unless otherwise specified.) Single phase, half wave,

For capacitive load, derate current by 20%.

Characteristic	Symbol	1N4001	1N4002	1N4003	1N4004	1N4005	1N4006	1N4007	Unit
Peak Repetitive Reverse Voltage Working Peak Reverse Voltage DC Blocking Voltage	V_{RRM} V_{RW} $M V_R$	50	100	200	400	1180	1180	1000	V
RMS Reverse Voltage	$V_{R(RMS)}$	35	70	140	2118	420	5118	700	V
Average Rectified Output Current (Note 1) @ $T_A = +75^\circ\text{C}$	I_O	1.0							A
Non-Repetitive Peak Forward Surge Current 8.3ms Single Half Sine-Wave Superimposed on Rated Load	I_{FSM}	30							A
Forward Voltage @ $I_F = 1.0\text{A}$	V_{FM}	1.0							V
Peak Reverse Current @ $T_A = +25^\circ\text{C}$ at Rated DC Blocking Voltage @ $T_A = +100^\circ\text{C}$	I_{RM}	5.0 50							μA
Typical Junction Capacitance (Note 2)	C_j	15					8		pF
Typical Thermal Resistance Junction to Ambient	$R_{\theta JA}$	100							K/W
Maximum DC Blocking Voltage Temperature	T_A	+150							$^\circ\text{C}$
Operating and Storage Temperature Range	T_J, T_{STG}	-65 to +150							$^\circ\text{C}$

ZENER DIODE :

TOSHIBA

1Z6.2~1Z390,1Z6.8A~1Z30A

TOSHIBA ZENER DIODE SILICON DIFFUSED JUNCTION TYPE

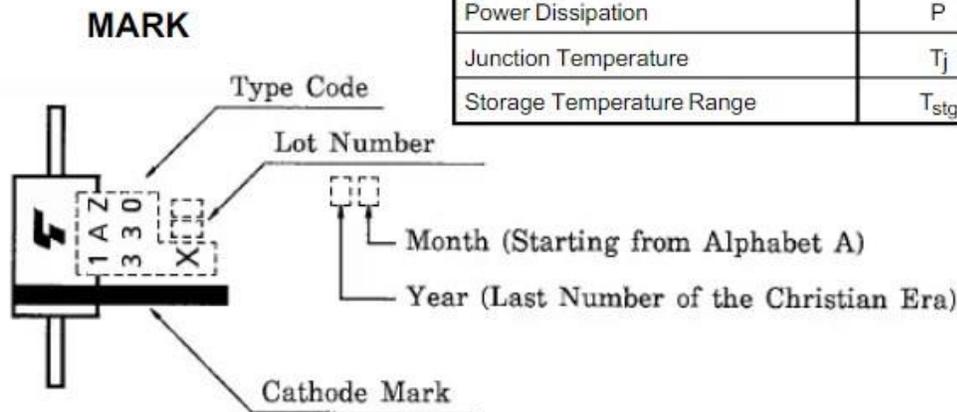
1Z6.2~1Z390,1Z6.8A~1Z30A

CONSTANT VOLTAGE REGULATION
TRANSIENT SUPPRESSORS

- Average Power Dissipation : P = 1W
- Peak Reverse Power Dissipation : PRSM = 200W at $t_w = 200\mu s$
- Zener Voltage : VZ = 6.2 ~ 390V
- Tolerance of Zener Voltage
 1Z6.2 Series : ±10%
 1Z6.8A Series : ±5%
- Plastic Mold Package

MAXIMUM RATINGS (Ta=25°C)

CHARACTERISTIC	SYMBOL	RATING	UNIT
Power Dissipation	P	1	W
Junction Temperature	T _j	-40~150	°C
Storage Temperature Range	T _{stg}	-40~150	°C



Color : Silver

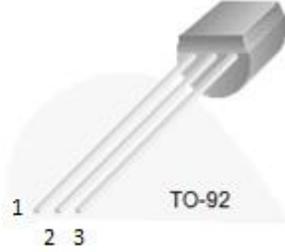
BJT & UJT :**BIPOLAR JUNCTION TRANSISTORS (BJT) :**



BC546 / BC547 / BC548 / BC549 / BC550 NPN Epitaxial Silicon Transistor

Features

- Switching and Amplifier
- High-Voltage: BC546, $V_{CE0} = 65\text{ V}$
- Low-Noise: BC549, BC550
- Complement to BC556, BC557, BC558, BC559, and BC560



1. Collector 2. Base 3. Emitter

Absolute Maximum Ratings

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only. Values are at $T_A = 25^\circ\text{C}$ unless otherwise noted.

Symbol	Parameter	Value	Unit
V_{CBO}	Collector-Base Voltage	BC546	80
		BC547 / BC550	50
		BC548 / BC549	30
V_{CEO}	Collector-Emitter Voltage	BC546	65
		BC547 / BC550	45
		BC548 / BC549	30
V_{EBO}	Emitter-Base Voltage	BC546 / BC547	6
		BC548 / BC549 / BC550	5
I_C	Collector Current (DC)	100	mA
P_C	Collector Power Dissipation	500	mW
T_J	Junction Temperature	150	$^\circ\text{C}$
T_{STG}	Storage Temperature Range	-65 to +150	$^\circ\text{C}$

Electrical Characteristics

Values are at $T_A = 25^\circ\text{C}$ unless otherwise noted.

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Unit		
I_{CBO}	Collector Cut-Off Current	$V_{CB} = 30\text{ V}, I_E = 0$			15	nA		
h_{FE}	DC Current Gain	$V_{CE} = 5\text{ V}, I_C = 2\text{ mA}$	110		800			
$V_{CE(sat)}$	Collector-Emitter Saturation Voltage	$I_C = 10\text{ mA}, I_B = 0.5\text{ mA}$		90	250	mV		
		$I_C = 100\text{ mA}, I_B = 5\text{ mA}$		250	600			
$V_{BE(sat)}$	Base-Emitter Saturation Voltage	$I_C = 10\text{ mA}, I_B = 0.5\text{ mA}$		700		mV		
		$I_C = 100\text{ mA}, I_B = 5\text{ mA}$		900				
$V_{BE(on)}$	Base-Emitter On Voltage	$V_{CE} = 5\text{ V}, I_C = 2\text{ mA}$	580	660	700	mV		
		$V_{CE} = 5\text{ V}, I_C = 10\text{ mA}$			720			
f_T	Current Gain Bandwidth Product	$V_{CE} = 5\text{ V}, I_C = 10\text{ mA}, f = 100\text{ MHz}$		300		MHz		
C_{ob}	Output Capacitance	$V_{CB} = 10\text{ V}, I_E = 0, f = 1\text{ MHz}$		3.5	6.0	pF		
C_{ib}	Input Capacitance	$V_{EB} = 0.5\text{ V}, I_C = 0, f = 1\text{ MHz}$		9		pF		
NF	Noise Figure	BC546 / BC547 / BC548	$V_{CE} = 5\text{ V}, I_C = 200\text{ }\mu\text{A}, f = 1\text{ kHz}, R_G = 2\text{ k}\Omega$		2.0	10.0	dB	
		BC549 / BC550			1.2	4.0		
		BC549		$V_{CE} = 5\text{ V}, I_C = 200\text{ }\mu\text{A}, R_G = 2\text{ k}\Omega, f = 30\text{ to }15000\text{ MHz}$		1.4		4.0
		BC550				1.4		3.0

h_{FE} Classification

Classification	A	B	C
h_{FE}	110 ~ 220	200 ~ 450	420 ~ 800

UNIUNCTION TRANSISTOR (UJT) :

2N2646
2N2647

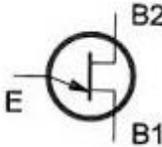
**SILICON
PN UNIUNCTION TRANSISTORS**



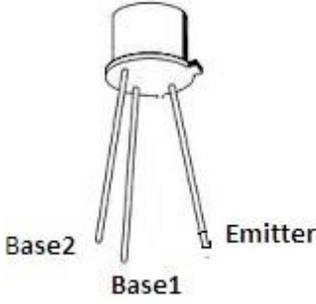
www.centrasemi.com

DESCRIPTION:
The CENTRAL SEMICONDUCTOR 2N2646 and 2N2647 devices are silicon PN Unijunction Transistors designed for general purpose industrial applications.

UJT Symbol & Terminal Identification



(a). Symbol



(b). Terminal Identification

MAXIMUM RATINGS: (T _A =25°C)		SYMBOL		UNITS
Emitter Reverse Voltage		V _{B2E}	30	V
Interbase Voltage		V _{B2B1}	35	V
RMS Emitter Current		I _e	50	mA
Peak Emitter Current (Duty Cycle ≤1%, PRR≤10pps)		I _e	2.0	A
RMS Power Dissipation		P _D	300	mW
Operating and Storage Junction Temperature		T _J , T _{stg}	-65 to +150	°C

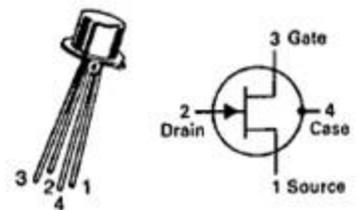
ELECTRICAL CHARACTERISTICS: (T _A =25°C unless otherwise noted)						
SYMBOL	TEST CONDITIONS	2N2646		2N2647		UNITS
		MIN	MAX	MIN	MAX	
η	V _{B2B1} =10V	0.56	0.75	0.68	0.82	
R _{BB}	V _{B2B1} =3.0V	4.7	9.1	4.7	9.1	kΩ
I _{EB2O}	V _{B2E} =30V	-	12	-	0.2	μA
I _V	V _{B2B1} =20V, R _{B2} =100Ω	4.0	-	8.0	18	mA
I _P	V _{B2B1} =25V	-	5.0	-	2.0	μA
V _{OB1}	V ₁ =20V	3.0	-	6.0	-	V

FIELD EFFECT TRANSISTOR (FET) :

MOTOROLA SC-{XSTRS/R F}

BFW10 BFW11

CASE 20-03, STYLE 1
TO-72 (TO-206A)



JFET
VHF/UHF AMPLIFIER
N-CHANNEL – DEPLETION

MAXIMUM RATINGS

Rating	Symbol	Value	Unit
Drain-Source Voltage	V _{DS}	30	V _{dc}
Drain-Gate Voltage	V _{DG}	30	V _{dc}
Reverse Gate-Source Voltage	V _{GSR}	-30	V _{dc}
Forward Gate Current	I _{GF}	10	mA _{dc}
Total Device Dissipation @ T _A = 25°C Derate above 25°C	P _D	300 1.71	mW mW/°C
Operating and Storage Junction Temperature Range	T _J , T _{stg}	-65 to +150	°C

ELECTRICAL CHARACTERISTICS (T_A = 25°C unless otherwise noted.)

Characteristic	Symbol	Min	Typ	Max	Unit
OFF CHARACTERISTICS					
Gate-Source Breakdown Voltage (I _G = 10 μA _{dc} , V _{DS} = 0)	V _{(BR)GSS}	30	—	—	V _{dc}
Gate-Source Cutoff Voltage (V _{DS} = 15 V _{dc} , I _D = 0.5 nA _{dc})	V _{GS(off)}	—	—	8 6	V _{dc}
Gate Reverse Current (V _{GS} = 20 V _{dc} , V _{DS} = 0)	I _{GSS}	—	—	0.1	nA _{dc}
Gate-Source Voltage (V _{DS} = 15 V _{dc} , I _D = 400 μA _{dc})	V _{GS}	2	—	7.5	V _{dc}
Gate-Source Voltage (V _{DS} = 15 V _{dc} , I _D = 50 μA _{dc})	V _{GS}	1.25	—	4	V _{dc}
ON CHARACTERISTICS					
Zero-Gate Voltage Drain Current (V _{DS} = 15 V _{dc} , V _{GS} = 0)	I _{DSS}	8 4	—	20 10	mA _{dc}
SMALL-SIGNAL CHARACTERISTICS					
Forward Transadmittance (V _{DS} = 15 V _{dc} , V _{GS} = 0, f = 1 kHz)	Y _{fs}	3.5 3.0	—	6.5 6.5	mmhos
Output Admittance (V _{DS} = 15 V _{dc} , V _{GS} = 0, f = 1.0 kHz)	Y _{os}	—	—	85 50	μmhos
Input Capacitance (V _{DS} = 15 V _{dc} , V _{GS} = 0 V _{dc} , f = 1.0 MHz)	C _{iss}	—	—	5.0	pF
Reverse Transfer Capacitance (V _{DS} = 15 V _{dc} , V _{GS} = 0 V _{dc} , f = 1.0 MHz)	C _{rss}	—	—	0.8	pF
Forward Transadmittance (V _{DS} = 15 V _{dc} , V _{GS} = 0, f = 200 MHz)	Y _{fs}	3.2	—	—	mmhos
Equivalent Noise Voltage (V _{DS} = 15 V _{dc} , V _{GS} = 0, f = 25 Hz)	e _n	—	—	75	nV/√Hz
Noise Figure (V _{DS} = 15 V _{dc} , V _{GS} = 0 V, see Figures 1, 2, 3)	NF	—	—	2.5	dB

N-channel silicon field-effect transistors

BF245A; BF245B;
BF245C

FEATURES

- Interchangeability of drain and source connections
- Frequencies up to 700 MHz.

APPLICATIONS

- LF, HF and DC amplifiers.

DESCRIPTION

General purpose N-channel symmetrical junction field-effect transistors in a plastic TO-92 variant package.

CAUTION

The device is supplied in an antistatic package. The gate-source input must be protected against static discharge during transport or handling.

PINNING

PIN	SYMBOL	DESCRIPTION
1	d	drain
2	s	source
3	g	gate

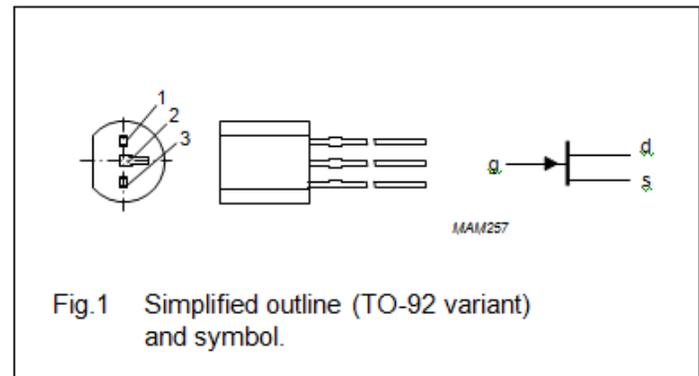


Fig. 1 Simplified outline (TO-92 variant) and symbol.

QUICK REFERENCE DATA

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
V_{DS}	drain-source voltage		-	-	± 30	V
V_{GSoff}	gate-source cut-off voltage	$I_D = 10 \text{ nA}$; $V_{DS} = 15 \text{ V}$	-0.25	-	-8	V
V_{GS0}	gate-source voltage	open drain	-	-	-30	V
I_{DSS}	drain current	$V_{DS} = 15 \text{ V}$; $V_{GS} = 0$				
	BF245A		2	-	6.5	mA
	BF245B		6	-	15	mA
	BF245C		12	-	25	mA
P_{tot}	total power dissipation	$T_{amb} = 75 \text{ }^\circ\text{C}$	-	-	300	mW
$ y_{fs} $	forward transfer admittance	$V_{DS} = 15 \text{ V}$; $V_{GS} = 0$; $f = 1 \text{ kHz}$; $T_{amb} = 25 \text{ }^\circ\text{C}$	3	-	6.5	mS
C_{rs}	reverse transfer capacitance	$V_{DS} = 20 \text{ V}$; $V_{GS} = -1 \text{ V}$; $f = 1 \text{ MHz}$; $T_{amb} = 25 \text{ }^\circ\text{C}$	-	1.1	-	pF

LIMITING VALUES

In accordance with the Absolute Maximum Rating System (IEC 134).

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
V_{DS}	drain-source voltage		–	± 30	V
V_{GDO}	gate-drain voltage	open source	–	–30	V
V_{GSO}	gate-source voltage	open drain	–	–30	V
I_D	drain current		–	25	mA
I_G	gate current		–	10	mA
P_{tot}	total power dissipation	up to $T_{amb} = 75\text{ }^{\circ}\text{C}$;	–	300	mW
		up to $T_{amb} = 90\text{ }^{\circ}\text{C}$; note 1	–	300	mW
T_{stg}	storage temperature		–65	+150	$^{\circ}\text{C}$
T_j	operating junction temperature		–	150	$^{\circ}\text{C}$

Note

1. Device mounted on a printed-circuit board, minimum lead length 3 mm, mounting pad for drain lead minimum 10 mm \times 10 mm.

THERMAL CHARACTERISTICS

SYMBOL	PARAMETER	CONDITIONS	VALUE	UNIT
$R_{th\ j-a}$	thermal resistance from junction to ambient	in free air	250	K/W
	thermal resistance from junction to ambient		200	K/W

STATIC CHARACTERISTICS $T_j = 25\text{ }^\circ\text{C}$; unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
$V_{(BR)GSS}$	gate-source breakdown voltage	$I_G = -1\text{ }\mu\text{A}$; $V_{DS} = 0$	-30	-	V
V_{GSoff}	gate-source cut-off voltage	$I_D = 10\text{ nA}$; $V_{DS} = 15\text{ V}$	-0.25	-8.0	V
V_{GS}	gate-source voltage	$I_D = 200\text{ }\mu\text{A}$; $V_{DS} = 15\text{ V}$			
	BF245A		-0.4	-2.2	V
	BF245B		-1.6	-3.8	V
	BF245C	-3.2	-7.5	V	
I_{DSS}	drain current	$V_{DS} = 15\text{ V}$; $V_{GS} = 0$; note 1			
	BF245A		2	6.5	mA
	BF245B		6	15	mA
	BF245C	12	25	mA	
I_{GSS}	gate cut-off current	$V_{GS} = -20\text{ V}$; $V_{DS} = 0$	-	-5	nA
		$V_{GS} = -20\text{ V}$; $V_{DS} = 0$; $T_j = 125\text{ }^\circ\text{C}$	-	-0.5	μA

Note

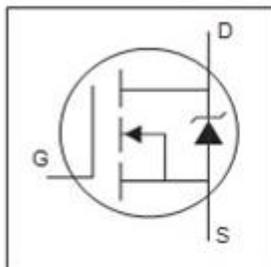
1. Measured under pulse conditions: $t_p = 300\text{ }\mu\text{s}$; $\delta \leq 0.02$.

DYNAMIC CHARACTERISTICSCommon source; $T_{amb} = 25\text{ }^\circ\text{C}$; unless otherwise specified.

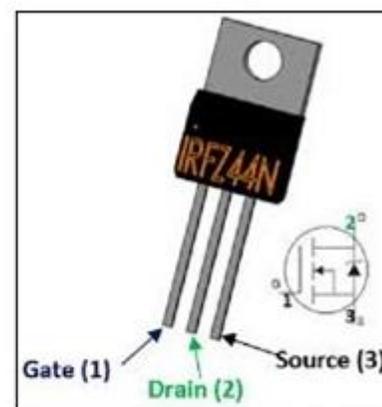
SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
C_{is}	input capacitance	$V_{DS} = 20\text{ V}$; $V_{GS} = -1\text{ V}$; $f = 1\text{ MHz}$	-	4	-	pF
C_{rs}	reverse transfer capacitance	$V_{DS} = 20\text{ V}$; $V_{GS} = -1\text{ V}$; $f = 1\text{ MHz}$	-	1.1	-	pF
C_{os}	output capacitance	$V_{DS} = 20\text{ V}$; $V_{GS} = -1\text{ V}$; $f = 1\text{ MHz}$	-	1.6	-	pF
g_{is}	input conductance	$V_{DS} = 15\text{ V}$; $V_{GS} = 0$; $f = 200\text{ MHz}$	-	250	-	μS
g_{os}	output conductance	$V_{DS} = 15\text{ V}$; $V_{GS} = 0$; $f = 200\text{ MHz}$	-	40	-	μS
$ y_{fs} $	forward transfer admittance	$V_{DS} = 15\text{ V}$; $V_{GS} = 0$; $f = 1\text{ kHz}$	3	-	6.5	mS
		$V_{DS} = 15\text{ V}$; $V_{GS} = 0$; $f = 200\text{ MHz}$	-	6	-	mS
$ y_{rs} $	reverse transfer admittance	$V_{DS} = 15\text{ V}$; $V_{GS} = 0$; $f = 200\text{ MHz}$	-	1.4	-	mS
$ y_{os} $	output admittance	$V_{DS} = 15\text{ V}$; $V_{GS} = 0$; $f = 1\text{ kHz}$	-	25	-	μS
f_{gfs}	cut-off frequency	$V_{DS} = 15\text{ V}$; $V_{GS} = 0$; $g_{fs} = 0.7$ of its value at 1 kHz	-	700	-	MHz
F	noise figure	$V_{DS} = 15\text{ V}$; $V_{GS} = 0$; $f = 100\text{ MHz}$; $R_G = 1\text{ k}\Omega$ (common source); input tuned to minimum noise	-	1.5	-	dB

MOSFET IRFZ 44N**IRFZ44NPbF**

HEXFET® Power MOSFET



$V_{DSS} = 55V$
$R_{DS(on)} = 17.5m\Omega$
$I_D = 49A$

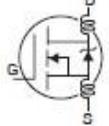
**Absolute Maximum Ratings**

	Parameter	Max.	Units
$I_D @ T_C = 25^\circ C$	Continuous Drain Current, $V_{GS} @ 10V$	49	A
$I_D @ T_C = 100^\circ C$	Continuous Drain Current, $V_{GS} @ 10V$	35	
I_{DM}	Pulsed Drain Current ①	160	
$P_D @ T_C = 25^\circ C$	Power Dissipation	94	W
	Linear Derating Factor	0.63	W/°C
V_{GS}	Gate-to-Source Voltage	± 20	V
I_{AR}	Avalanche Current ①	25	A
E_{AR}	Repetitive Avalanche Energy ①	9.4	mJ
dv/dt	Peak Diode Recovery dv/dt ③	5.0	V/ns
T_J	Operating Junction and	-55 to +175	°C
T_{STG}	Storage Temperature Range		
	Soldering Temperature, for 10 seconds		
	Mounting torque, 6-32 or M3 screw	10 lbf•in (1.1N•m)	

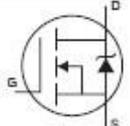
Thermal Resistance

	Parameter	Typ.	Max.	Units
$R_{\theta JC}$	Junction-to-Case	—	1.5	°C/W
$R_{\theta CS}$	Case-to-Sink, Flat, Greased Surface	0.50	—	
$R_{\theta JA}$	Junction-to-Ambient	—	62	

Electrical Characteristics @ $T_J = 25^\circ\text{C}$ (unless otherwise specified)

	Parameter	Min.	Typ.	Max.	Units	Conditions
$V_{(BR)DSS}$	Drain-to-Source Breakdown Voltage	55	—	—	V	$V_{GS} = 0V, I_D = 250\mu A$
$\Delta V_{(BR)DSS}/\Delta T_J$	Breakdown Voltage Temp. Coefficient	—	0.058	—	V/ $^\circ\text{C}$	Reference to $25^\circ\text{C}, I_D = 1\text{mA}$
$R_{DS(on)}$	Static Drain-to-Source On-Resistance	—	—	17.5	m Ω	$V_{GS} = 10V, I_D = 25A$ ④
$V_{GS(th)}$	Gate Threshold Voltage	2.0	—	4.0	V	$V_{DS} = V_{GS}, I_D = 250\mu A$
g_{fs}	Forward Transconductance	19	—	—	S	$V_{DS} = 25V, I_D = 25A$ ④
I_{DSS}	Drain-to-Source Leakage Current	—	—	25	μA	$V_{DS} = 55V, V_{GS} = 0V$
		—	—	250		$V_{DS} = 44V, V_{GS} = 0V, T_J = 150^\circ\text{C}$
I_{GSS}	Gate-to-Source Forward Leakage	—	—	100	nA	$V_{GS} = 20V$
	Gate-to-Source Reverse Leakage	—	—	-100		$V_{GS} = -20V$
Q_g	Total Gate Charge	—	—	63	nC	$I_D = 25A$
Q_{gs}	Gate-to-Source Charge	—	—	14		$V_{DS} = 44V$
Q_{gd}	Gate-to-Drain ("Miller") Charge	—	—	23		$V_{GS} = 10V$, See Fig. 6 and 13
$t_{d(on)}$	Turn-On Delay Time	—	12	—	ns	$V_{DD} = 28V$
t_r	Rise Time	—	60	—		$I_D = 25A$
$t_{d(off)}$	Turn-Off Delay Time	—	44	—		$R_G = 12\Omega$
t_f	Fall Time	—	45	—		$V_{GS} = 10V$, See Fig. 10 ④
L_D	Internal Drain Inductance	—	4.5	—	nH	Between lead, 6mm (0.25in.) from package and center of die contact
L_S	Internal Source Inductance	—	7.5	—		
C_{iss}	Input Capacitance	—	1470	—	pF	$V_{GS} = 0V$
C_{oss}	Output Capacitance	—	360	—		$V_{DS} = 25V$
C_{rss}	Reverse Transfer Capacitance	—	88	—		$f = 1.0\text{MHz}$, See Fig. 5
E_{AS}	Single Pulse Avalanche Energy②	—	530③	150⑥	mJ	$I_{AS} = 25A, L = 0.47\text{mH}$

Source-Drain Ratings and Characteristics

	Parameter	Min.	Typ.	Max.	Units	Conditions
I_S	Continuous Source Current (Body Diode)	—	—	49	A	MOSFET symbol showing the integral reverse p-n junction diode. 
I_{SM}	Pulsed Source Current (Body Diode)①	—	—	160		
V_{SD}	Diode Forward Voltage	—	—	1.3	V	$T_J = 25^\circ\text{C}, I_S = 25A, V_{GS} = 0V$ ④
t_{rr}	Reverse Recovery Time	—	63	95	ns	$T_J = 25^\circ\text{C}, I_F = 25A$
Q_{rr}	Reverse Recovery Charge	—	170	260	nC	$di/dt = 100A/\mu s$ ④
t_{on}	Forward Turn-On Time	Intrinsic turn-on time is negligible (turn-on is dominated by L_S+L_D)				

Notes:

- ① Repetitive rating; pulse width limited by max. junction temperature. (See fig. 11)
- ② Starting $T_J = 25^\circ\text{C}$, $L = 0.48\text{mH}$
 $R_G = 25\Omega$, $I_{AS} = 25A$. (See Figure 12)

③ $I_{SD} \leq 25A$, $di/dt \leq 230A/\mu s$, $V_{DD} \leq V_{(BR)DSS}$,
 $T_J \leq 175^\circ\text{C}$

④ Pulse width $\leq 400\mu s$; duty cycle $\leq 2\%$.

⑤ This is a typical value at device destruction and represents operation outside rated limits.

⑥ This is a calculated value limited to $T_J = 175^\circ\text{C}$.

B. RULES

RULES TO BE FOLLOWED WHILE OPERATING THE REGULATED POWERB SUPPLY (RPS) & CRO :

The flowing rules should be followed before switch ON the Regulated Power Supply,

1. Initially Keep the *voltage Course & Voltage fine controls* of RPS at minimum position. Later (After switch ON the RPS) can vary these controls slowly to get the required voltage.
2. Always keep the Current Limit control at maximum position, Otherwise the display can shows the constant voltage instead of varying.

Trouble shooting while operating the rps :

The following trouble shooting can done while operating the RPS,

During connecting the RPS to the circuit and varying the Voltage Course & Voltage Fine Controls, If it displays the voltage as constant or above 30V then it can said that either the circuit is shorted OR the Current Limit control is not kept at maximum position. This problem can solve to prevent the circuit from shorted and by keeping the Current Limit control at maximum.

RULES TO OPERATE THE CRO:

The following rules should be follows before operate the CRO.

1. Keep the following controls at middle position or vary until the electron beam is generated.
 - a) INTENSITY
 - b) FOCUS
 - c)  (Horizontal position)
(Horizontal position common for both channels)
 - d)  Vertical Position (Vertical position individual per each channel)
 - e) LEVEL (Trigger Level)
2. Keep the following controls at maximum position.
 - (a). **VARIABLE** controls of VOLTS/DIV switch in both channels.
 - (b). **SWP.VAR** (Sweep Variation)
3. Keep the following switches at releasing mode.
 - a) $\times 10$ MAG
 - b) TRIG.ALT
 - c) SLOPE
 - d) ALT/CHOP
 - e) CH2 INV
4. Initially should keep the **TIME/DIV** control at 1mS position, later can change this switch depending upon our requirement , i.e. if we can't get the signal clearly on the CRT, then we can vary this switch until to get the signal.
5. Set the channel selector control **MODE** at the appropriate position i.e. if we want to see the signal in channel1, set this control at CH1, in channel2 set at CH2, in both channels set at DUAL. To add the signals (algebraically sum or difference) available in both channels set at ADD.
6. **AC/GND/DC**: Before setting the signals on CRT, first we should keep the electron beam on referenceline. To set this beam on reference line, keep this control at GND positio and then vary vertical position control until to get the beam on the reference line. After that to see the applied signals, keep this control at AC or DC positions.
7. Always keep the **TRIGGER MODE** control at AUTO position.
8. Keep the **SOURCE** control at approximate channel. It means if MODE control is selected to CH1, then the SOURCE control should select to CH1. If MODE control at CH2, set the SOURCE control at CH2. If MODE control at DUAL or ADD, set the SOURCE control either at CH1 or CH2.

RULES FOR HOW TO WRITE THE OBSERVATION AND RECORDS:

1. Make the top & right margins in each page of right side.
2. In top margin make the headings as Experiment No., date and name of the experiment.
3. Circuit diagrams, tabular columns, expected graphs and parameters/calculations should write on leftside page (even No. page) .
4. Aim, apparatus, components, theory, procedure, applications, conclusion and result should write on right side page (Odd No. Page) .
5. Headings should underline with any other ink except red, orange and green.
6. The every new experiment should start with right side page.
7. leave the half of the page under the heading of *theory*.